

VINcoMNPC X4
1200 V/400 A
Features

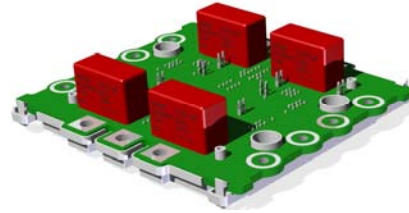
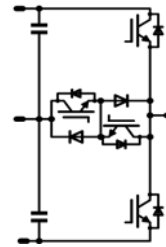
- Mixed voltage NPC
- Low inductive
- High power screw interface
- Integrated DC-snubber capacitors

Target Applications

- Solar inverter
- UPS
- High speed motor drive

Types

- 70-W212NMA400SC-M209P

VINco X4 housing

Schematic


Maximum Ratings

 $T_j=25^{\circ}\text{C}$, unless otherwise specified

Parameter	Symbol	Condition	Value	Unit
half bridge IGBT (T1 , T4)				
Collector-emitter break down voltage	V_{CE}		1200	V
DC collector current	I_C	$T_j=T_{j,max}$ $T_n=80^{\circ}\text{C}$ $T_c=80^{\circ}\text{C}$	338 439	A
Repetitive peak collector current	$I_{C,pulse}$	t_p limited by $T_{j,max}$	1200	A
Power dissipation per IGBT	P_{tot}	$T_j=T_{j,max}$ $T_n=80^{\circ}\text{C}$ $T_c=80^{\circ}\text{C}$	729 1104	W
Gate-emitter peak voltage	V_{GE}		± 20	V
Short circuit ratings	t_{SC} V_{CC}	$T_j \leq 150^{\circ}\text{C}$ $V_{GE} = 15\text{V}$	10 800	μs V
Turn off safe operating area (RBSOA)	$I_{c,max}$	$V_{CE,max} = 1200\text{V}$ $T_{vj,max} = 150^{\circ}\text{C}$	800	A
Maximum Junction Temperature	$T_{j,max}$		175	$^{\circ}\text{C}$
neutral point FWD (D2 , D3)				
Peak Repetitive Reverse Voltage	V_{RRM}	$T_j=25^{\circ}\text{C}$	600	V
DC forward current	I_F	$T_j=T_{j,max}$ $T_n=80^{\circ}\text{C}$ $T_c=80^{\circ}\text{C}$	309 415	A
Surge forward current	I_{FSM}	$t_p = 10 \text{ ms, sine halfwave}$ $T_{vj} < 150^{\circ}\text{C}$	890	A
I ² t-value	I^2t		3960	A ² s
Repetitive peak forward current	I_{FRM}	$t_p = 1 \text{ ms}$ $T_{vj} < 150^{\circ}\text{C}$	800	A
Power dissipation per FWD	P_{tot}	$T_j=T_{j,max}$ $T_n=80^{\circ}\text{C}$ $T_c=80^{\circ}\text{C}$	421 637	W
Maximum Junction Temperature	$T_{j,max}$		175	$^{\circ}\text{C}$

Maximum Ratings

 $T_j=25^{\circ}\text{C}$, unless otherwise specified

Parameter	Symbol	Condition	Value	Unit
neutral point IGBT (T2 , T3)				
Collector-emitter break down voltage	V_{CE}		600	V
DC collector current	I_C	$T_j=T_{jmax}$ $T_h=80^{\circ}\text{C}$ $T_c=80^{\circ}\text{C}$	329 430	A
Repetitive peak collector current	I_{Cpuls}	t_p limited by T_{jmax}	1200	A
Power dissipation per IGBT	P_{tot}	$T_j=T_{jmax}$ $T_h=80^{\circ}\text{C}$ $T_c=80^{\circ}\text{C}$	574 870	W
Gate-emitter peak voltage	V_{GE}		± 20	V
Short circuit ratings	t_{SC}	$T_j \leq 150^{\circ}\text{C}$	6	μs
	V_{CC}	$V_{GE}=15\text{V}$	360	V
Turn off safe operating area (RBSOA)	I_{cmax}	$V_{CE max} = 1200\text{V}$ $T_{vj max} = 150^{\circ}\text{C}$	800	A
Maximum Junction Temperature	T_{jmax}		175	$^{\circ}\text{C}$

half bridge FWD (D1 , D4)

Peak Repetitive Reverse Voltage	V_{RRM}	$T_j=25^{\circ}\text{C}$	1200	V
DC forward current	I_F	$T_j=T_{jmax}$ $T_h=80^{\circ}\text{C}$ $T_c=80^{\circ}\text{C}$	270 356	A
Surge forward current	I_{FSM}	$t_p=10\text{ms}$, sin 180° $T_j=150^{\circ}\text{C}$	2200	A
I ² t-value	I^2t			
Repetitive peak forward current	I_{FRM}	t_p limited by T_{jmax}	1200	A
Power dissipation per FWD	P_{tot}	$T_j=T_{jmax}$ $T_h=80^{\circ}\text{C}$ $T_c=80^{\circ}\text{C}$	540 818	W
Maximum Junction Temperature	T_{jmax}		175	$^{\circ}\text{C}$

DC link Capacitor

Max.DC voltage	V_{MAX}	$T_c=100^{\circ}\text{C}$	630	V
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General Module Properties

Material of module baseplate			Cu	
Material of internal insulation			Al_2O_3	

Thermal Properties

Storage temperature	T_{stg}		$-40 \dots +125$	$^{\circ}\text{C}$
Operation temperature under switching condition	T_{op}		$-40 \dots +(T_{jmax} - 25)$	$^{\circ}\text{C}$

Insulation Properties

Insulation voltage	V_{is}	$t=2\text{s}$ DC voltage	4000	V
Creepage distance			min 12,7	mm
Clearance			min 12,7	mm
Comparative tracking index	CTI		>200	

Characteristic Values

Parameter	Symbol	Conditions				Value			Unit	
		V_{GE} [V] or V_{ES} [V]	V_i [V] or V_{CE} [V] or V_{DS} [V]	I_c [A] or I_f [A] or I_b [A]	T_j	Min	Typ	Max		
half bridge IGBT (T1 , T4)										
Gate emitter threshold voltage	$V_{GE(th)}$	$V_{CE}=V_{GE}$			0,0152	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	5 1,5	5,8 1,97 2,23	6,5 2,4	V
Collector-emitter saturation voltage	$V_{CE(sat)}$		15		400	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$				V
Collector-emitter cut-off current incl. FWD	I_{CES}		0	1200		$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$			0,6	mA
Gate-emitter leakage current	I_{GES}		20	0		$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$			3000	nA
Integrated Gate resistor	R_{gint}							1,88		Ω
Turn-on delay time	$t_{d(on)}$					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		235 247		ns
Rise time	t_r					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		46 55		
Turn-off delay time	$t_{d(off)}$	$R_{goff}=1\ \Omega$ $R_{gon}=1\ \Omega$	± 15	350	400	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		292 354		
Fall time	t_f					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		55 92		
Turn-on energy loss per pulse	E_{on}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		7,95 12,30		mWs
Turn-off energy loss per pulse	E_{off}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		13,25 22,08		
Input capacitance	C_{ies}							24600		pF
Output capacitance	C_{oss}	$f=1\text{MHz}$	0	25		$T_j=25^\circ\text{C}$		1620		
Reverse transfer capacitance	C_{rss}							1380		
Gate charge	Q_{gate}		± 15	960	400	$T_j=25^\circ\text{C}$		2030		nC
Thermal resistance chip to heatsink per chip	R_{thJH}	Thermal grease thickness $\leq 50\mu\text{m}$ $\lambda = 1\ \text{W/mK}$						0,13		K/W
Thermal resistance chip to case per chip	R_{thJC}							0,09		
neutral point FWD (D2 , D3)										
FWD forward voltage	V_F				400	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	1,2	1,67 1,56	2,2	V
Peak reverse recovery current	I_{RRM}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		204 262		A
Reverse recovery time	t_{rr}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		183 295		ns
Reverse recovered charge	Q_{rr}	$R_{gon}=1\ \Omega$	± 15	350	400	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		17 33		μC
Peak rate of fall of recovery current	$di(\text{rec})_{\text{max}}/dt$					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		3129 1705		A/ μs
Reverse recovered energy	E_{rec}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		3,78 7,44		mWs
Thermal resistance chip to heatsink per chip	R_{thJH}	Thermal grease thickness $\leq 50\mu\text{m}$ $\lambda = 1\ \text{W/mK}$						0,23		K/W
Thermal resistance chip to case per chip	R_{thJC}							0,15		
neutral point IGBT (T2 , T3)										
Gate emitter threshold voltage	$V_{GE(th)}$	$V_{CE}=V_{GE}$			0,0064	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	5	5,8	6,5	V
Collector-emitter saturation voltage	$V_{CE(sat)}$		15		400	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	1	1,56 1,80	2,2	V
Collector-emitter cut-off incl FWD	I_{CES}		0	600		$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$			0,1	mA
Gate-emitter leakage current	I_{GES}		20	0		$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$			3000	nA
Integrated Gate resistor	R_{gint}							0,5		Ω
Turn-on delay time	$t_{d(on)}$					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		201 204		ns
Rise time	t_r					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		29 32		
Turn-off delay time	$t_{d(off)}$	$R_{goff}=1\ \Omega$ $R_{gon}=1\ \Omega$	± 15	350	400	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		248 272		
Fall time	t_f					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		71 88		
Turn-on energy loss per pulse	E_{on}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		3,93 5,61		mWs
Turn-off energy loss per pulse	E_{off}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		10,49 14,07		
Input capacitance	C_{ies}							24640		pF
Output capacitance	C_{oss}	$f=1\text{MHz}$	0	25		$T_j=25^\circ\text{C}$		1536		
Reverse transfer capacitance	C_{rss}							732		
Gate charge	Q_{gate}		± 15	480	400	$T_j=25^\circ\text{C}$		2480		nC
Thermal resistance chip to heatsink per chip	R_{thJH}	Thermal grease thickness $\leq 50\mu\text{m}$ $\lambda = 1\ \text{W/mK}$						0,17		K/W
Thermal resistance chip to case per chip	R_{thJC}							0,11		

Characteristic Values

Parameter	Symbol	Conditions				Value			Unit
		V_{GS} [V] or V_{DS} [V]	V_r [V] or V_{CE} [V] or V_{DS} [V]	I_c [A] or I_r [A] or I_b [A]	T_j	Min	Typ	Max	
half bridge FWD (D1 , D4)									
FWD forward voltage	V_F			400		$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	1 2,29 2,37	3	V
Reverse leakage current	I_r		1200			$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$		480	μA
Peak reverse recovery current	I_{RRM}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	410 521		A
Reverse recovery time	t_{rr}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	63 149		ns
Reverse recovered charge	Q_{rr}	Rgon=1 Ω	± 15	350	400	$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	24 49		μC
Peak rate of fall of recovery current	$di(\text{rec})_{\text{max}}/dt$					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	18915 15110		A/ μs
Reverse recovery energy	E_{rec}					$T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	5,79 12,71		mWs
Thermal resistance chip to heatsink per chip	R_{thJH}	Thermal grease thickness $\leq 50\mu\text{m}$						0,18	K/W
Thermal resistance chip to case per chip	R_{thJC}	$\lambda = 1 \text{ W/mK}$						0,12	
DC link Capacitor									
C value	C							2 * 0,68	μF
Stray inductance of on board capacitors	ESL							26/2	nH
Series resistance of on board capacitors	ESR							14/2	m Ω
Thermistor									
Rated resistance	R					$T_j=25^\circ\text{C}$		22000	Ω
Deviation of R_{100}	$\Delta R/R$	$R_{100}=1486 \Omega$				$T_j=100^\circ\text{C}$	-5	5	%
Power dissipation	P					$T_j=25^\circ\text{C}$		200	mW
Power dissipation constant						$T_j=25^\circ\text{C}$		2	mW/K
B-value	$B_{(25/50)}$	Tol. $\pm 3\%$				$T_j=25^\circ\text{C}$		3950	K
B-value	$B_{(25/100)}$	Tol. $\pm 3\%$				$T_j=25^\circ\text{C}$		3996	K
Vincotech NTC Reference						$T_j=25^\circ\text{C}$		B	
Module Properties									
Module inductance (from chips to PCB)	L_{SCE}							5	nH
Module inductance (from PCB to PCB using Intercon board)	L_{SCE}							3	nH
Resistance of Intercon boards (from PCB to PCB using Intercon board)	R_{CC1+EE}					$T_c=25^\circ\text{C}$, per switch		1,5	m Ω
Mounting torque	M	Screw M4 - mounting according to valid application note FSWB1-4TY-M-*-HI					2	2,2	Nm
Mounting torque	M	Screw M5 - mounting according to valid application note FSWB1-4TY-M-*-HI					4	6	Nm
Terminal connection torque	M	Screw M6 - mounting according to valid application note FSWB1-4TY-M-*-HI					2,5	5	Nm
Weight	G							710	g

Buck

half bridge IGBT and neutral point FWD

Figure 1 IGBT

Typical output characteristics

$$I_C = f(V_{CE})$$

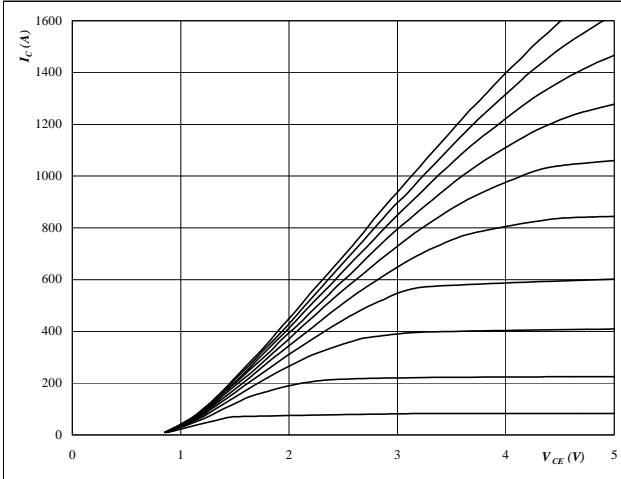

At
 $t_p = 350 \mu s$
 $T_j = 25 \text{ } ^\circ C$
 V_{GE} from 8 V to 17 V in steps of 1 V

