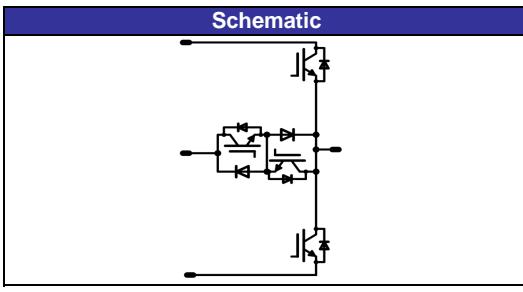


VINcoMNPC X4		1200 V/600 A
Features		
<ul style="list-style-type: none"> • Mixed voltage NPC • Low inductive • High power screw interface 		
Target Applications		
<ul style="list-style-type: none"> • Solar inverter • UPS • High speed motor drive 		
Types		
<ul style="list-style-type: none"> • 70-W212NMA600NB02-M200P62 		
VINco X4 housing		
		
Schematic		
		

Maximum Ratings

T_j=25°C, unless otherwise specified

Parameter	Symbol	Condition	Value	Unit
half bridge IGBT (T1, T4) (T1 , T4)				
Collector-emitter break down voltage	V _{CE}		1200	V
DC collector current	I _C	T _j =T _j max T _h =80°C	517	A
Pulsed collector current	I _{Cpulse}	t _p limited by T _j max	1200	A
Power dissipation per IGBT	P _{tot}	T _j =T _j max T _h =80°C	1051	W
Gate-emitter peak voltage	V _{GE}		±20	V
Short circuit ratings	t _{SC} V _{CC}	T _j ≤150°C V _{GE} =15V	10 850	μs V
Maximum Junction Temperature	T _j max		175	°C

neutral point FWD (D2, D3) (D2 , D3)

Peak Repetitive Reverse Voltage	V _{RRM}		650	V
DC forward current	I _F	T _j =T _j max T _h =80°C	254	A
Repetitive peak forward current	I _{FRM}	t _p = 1 ms T _{vj} < 150°C	800	A
Power dissipation per FWD	P _{tot}	T _j =T _j max T _h =80°C	354	W
Maximum Junction Temperature	T _j max		175	°C

Maximum Ratings

T_j=25°C, unless otherwise specified

Parameter	Symbol	Condition	Value	Unit

neutral point IGBT (T2, T3) (T2 , T3)

Collector-emitter break down voltage	V _{CE}		650	V
DC collector current	I _C	T _j =T _j max T _h =80°C	344	A
Pulsed collector current	I _{Cpuls}	t _p limited by T _j max	1200	A
Power dissipation per IGBT	P _{tot}	T _j =T _j max T _h =80°C	629	W
Gate-emitter peak voltage	V _{GE}		±20	V
Short circuit ratings	t _{SC} V _{CC}	T _j ≤150°C V _{GE} =15V	10 360	μs V
Turn off safe operating area (RBSOA)	I _{cmax}	V _{CE} max = 1200V T _{vj} max= 150°C	800	A
Maximum Junction Temperature	T _j max		175	°C

half bridge FWD (D1, D4) (D1 , D4)

Peak Repetitive Reverse Voltage	V _{RRM}		1200	V
DC forward current	I _F	T _j =T _j max T _h =80°C	272	A
Surge forward current	I _{FSM}		1100	A
I ² t-value	I ² t	t _p =10ms , sin 180° T _j =150°C	3026	A ² s
Repetitive peak forward current	I _{FRM}	t _p limited by T _j max	1200	A
Power dissipation per FWD	P _{tot}	T _j =T _j max T _h =80°C	596	W
Maximum Junction Temperature	T _j max		175	°C

Maximum Ratings

T_j=25°C, unless otherwise specified

Parameter	Symbol	Condition	Value	Unit

General Module Properties

Material of module baseplate			Cu	
Material of internal insulation			Al ₂ O ₃	

Thermal Properties

Storage temperature	T _{stg}		-40...+125	°C
Operation temperature under switching condition	T _{op}		-40...+(T _{jmax} - 25)	°C

Insulation Properties

Insulation voltage	V _{is}	t=2s	DC voltage	4000	V
Creepage distance				min 12,7	mm
Clearance				min 12,7	mm
Comparative tracking index	CTI			>200	

Characteristic Values

Parameter	Symbol		Conditions				Value			Unit
			V_{GE} [V] or V_{GS} [V]	V_r [V] or V_{CE} [V] or V_{DS} [V]	I_c [A] or I_r [A] or I_b [A]	T_j	Min	Typ	Max	
half bridge IGBT (T1, T4) (T1 , T4)										
Gate emitter threshold voltage	$V_{GE(th)}$	$V_{CE}=V_{GE}$			0,03	$T_j=25^\circ C$ $T_j=125^\circ C$	5,4	6	6,6	V
Collector-emitter saturation voltage	$V_{CE(sat)}$		15		600	$T_j=25^\circ C$ $T_j=125^\circ C$		1,86 2,11		V
Collector-emitter cut-off current incl. FWD	I_{CES}		0	1200		$T_j=25^\circ C$ $T_j=125^\circ C$			0,1	mA
Gate-emitter leakage current	I_{GES}		20	0		$T_j=25^\circ C$ $T_j=125^\circ C$			1500	nA
Integrated Gate resistor	R_{gint}							3,25		Ω
Turn-on delay time	$t_{d(on)}$					$T_j=25^\circ C$ $T_j=125^\circ C$	323 340			
Rise time	t_r					$T_j=25^\circ C$ $T_j=125^\circ C$	73 91			
Turn-off delay time	$t_{d(off)}$	$R_{goff}=0,5 \Omega$ $R_{gon}=0,5 \Omega$	± 15	350	600	$T_j=25^\circ C$ $T_j=125^\circ C$	234 274			ns
Fall time	t_f					$T_j=25^\circ C$ $T_j=125^\circ C$	48 66			
Turn-on energy loss per pulse	E_{on}					$T_j=25^\circ C$ $T_j=125^\circ C$	23 34			mWs
Turn-off energy loss per pulse	E_{off}					$T_j=25^\circ C$ $T_j=125^\circ C$	18 26			
Input capacitance	C_{ies}					$T_j=25^\circ C$			60000	
Output capacitance	C_{oss}	$f=1MHz$	0	10		$T_j=25^\circ C$			12000	pF
Reverse transfer capacitance	C_{rss}								1000	
Thermal resistance chip to heatsink per chip	R_{thJH}	100um preapplied PCM						0,09		K/W
Thermal resistance chip to heatsink per chip	R_{thJH}	100um grease 1W/mK						0,11		K/W
neutral point FWD (D2, D3) (D2 , D3)										
FWD forward voltage	V_F				400	$T_j=25^\circ C$ $T_j=125^\circ C$		1,66 1,60		V
Peak reverse recovery current	I_{RRM}					$T_j=25^\circ C$ $T_j=125^\circ C$	158 192			A
Reverse recovery time	t_r					$T_j=25^\circ C$ $T_j=125^\circ C$	281 417			ns
Reverse recovered charge	Q_{rr}	$R_{gon}