Figure 2 IGBT

Typical output characteristics

$$I_C = f(V_{CE})$$

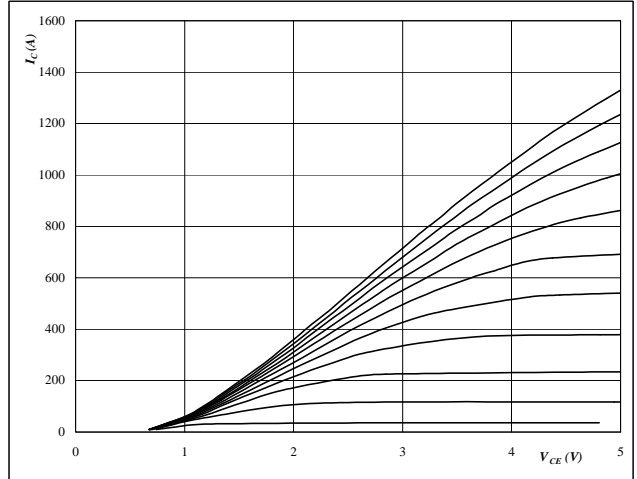
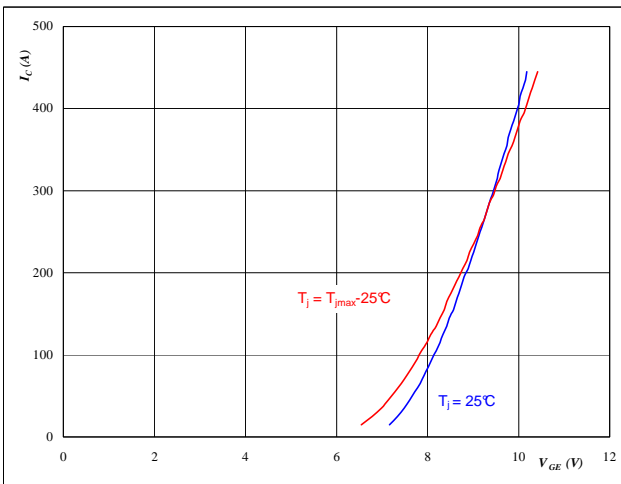

At
 $t_p = 350 \mu s$
 $T_j = 125 \text{ } ^\circ C$
 V_{GE} from 7 V to 17 V in steps of 1 V

Figure 3 IGBT

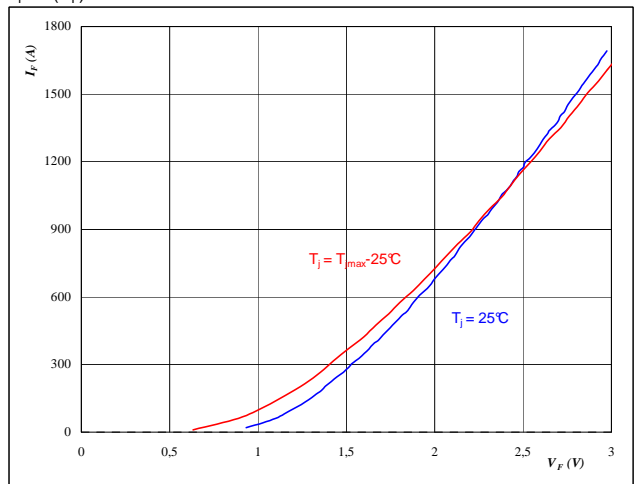
Typical transfer characteristics

$$I_C = f(V_{GE})$$


At
 $t_p = 350 \mu s$
 $V_{CE} = 10 \text{ V}$
Figure 4 FWD

Typical FWD forward current as a function of forward voltage

$$I_F = f(V_F)$$


At
 $t_p = 350 \mu s$

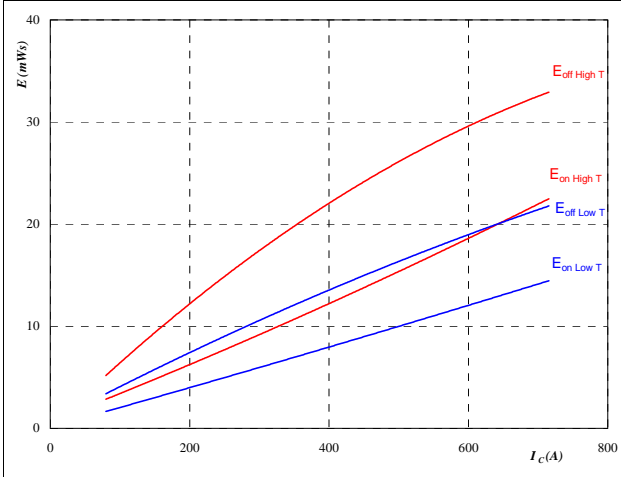
Buck

half bridge IGBT and neutral point FWD

Figure 5 IGBT

Typical switching energy losses
 as a function of collector current

$$E = f(I_C)$$



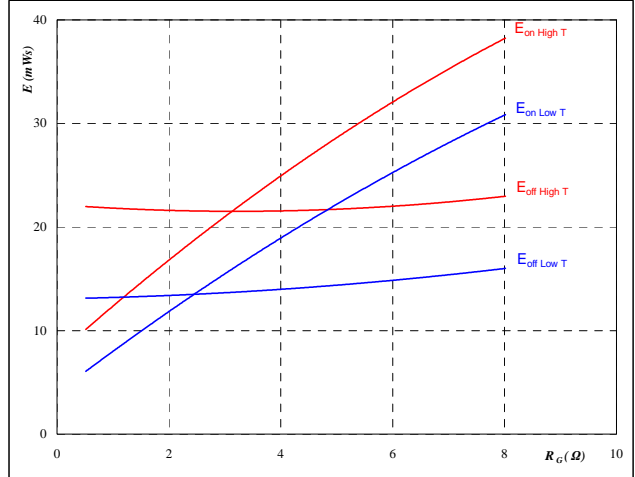
With an inductive load at

$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω
$R_{goff} =$	1	Ω

Figure 6 IGBT

Typical switching energy losses
 as a function of gate resistor

$$E = f(R_G)$$



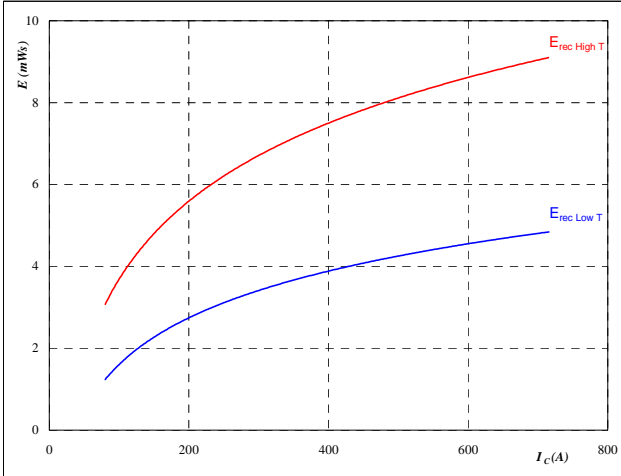
With an inductive load at

$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$I_C =$	400	A

Figure 7 FWD

Typical reverse recovery energy loss
 as a function of collector current

$$E_{rec} = f(I_C)$$



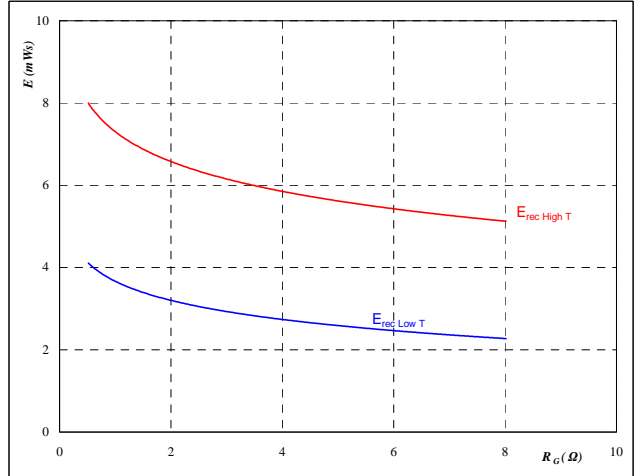
With an inductive load at

$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω

Figure 8 FWD

Typical reverse recovery energy loss
 as a function of gate resistor

$$E_{rec} = f(R_G)$$



With an inductive load at

$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$I_C =$	400	A

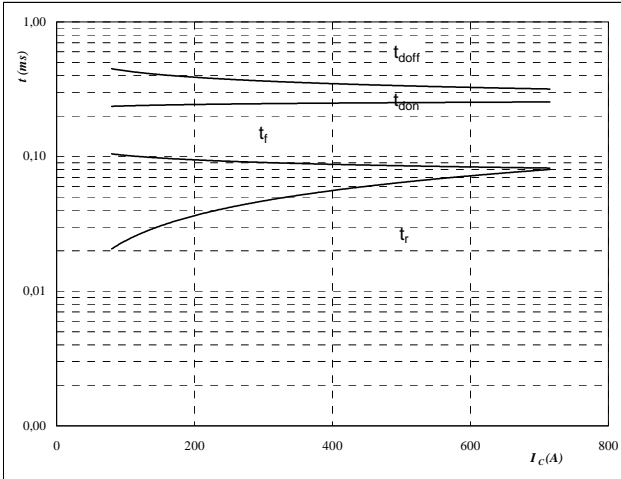
Buck

half bridge IGBT and neutral point FWD

Figure 9 IGBT

Typical switching times as a function of collector current

$t = f(I_C)$



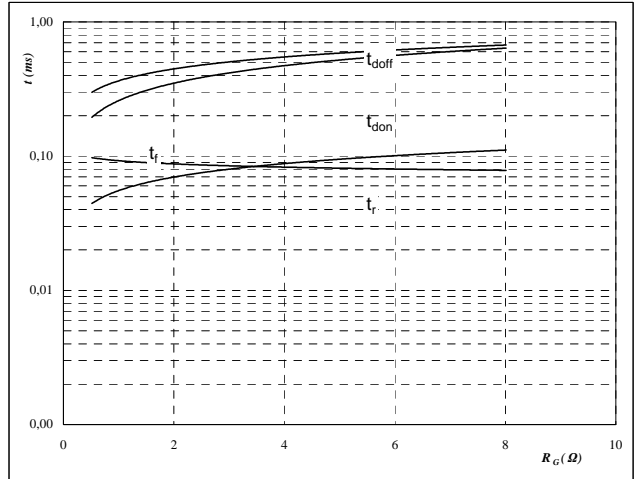
With an inductive load at

$T_J =$	125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1,0	Ω
$R_{goff} =$	1,0	Ω

Figure 10 IGBT

Typical switching times as a function of gate resistor

$t = f(R_G)$



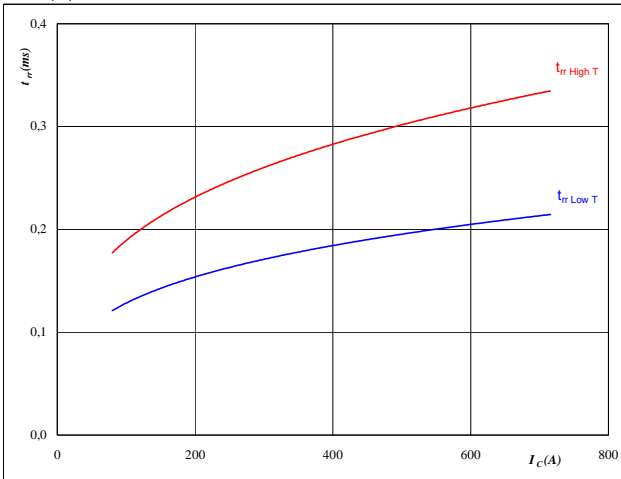
With an inductive load at

$T_J =$	125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$I_C =$	400	A

Figure 11 FWD

Typical reverse recovery time as a function of collector current

$t_{rr} = f(I_C)$



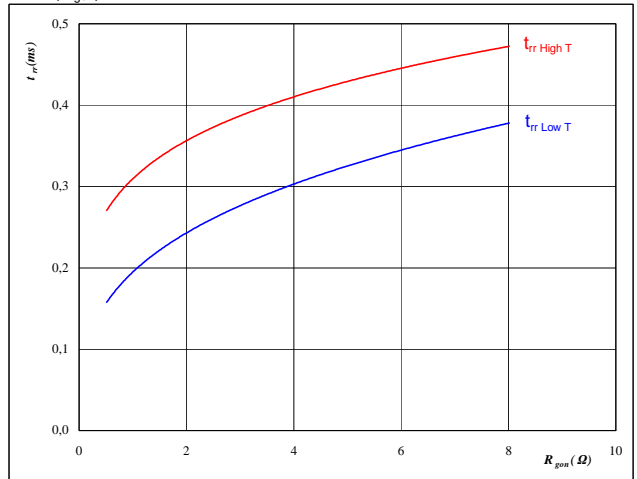
At

$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω

Figure 12 FWD

Typical reverse recovery time as a function of IGBT turn on gate resistor

$t_{rr} = f(R_{gon})$



At

$T_J =$	25/125	°C
$V_R =$	350	V
$I_F =$	400	A
$V_{GE} =$	±15	V

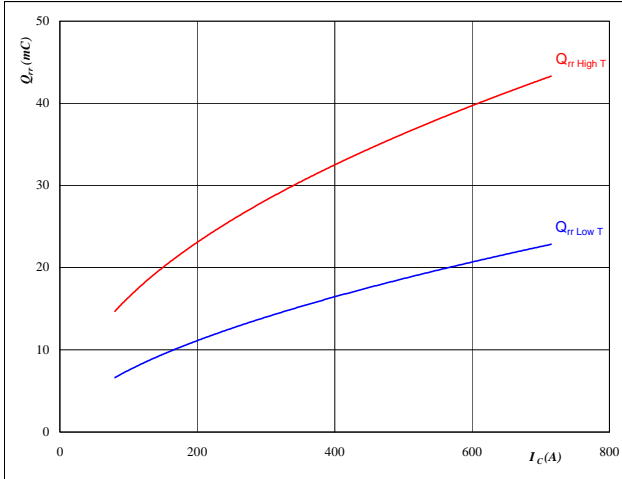
Buck

half bridge IGBT and neutral point FWD

Figure 13 FWD

Typical reverse recovery charge as a function of collector current

$$Q_{rr} = f(I_C)$$

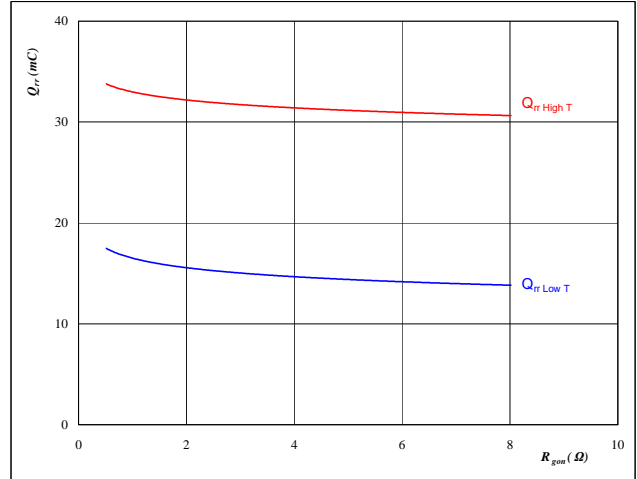

At

$T_j =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω

Figure 14 FWD

Typical reverse recovery charge as a function of IGBT turn on gate resistor

$$Q_{rr} = f(R_{gon})$$

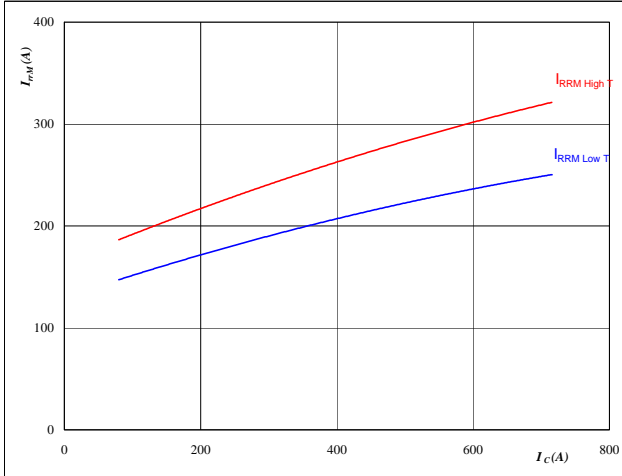

At

$T_j =$	25/125	°C
$V_R =$	350	V
$I_F =$	400	A
$V_{GE} =$	±15	V

Figure 15 FWD

Typical reverse recovery current as a function of collector current

$$I_{RRM} = f(I_C)$$

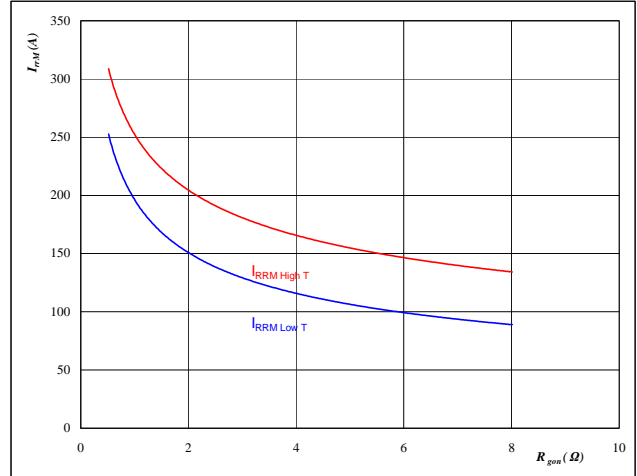

At

$T_j =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω

Figure 16 FWD

Typical reverse recovery current as a function of IGBT turn on gate resistor

$$I_{RRM} = f(R_{gon})$$


At

$T_j =$	25/125	°C
$V_R =$	350	V
$I_F =$	400	A
$V_{GE} =$	±15	V

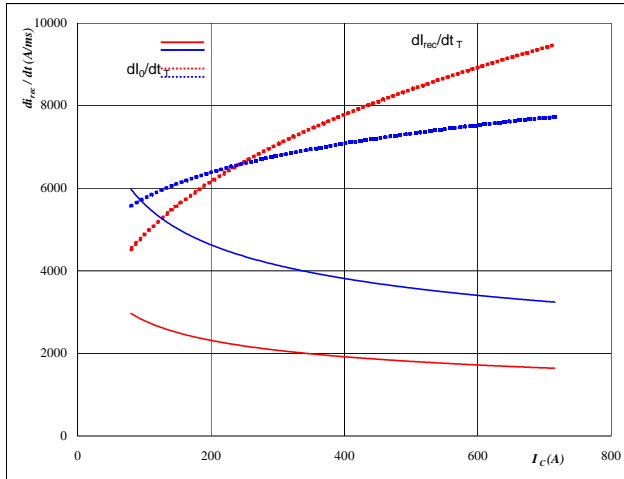
Buck

half bridge IGBT and neutral point FWD

Figure 17 FWD

Typical rate of fall of forward and reverse recovery current as a function of collector current

$$di_o/dt, di_{rec}/dt = f(I_c)$$

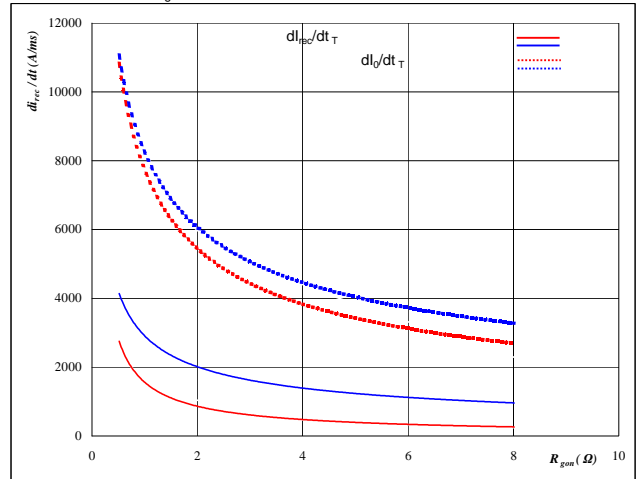


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 1 \text{ } \Omega$

Figure 18 FWD

Typical rate of fall of forward and reverse recovery current as a function of IGBT turn on gate resistor

$$di_o/dt, di_{rec}/dt = f(R_{gon})$$

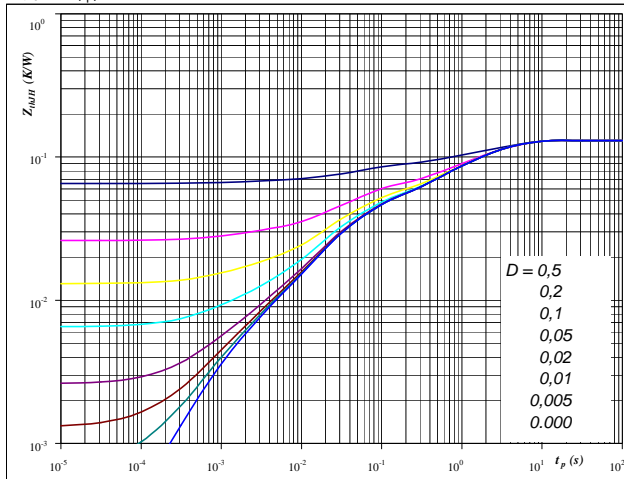


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 400 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$

Figure 19 IGBT

IGBT transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$



At
 $D = t_p / T$
 $R_{thJH} = 0,13 \text{ K/W}$

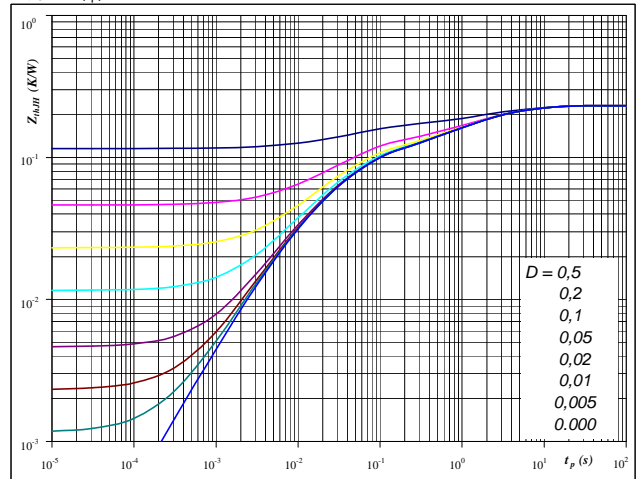
IGBT thermal model values

Thermal grease	
R (C/W)	Tau (s)
0,06	2,5E+00
0,03	4,7E-01
0,03	3,9E-02
0,01	1,2E-02
0,00	1,2E-03

Figure 20 FWD

FWD transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$



At
 $D = t_p / T$
 $R_{thJH} = 0,23 \text{ K/W}$

FWD thermal model values

Thermal grease	
R (C/W)	Tau (s)
0,05	5,2E+00
0,07	1,1E+00
0,02	2,0E-01
0,06	4,6E-02
0,02	1,7E-02

Buck

half bridge IGBT and neutral point FWD

Figure 21 IGBT

Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$

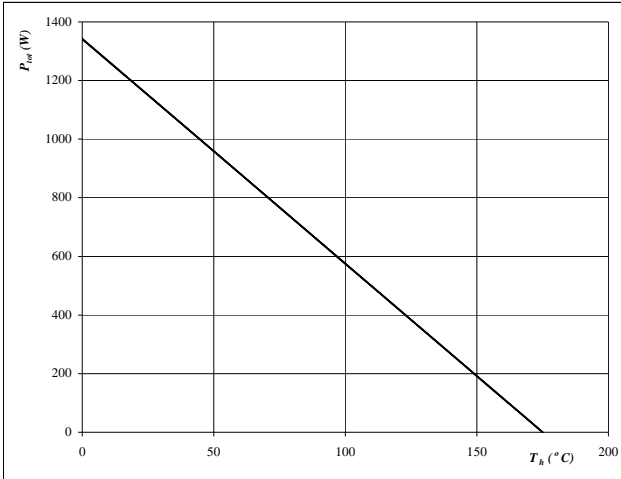

At
 $T_j = 175$ °C

Figure 22 IGBT

Collector current as a function of heatsink temperature

$$I_C = f(T_h)$$

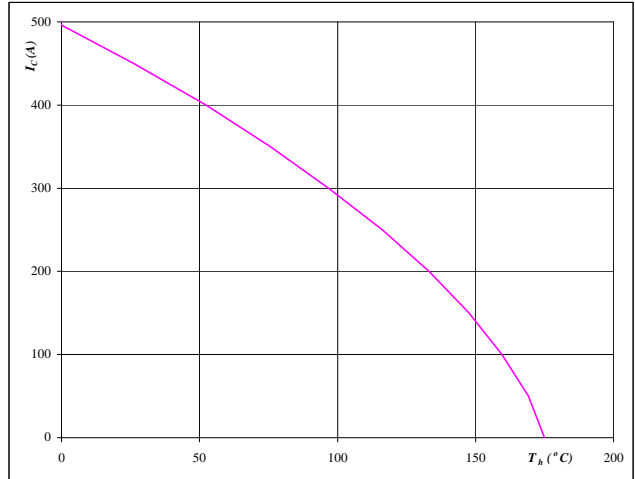

At
 $T_j = 175$ °C
 $V_{GE} = 15$ V

Figure 23 FWD

Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$

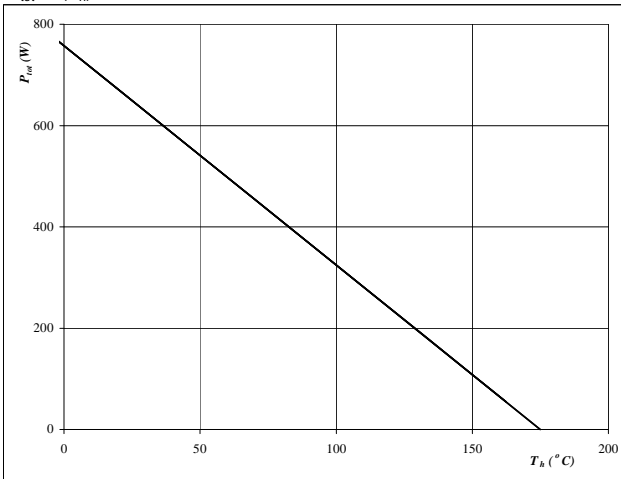
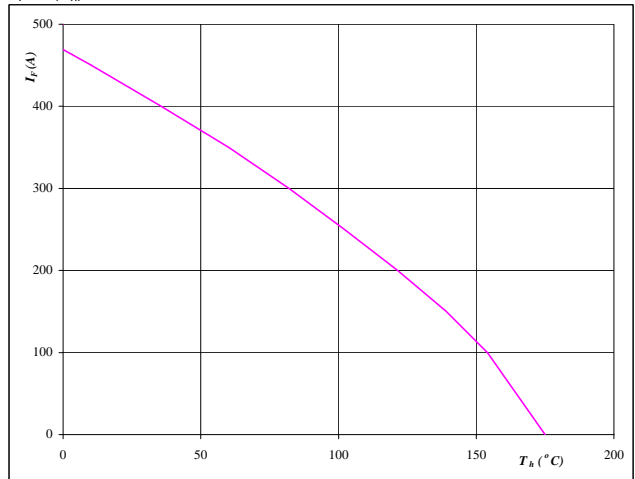

At
 $T_j = 175$ °C

Figure 24 FWD

Forward current as a function of heatsink temperature

$$I_F = f(T_h)$$


At
 $T_j = 175$ °C

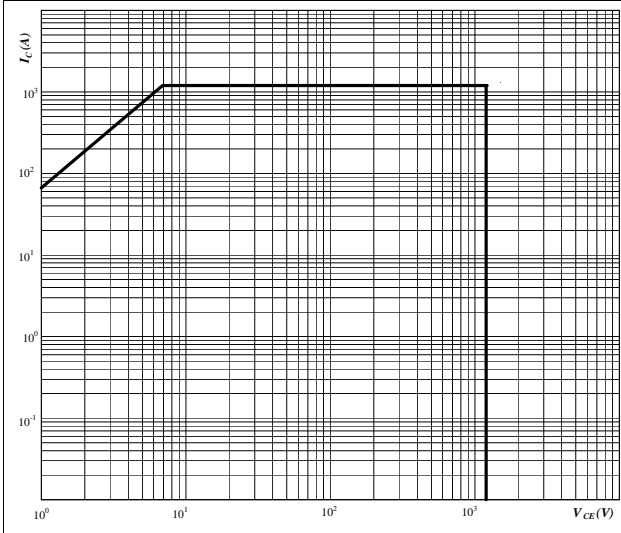
Buck

half bridge IGBT and neutral point FWD

Figure 25 IGBT

Safe operating area as a function of collector-emitter voltage

$$I_C = f(V_{CE})$$

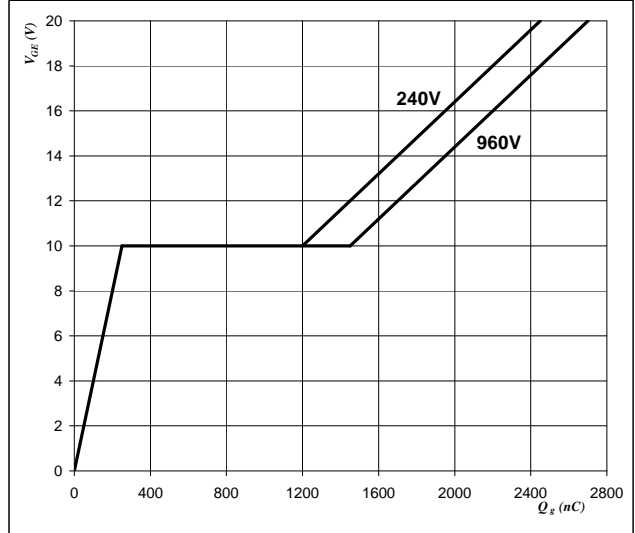


At
 D = single pulse
 Th = 80 °C
 V_{GE} = ±15 V
 T_j = T_{jmax} °C

Figure 26 IGBT

Gate voltage vs Gate charge

$$V_{GE} = f(Q_g)$$

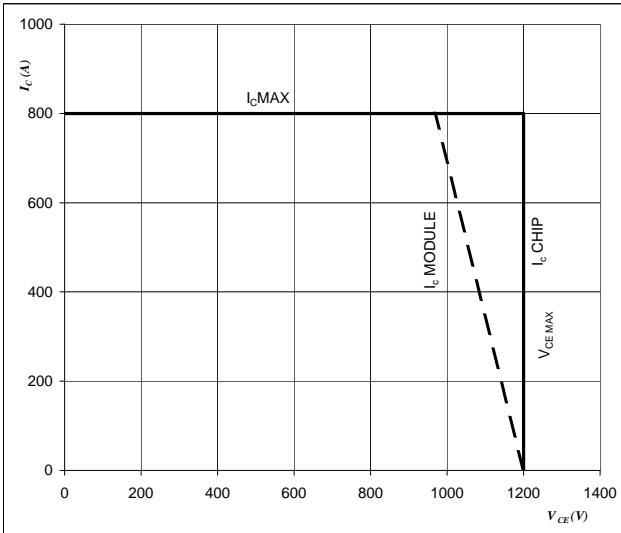


At
 I_C = 400 A

Figure 27 IGBT

Reverse bias safe operating area

$$I_C = f(V_{CE})$$



At
 T_j = T_{jmax} - 25 °C
 U_{ocminus} = U_{ccplus}
 Switching mode : 3 level switching

Boost

neutral point IGBT and half bridge FWD

Figure 1 IGBT

Typical output characteristics

$$I_C = f(V_{CE})$$

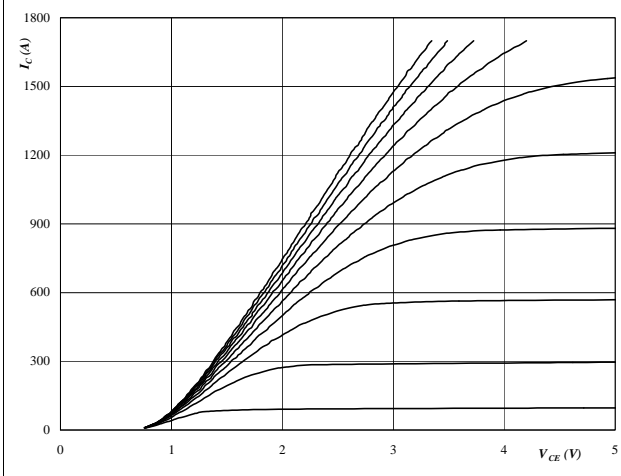

At
 $t_p = 350 \mu s$
 $T_j = 25 \text{ } ^\circ C$
 V_{GE} from 8 V to 17 V in steps of 1 V

Figure 2 IGBT

Typical output characteristics

$$I_C = f(V_{CE})$$

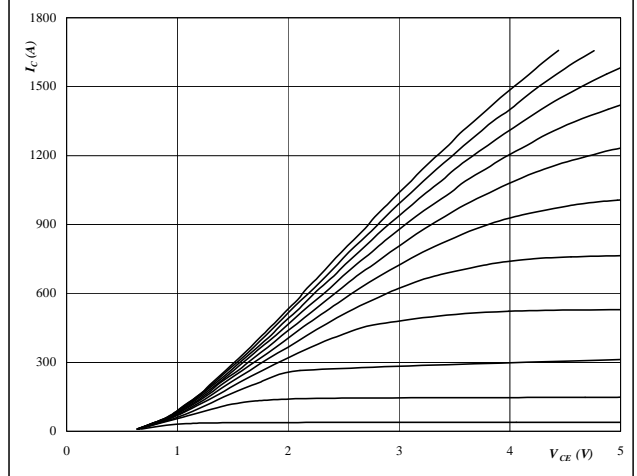
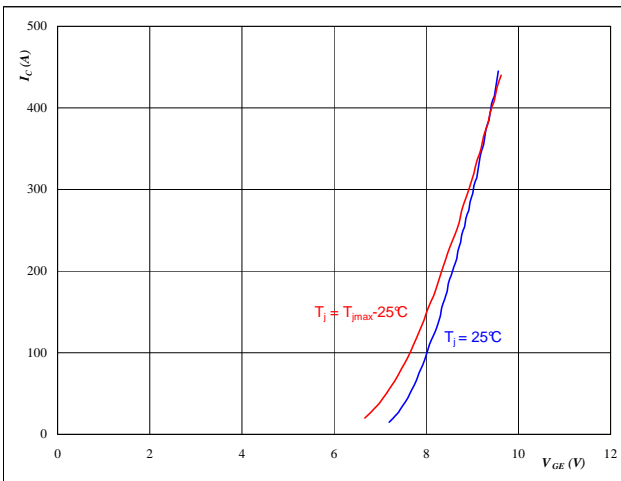

At
 $t_p = 350 \mu s$
 $T_j = 125 \text{ } ^\circ C$
 V_{GE} from 7 V to 17 V in steps of 1 V

Figure 3 IGBT

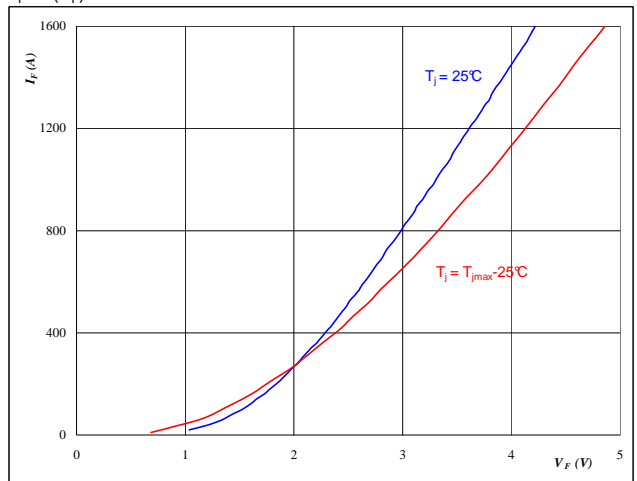
Typical transfer characteristics

$$I_C = f(V_{GE})$$


At
 $t_p = 350 \mu s$
 $V_{CE} = 0 \text{ V}$
Figure 4 FWD

Typical FWD forward current as a function of forward voltage

$$I_F = f(V_F)$$


At
 $t_p = 350 \mu s$

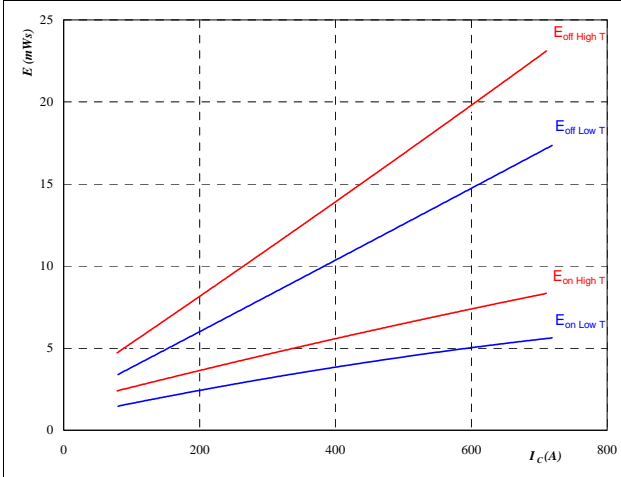
Boost

neutral point IGBT and half bridge FWD

Figure 5 IGBT

Typical switching energy losses
 as a function of collector current

$$E = f(I_C)$$

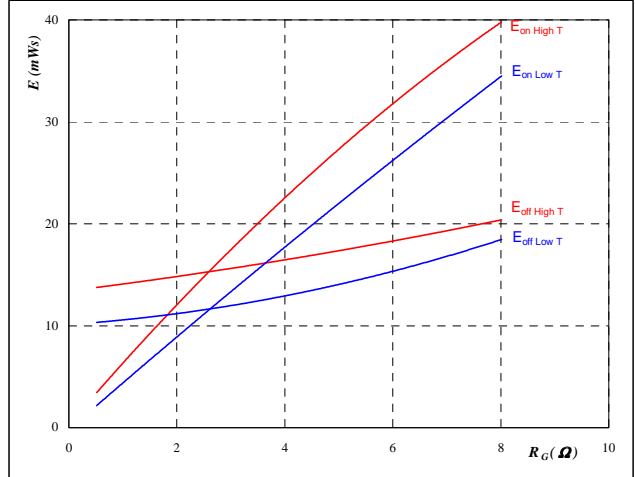


With an inductive load at

 $T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 1 \text{ } \Omega$
 $R_{goff} = 1 \text{ } \Omega$
Figure 6 IGBT

Typical switching energy losses
 as a function of gate resistor

$$E = f(R_G)$$

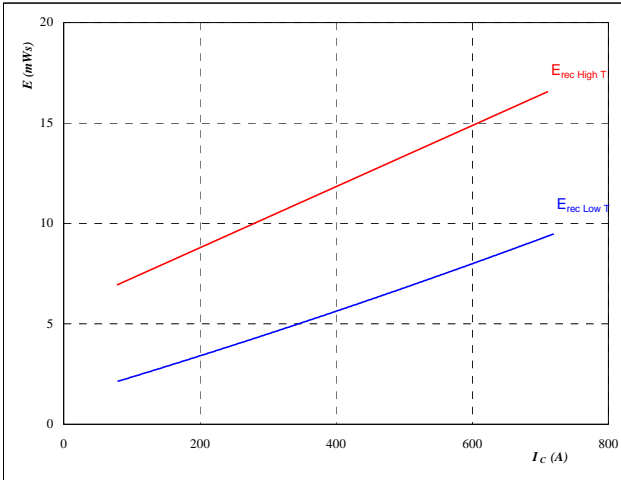


With an inductive load at

 $T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $I_C = 400 \text{ A}$
Figure 7 FWD

Typical reverse recovery energy loss
 as a function of collector current

$$E_{rec} = f(I_C)$$

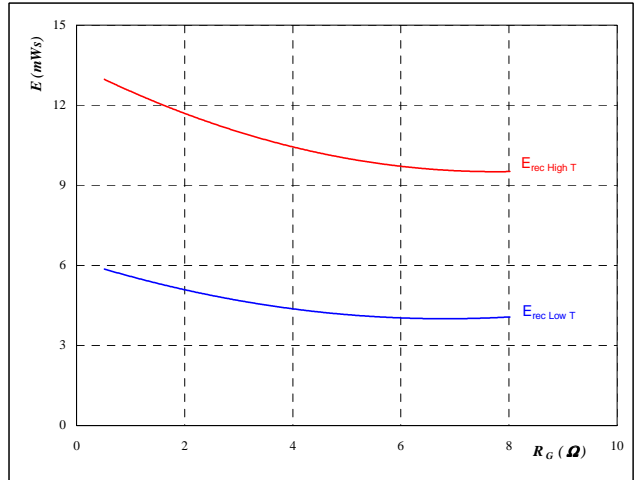


With an inductive load at

 $T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 1 \text{ } \Omega$
Figure 8 FWD

Typical reverse recovery energy loss
 as a function of gate resistor

$$E_{rec} = f(R_G)$$



With an inductive load at

 $T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $I_C = 400 \text{ A}$

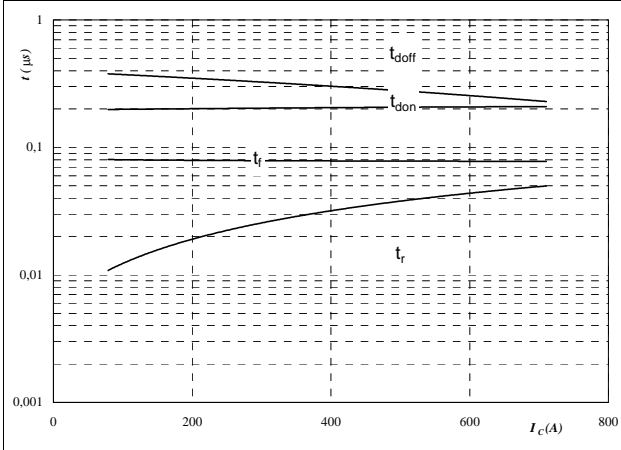
Boost

neutral point IGBT and half bridge FWD

Figure 9 IGBT

Typical switching times as a function of collector current

$$t = f(I_c)$$



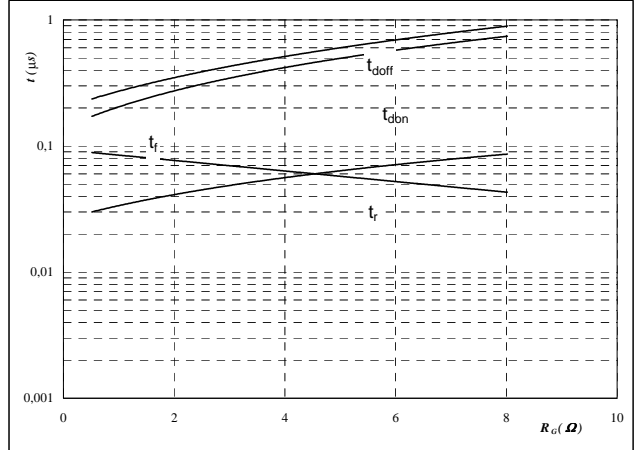
With an inductive load at

$T_j =$	125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω
$R_{goff} =$	1	Ω

Figure 10 IGBT

Typical switching times as a function of gate resistor

$$t = f(R_G)$$



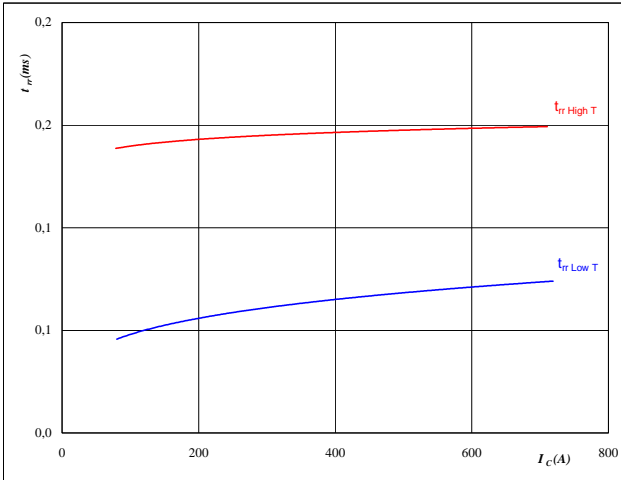
With an inductive load at

$T_j =$	125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$I_c =$	400	A

Figure 11 FWD

Typical reverse recovery time as a function of collector current

$$t_{rr} = f(I_c)$$



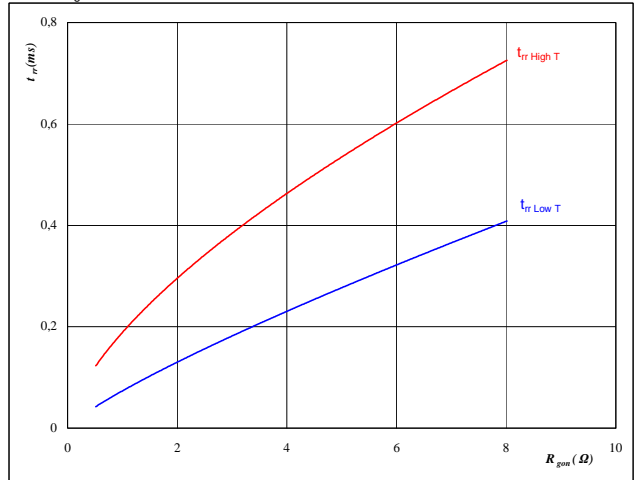
At

$T_j =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω

Figure 12 FWD

Typical reverse recovery time as a function of IGBT turn on gate resistor

$$t_{rr} = f(R_{gon})$$



At

$T_j =$	25/125	°C
$V_R =$	350	V
$I_F =$	400	A
$V_{GE} =$	±15	V

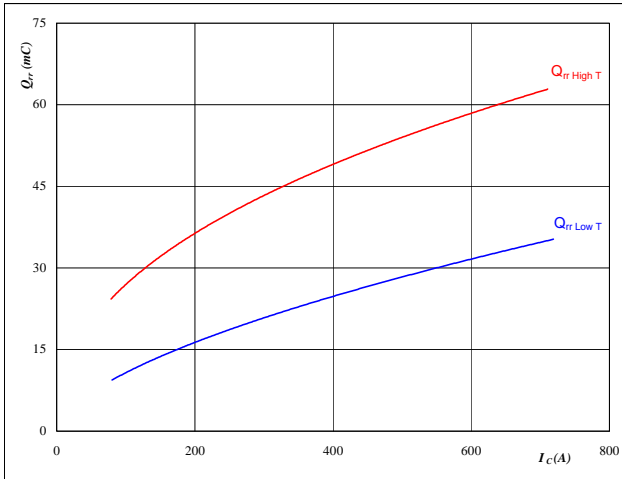
Boost

neutral point IGBT and half bridge FWD

Figure 13 FWD

Typical reverse recovery charge as a function of collector current

$$Q_{rr} = f(I_C)$$

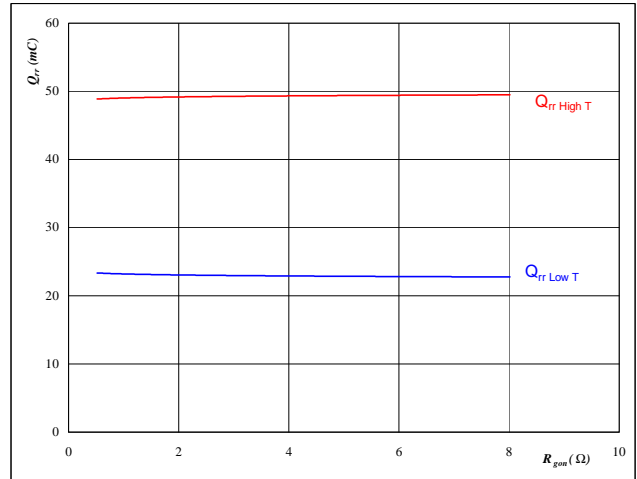

At

$T_j =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω

Figure 14 FWD

Typical reverse recovery charge as a function of IGBT turn on gate resistor

$$Q_{rr} = f(R_{gon})$$

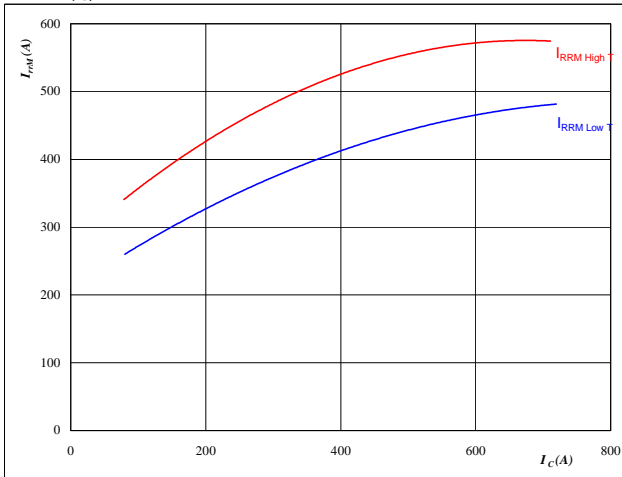

At

$T_j =$	25/125	°C
$V_R =$	350	V
$I_F =$	400	A
$V_{GE} =$	±15	V

Figure 15 FWD

Typical reverse recovery current as a function of collector current

$$I_{RRM} = f(I_C)$$

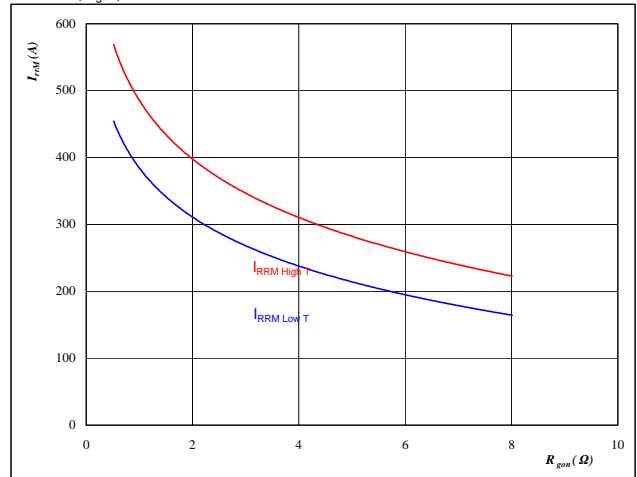

At

$T_j =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	1	Ω

Figure 16 FWD

Typical reverse recovery current as a function of IGBT turn on gate resistor

$$I_{RRM} = f(R_{gon})$$


At

$T_j =$	25/125	°C
$V_R =$	350	V
$I_F =$	400	A
$V_{GE} =$	±15	V

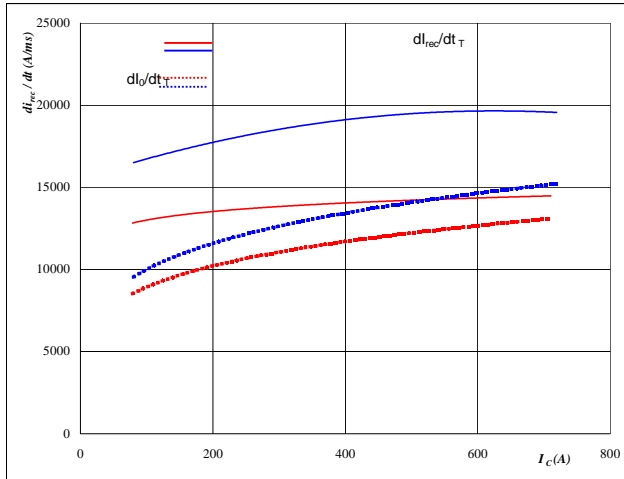
Boost

neutral point IGBT and half bridge FWD

Figure 17 FWD

Typical rate of fall of forward and reverse recovery current as a function of collector current

$$di_o/dt, di_{rec}/dt = f(I_c)$$

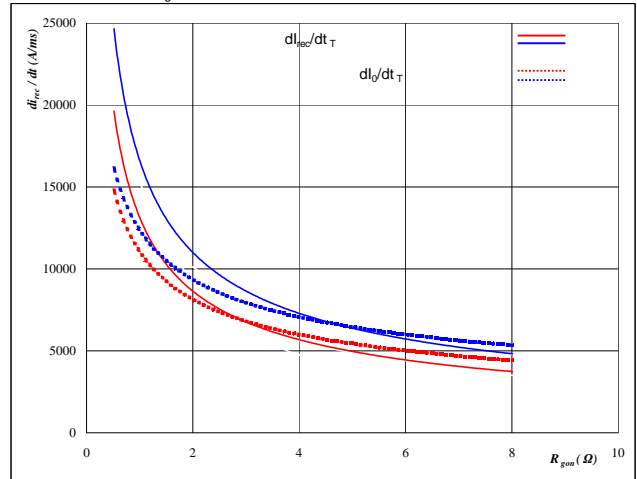


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 1 \text{ } \Omega$

Figure 18 FWD

Typical rate of fall of forward and reverse recovery current as a function of IGBT turn on gate resistor

$$di_o/dt, di_{rec}/dt = f(R_{gon})$$

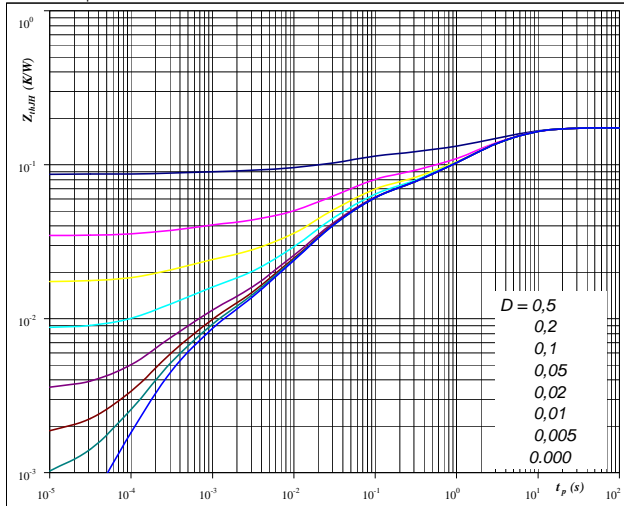


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 400 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$

Figure 19 IGBT

IGBT transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$



At
 $D = t_p / T$
 $R_{thJH} = 0,17 \text{ K/W}$

IGBT thermal model values

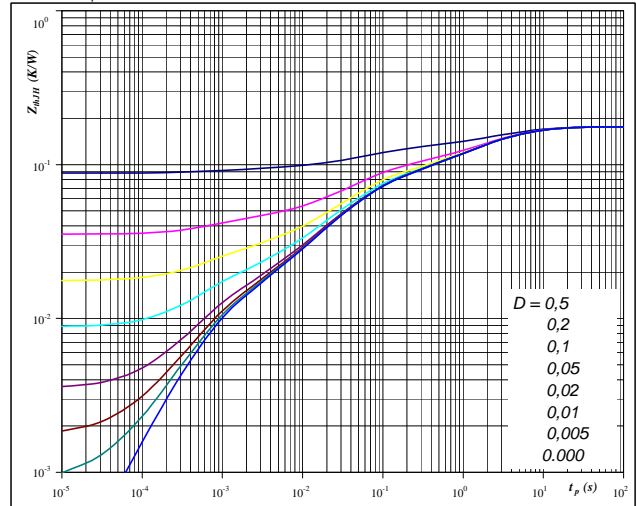
IGBT thermal model values

R (C/W)	Tau (s)
0,03	8,9E+00
0,07	2,2E+00
0,02	3,7E-01
0,04	4,3E-02
0,01	1,1E-02
0,00	1,9E-03

Figure 20 FWD

FWD transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$



At
 $D = t_p / T$
 $R_{thJH} = 0,18 \text{ K/W}$

FWD thermal model values

FWD thermal model values

R (C/W)	Tau (s)
0,02	9,8E+00
0,05	2,5E+00
0,03	6,5E-01
0,03	8,1E-02
0,03	2,7E-02
0,01	4,1E-03

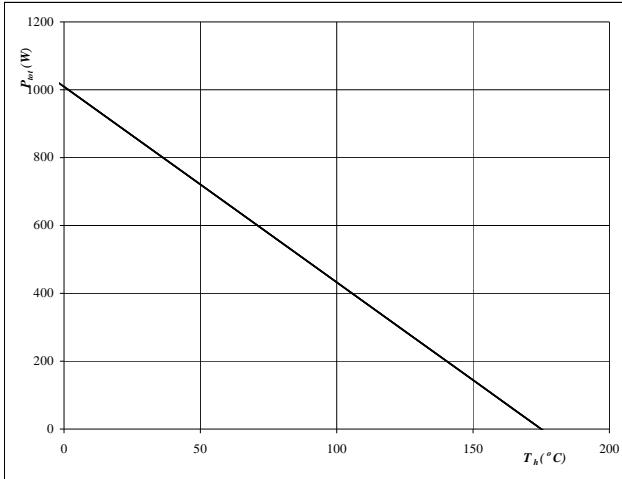
Boost

neutral point IGBT and half bridge FWD

Figure 21 IGBT

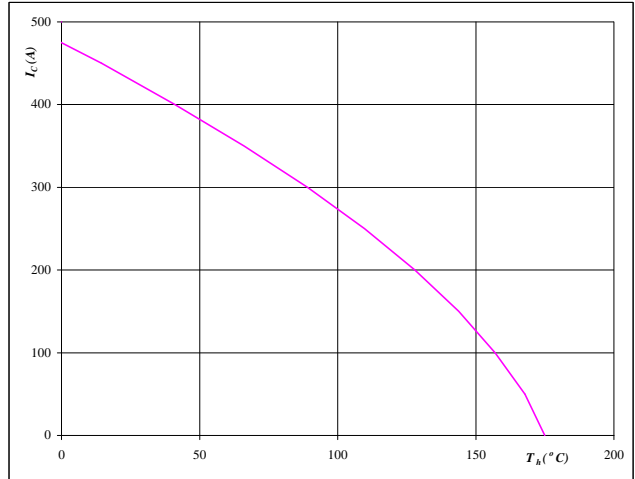
Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$


At
 $T_j = 175 \text{ } ^\circ\text{C}$
Figure 22 IGBT

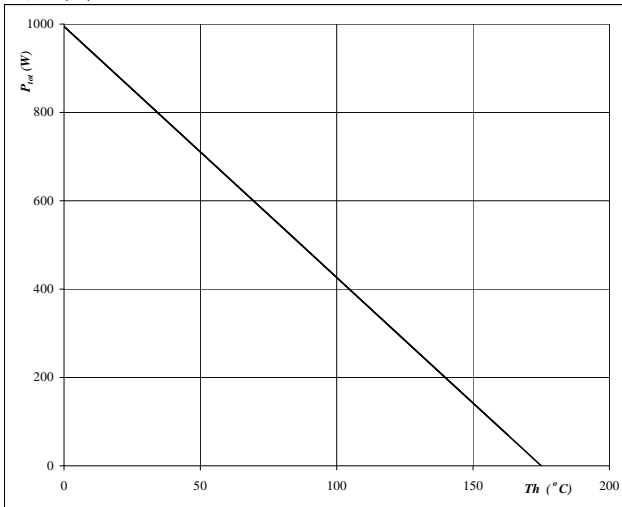
Collector current as a function of heatsink temperature

$$I_C = f(T_h)$$


At
 $T_j = 175 \text{ } ^\circ\text{C}$
 $V_{GE} = 15 \text{ V}$
Figure 23 FWD

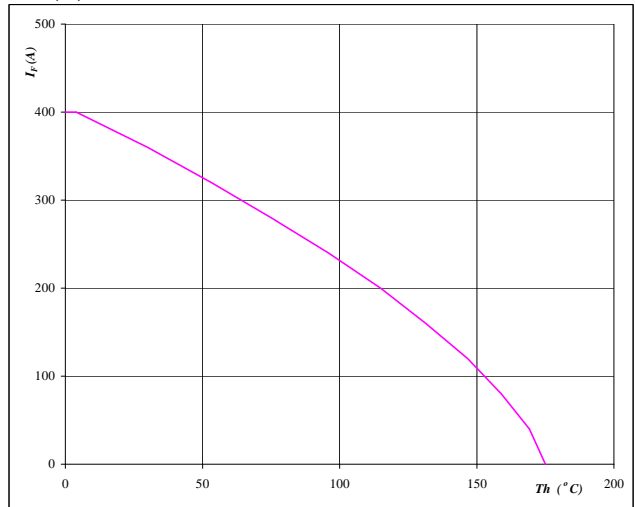
Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$


At
 $T_j = 175 \text{ } ^\circ\text{C}$
Figure 24 FWD

Forward current as a function of heatsink temperature

$$I_F = f(T_h)$$

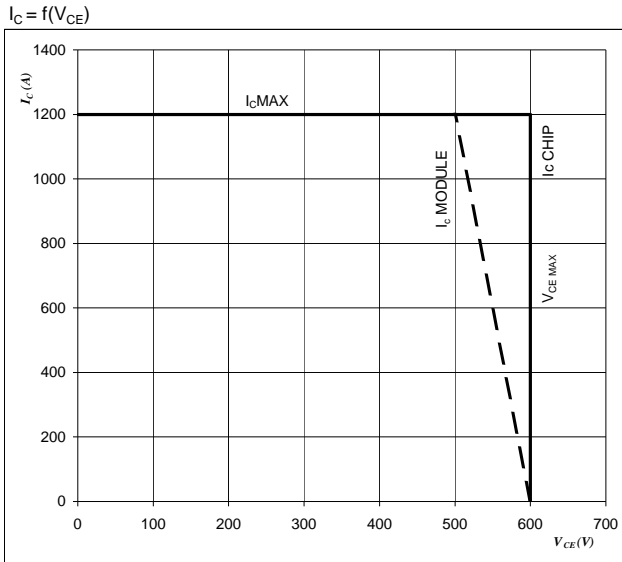

At
 $T_j = 175 \text{ } ^\circ\text{C}$

Boost

neutral point IGBT

Figure 25 IGBT

Reverse bias safe operating area



At

$$T_J = T_{j\text{max}} - 25 \text{ } ^\circ\text{C}$$

$$U_{oc\text{minus}} = U_{cc\text{plus}}$$

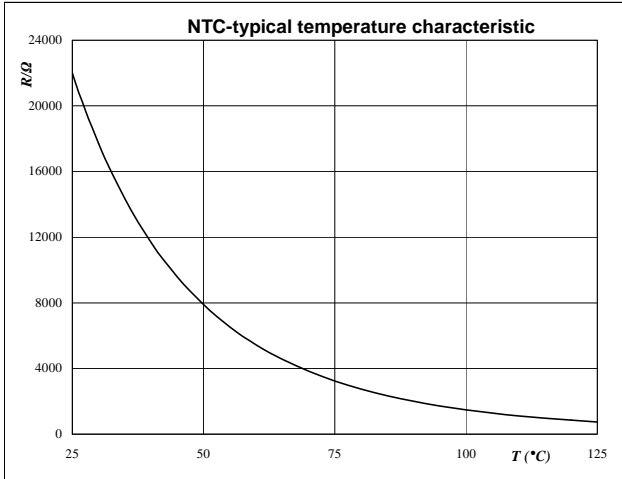
Switching mode : 3 level switching

Thermistor

Figure 1 Thermistor

Typical NTC characteristic
as a function of temperature

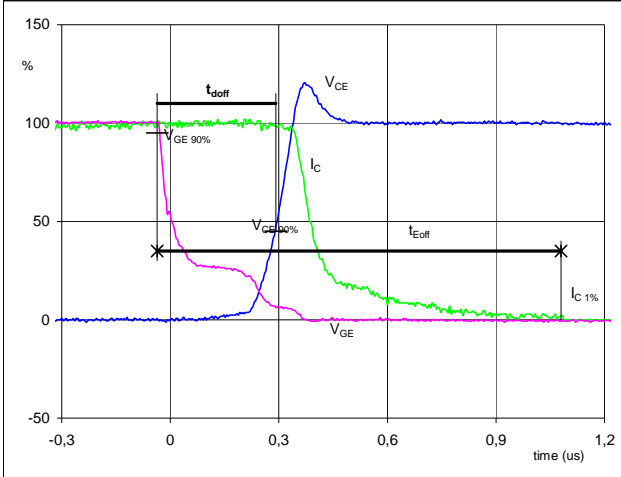
$$R_T = f(T)$$



Switching Definitions half bridge IGBT

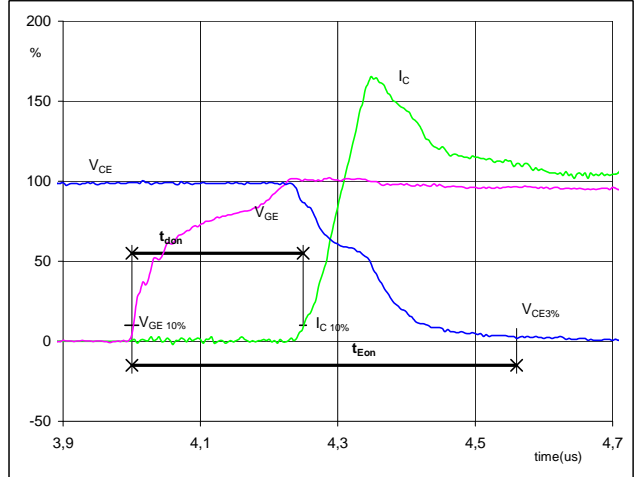
General conditions	
T_j	= 125 °C
R_{gon}	= 1 Ω
R_{goff}	= 1 Ω

Figure 1 half bridge IGBT

Turn-off Switching Waveforms & definition of t_{doff} , t_{Eoff}
 (t_{Eoff} = integrating time for E_{off})


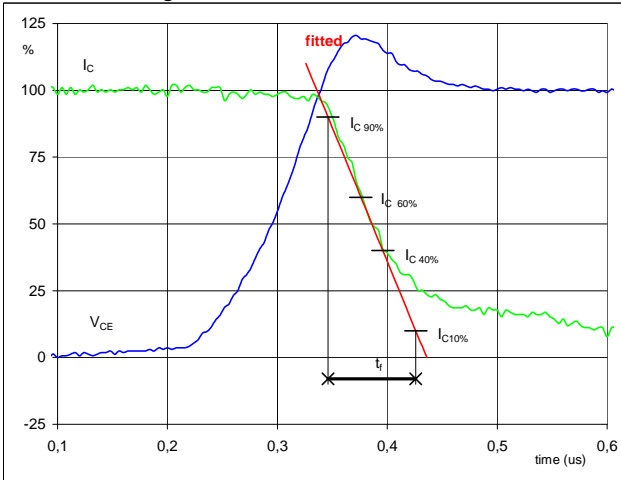
$V_{GE}(0\%) =$	-15	V
$V_{GE}(100\%) =$	15	V
$V_C(100\%) =$	350	V
$I_C(100\%) =$	400	A
$t_{doff} =$	0,35	μs
$t_{Eoff} =$	1,12	μs

Figure 2 half bridge IGBT

Turn-on Switching Waveforms & definition of t_{don} , t_{Eon}
 (t_{Eon} = integrating time for E_{on})


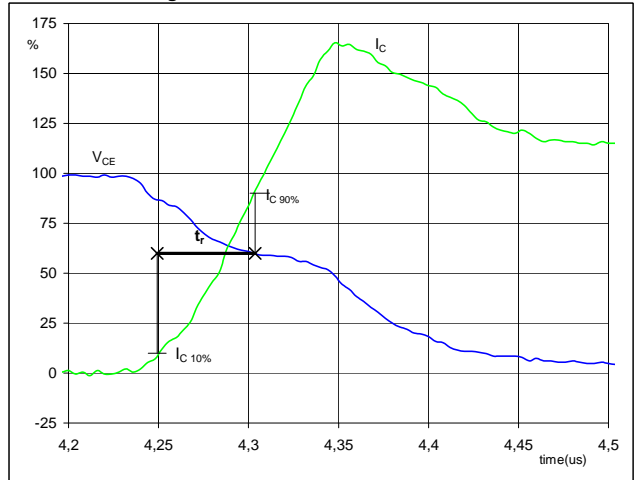
$V_{GE}(0\%) =$	-15	V
$V_{GE}(100\%) =$	15	V
$V_C(100\%) =$	350	V
$I_C(100\%) =$	400	A
$t_{don} =$	0,25	μs
$t_{Eon} =$	0,56	μs

Figure 3 half bridge IGBT

Turn-off Switching Waveforms & definition of t_f


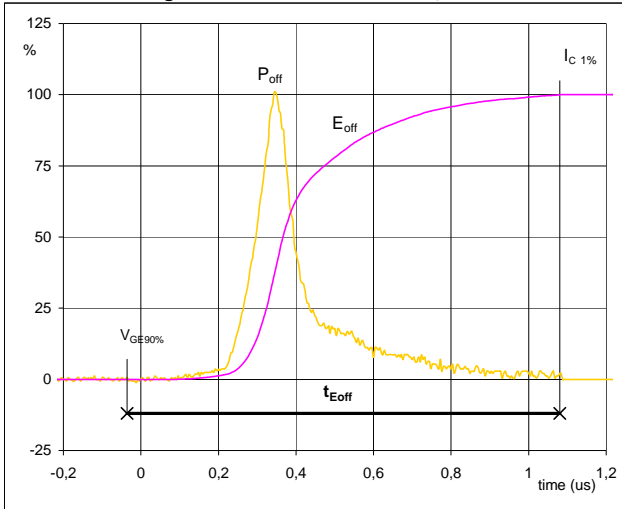
$V_C(100\%) =$	350	V
$I_C(100\%) =$	400	A
$t_f =$	0,09	μs

Figure 4 half bridge IGBT

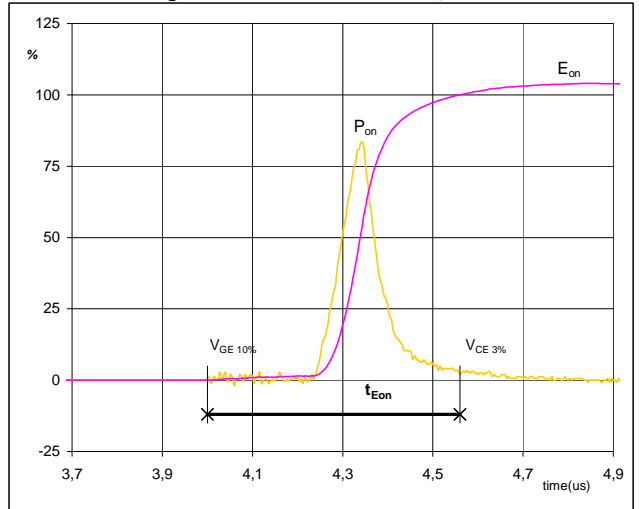
Turn-on Switching Waveforms & definition of t_r


$V_C(100\%) =$	350	V
$I_C(100\%) =$	400	A
$t_r =$	0,06	μs

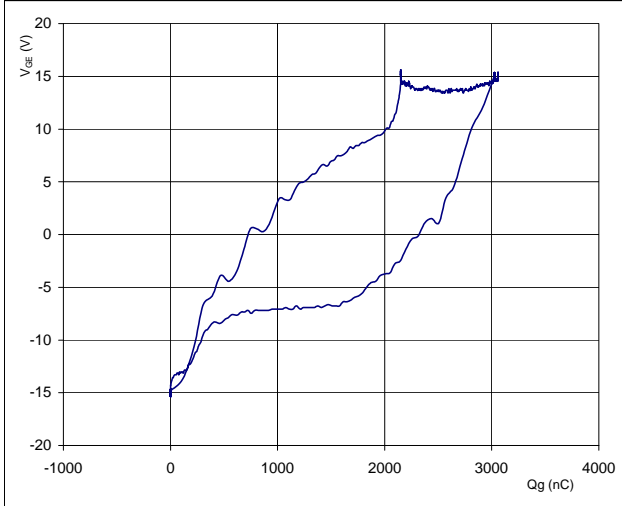
Switching Definitions half bridge IGBT

Figure 5 half bridge IGBT
Turn-off Switching Waveforms & definition of t_{Eoff}


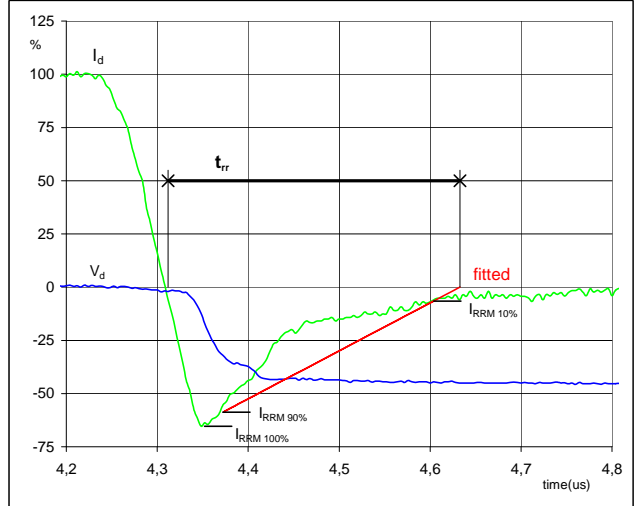
$P_{off} (100\%) = 140,00$ kW
 $E_{off} (100\%) = 22,08$ mJ
 $t_{Eoff} = 1,12$ μ s

Figure 6 half bridge IGBT
Turn-on Switching Waveforms & definition of t_{Eon}


$P_{on} (100\%) = 140,00$ kW
 $E_{on} (100\%) = 12,30$ mJ
 $t_{Eon} = 0,56$ μ s

Figure 7 half bridge IGBT
Gate voltage vs Gate charge (measured)


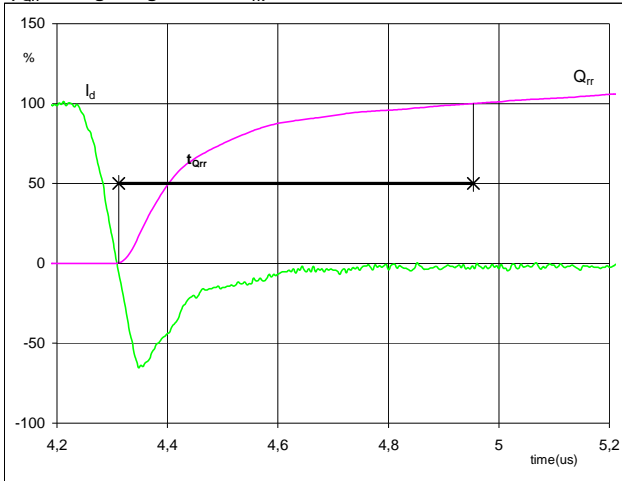
$V_{GEoff} = -15$ V
 $V_{GEon} = 15$ V
 $V_C (100\%) = 350$ V
 $I_C (100\%) = 400$ A
 $Q_g = 3059$ nC

Figure 8 neutral point FWD
Turn-off Switching Waveforms & definition of t_{rr}


$V_d (100\%) = 350$ V
 $I_d (100\%) = 400$ A
 $I_{RRM} (100\%) = -262$ A
 $t_{rr} = 0,30$ μ s

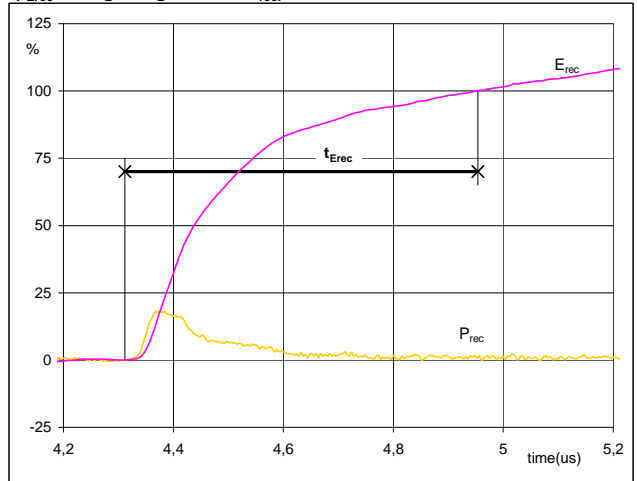
Switching Definitions half bridge IGBT

Figure 9 neutral point FWD

Turn-on Switching Waveforms & definition of t_{Qrr}
 (t_{Qrr} = integrating time for Q_{rr})


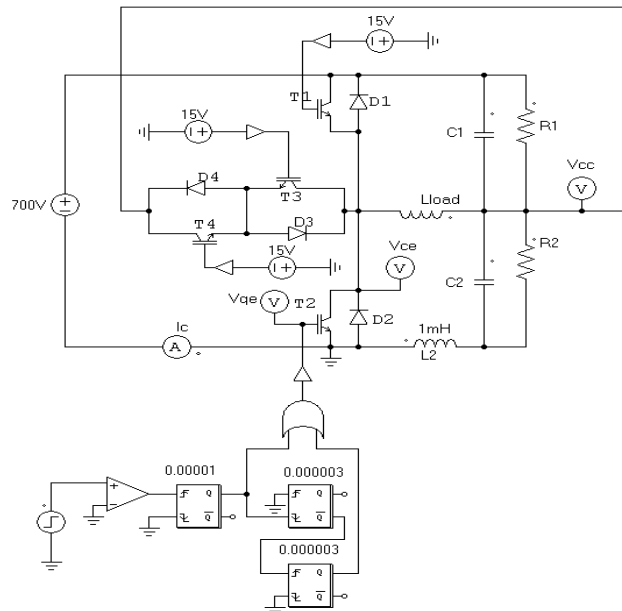
I_d (100%) =	400	A
Q_{rr} (100%) =	33,04	μC
t_{Qrr} =	0,64	μs

Figure 10 neutral point FWD

Turn-on Switching Waveforms & definition of t_{Erec}
 (t_{Erec} = integrating time for E_{rec})


P_{rec} (100%) =	140,00	kW
E_{rec} (100%) =	7,44	mJ
t_{Erec} =	0,64	μs

half bridge IGBT switching measurement circuit

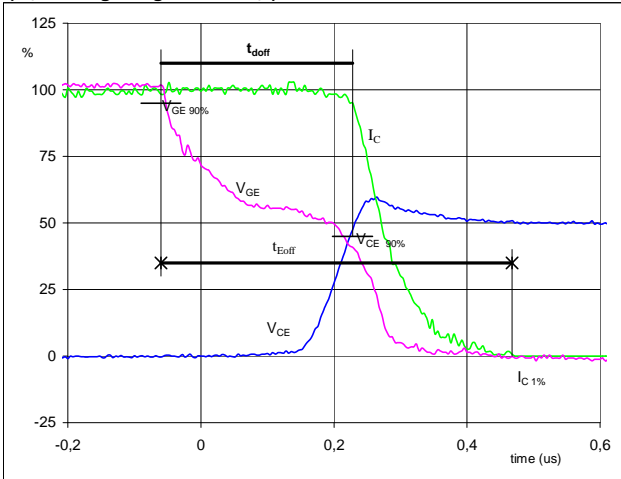
Figure 11


Switching Definitions neutral point IGBT

General conditions	
T_j	= 125 °C
R_{gon}	= 1 Ω
R_{goff}	= 1 Ω

Figure 1 neutral point IGBT

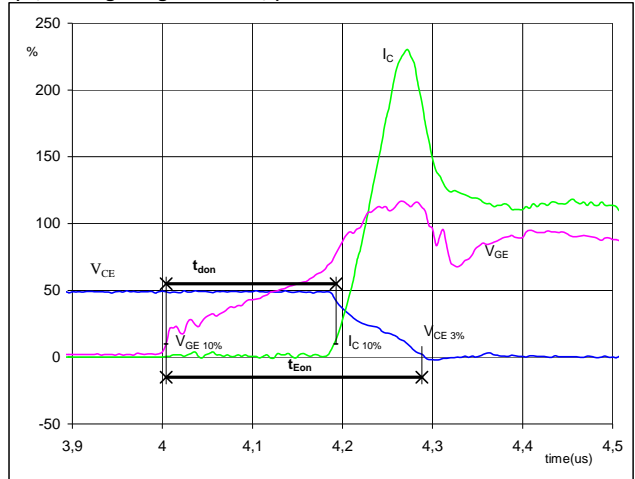
Turn-off Switching Waveforms & definition of t_{doff} , t_{Eoff}
(t_{Eoff} = integrating time for E_{off})



$V_{GE}(0\%) =$	-15	V
$V_{GE}(100\%) =$	15	V
$V_C(100\%) =$	700	V
$I_C(100\%) =$	400	A
$t_{doff} =$	0,23	μs
$t_{Eoff} =$	0,58	μs

Figure 2 neutral point IGBT

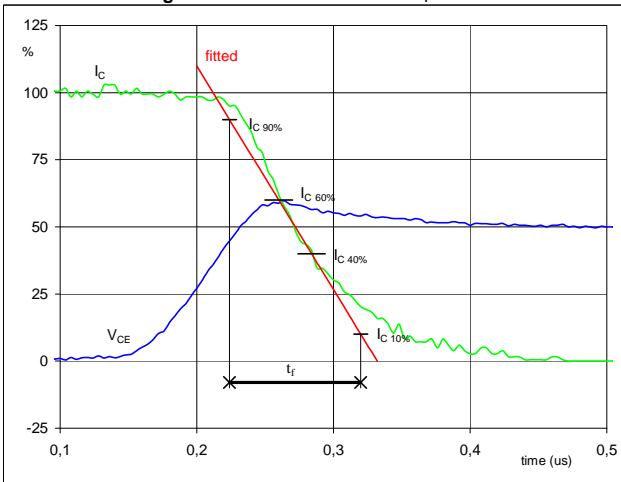
Turn-on Switching Waveforms & definition of t_{don} , t_{Eon}
(t_{Eon} = integrating time for E_{on})



$V_{GE}(0\%) =$	-15	V
$V_{GE}(100\%) =$	15	V
$V_C(100\%) =$	700	V
$I_C(100\%) =$	400	A
$t_{don} =$	0,20	μs
$t_{Eon} =$	0,38	μs

Figure 3 neutral point IGBT

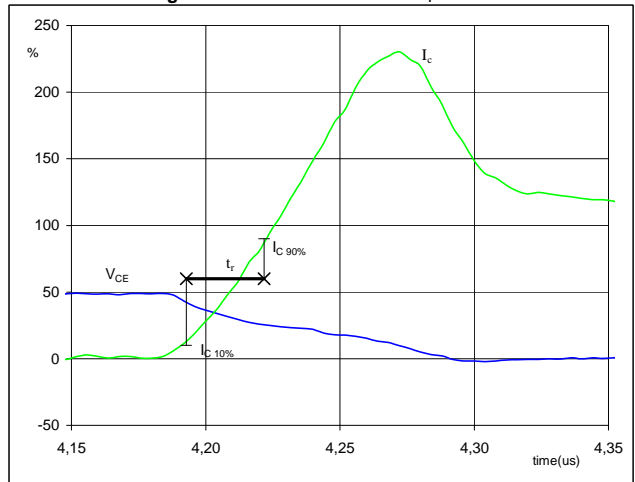
Turn-off Switching Waveforms & definition of t_f



$V_C(100\%) =$	700	V
$I_C(100\%) =$	400	A
$t_f =$	0,088	μs

Figure 4 neutral point IGBT

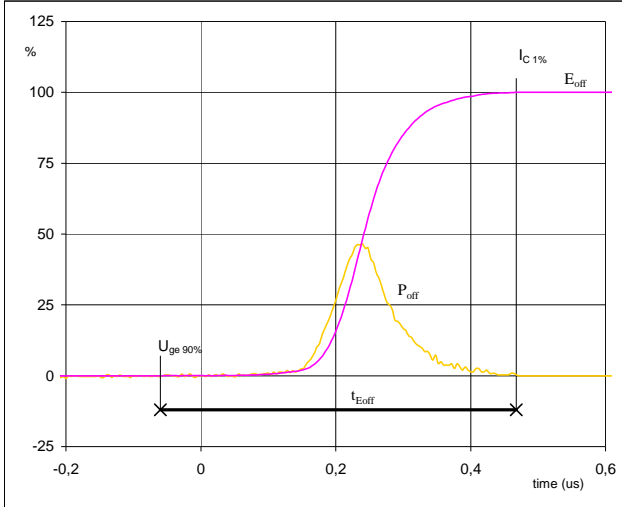
Turn-on Switching Waveforms & definition of t_r



$V_C(100\%) =$	700	V
$I_C(100\%) =$	400	A
$t_r =$	0,032	μs

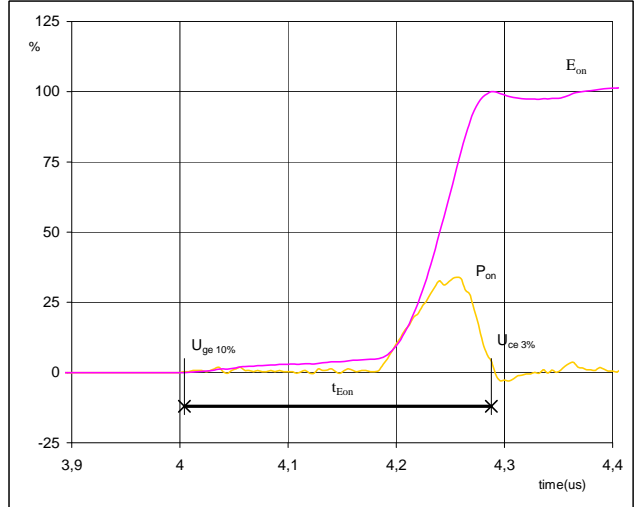
Switching Definitions neutral point IGBT

Figure 5 neutral point IGBT

Turn-off Switching Waveforms & definition of t_{Eoff}


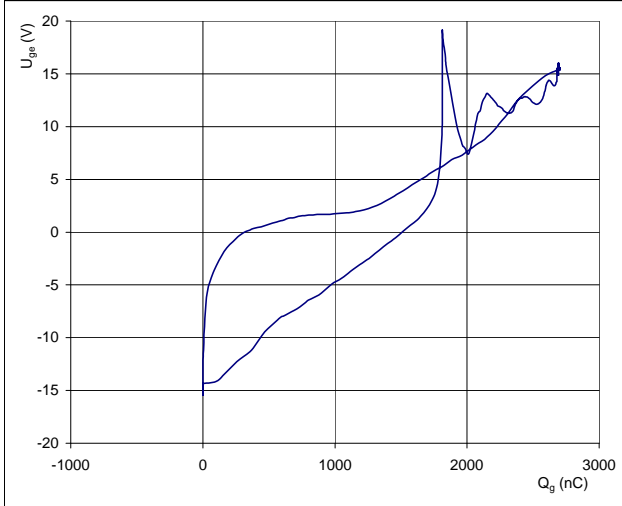
$P_{off} (100\%) = 280,22 \text{ kW}$
 $E_{off} (100\%) = 14,07 \text{ mJ}$
 $t_{Eoff} = 0,58 \text{ }\mu\text{s}$

Figure 6 neutral point IGBT

Turn-on Switching Waveforms & definition of t_{Eon}


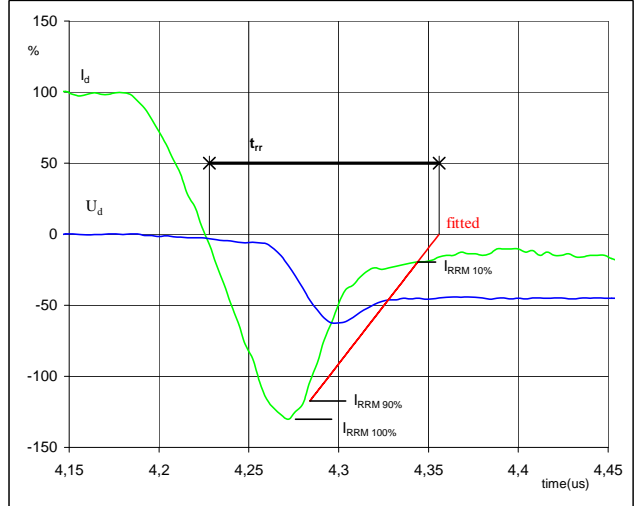
$P_{on} (100\%) = 280,2184 \text{ kW}$
 $E_{on} (100\%) = 13,39 \text{ mJ}$
 $t_{Eon} = 0,38 \text{ }\mu\text{s}$

Figure 7 neutral point IGBT

Gate voltage vs Gate charge (measured)


$V_{GEoff} = -15 \text{ V}$
 $V_{GEon} = 15 \text{ V}$
 $V_C (100\%) = 700 \text{ V}$
 $I_C (100\%) = 400 \text{ A}$
 $Q_g = 3442 \text{ nC}$

Figure 8 half bridge FWD

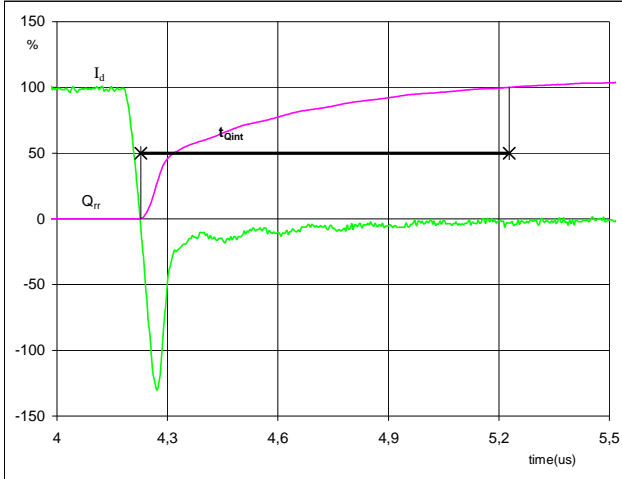
Turn-off Switching Waveforms & definition of t_{rr}


$V_d (100\%) = 700 \text{ V}$
 $I_d (100\%) = 400 \text{ A}$
 $I_{RRM} (100\%) = -521 \text{ A}$
 $t_{rr} = 0,15 \text{ }\mu\text{s}$

Switching Definitions neutral point IGBT

Figure 9 half bridge FWD

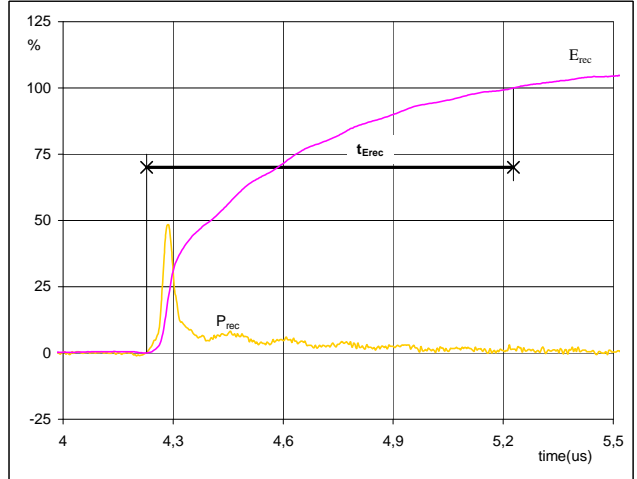
Turn-on Switching Waveforms & definition of t_{Qrr}
(t_{Qrr} = integrating time for Q_{rr})



I_d (100%) = 400 A
 Q_{rr} (100%) = 49,18 μ C
 t_{Qint} = 0,33 μ s

Figure 10 half bridge FWD

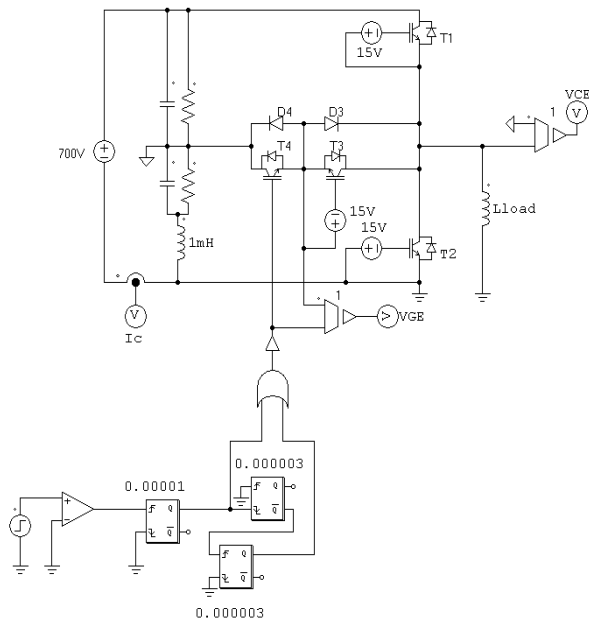
Turn-on Switching Waveforms & definition of t_{Erec}
(t_{Erec} = integrating time for E_{rec})



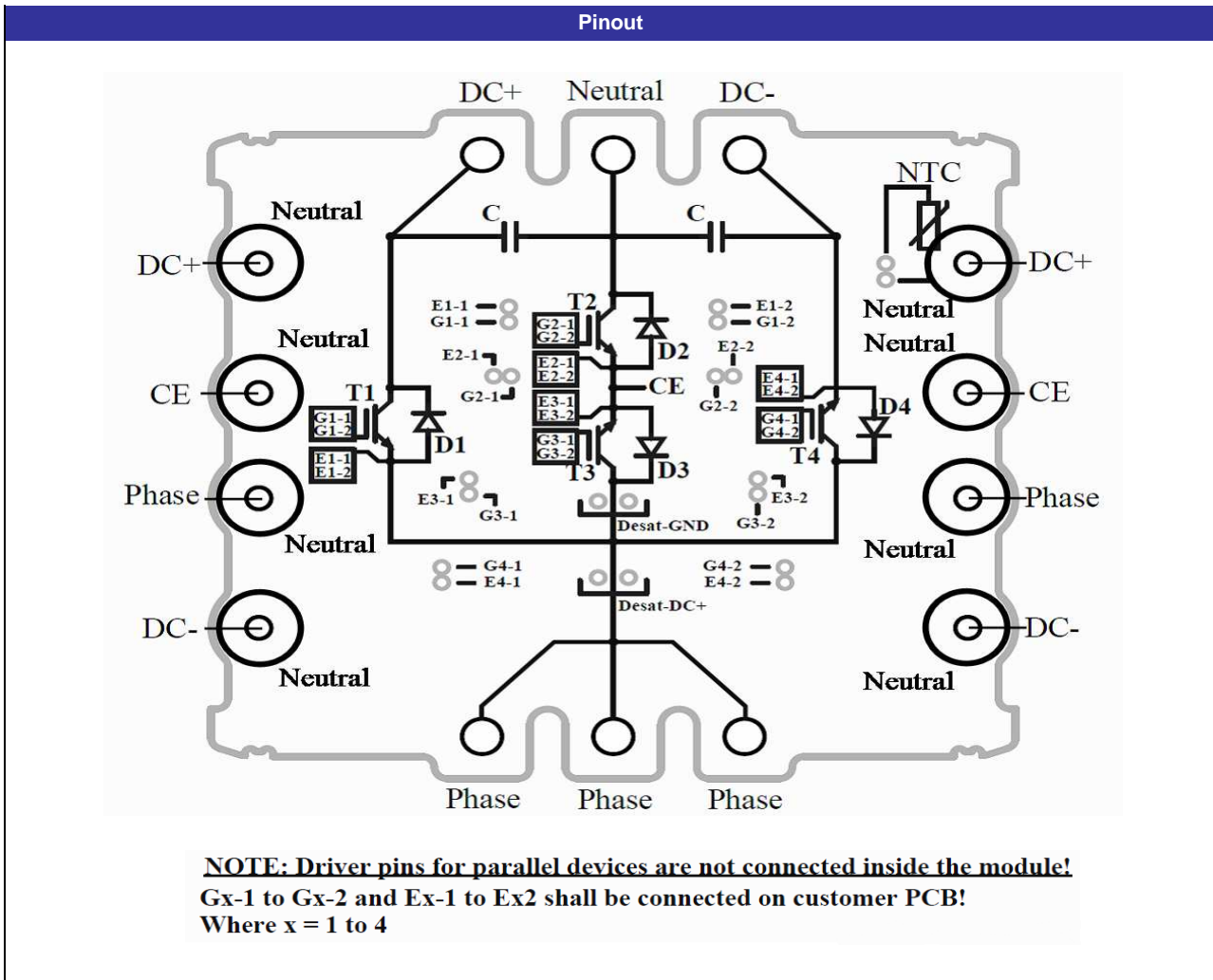
P_{rec} (100%) = 280,22 kW
 E_{rec} (100%) = 12,71 mJ
 t_{Erec} = 0,33 μ s

neutral point IGBT switching measurement circuit

Figure 11



Ordering Code and Marking - Outline - Pinout



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2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.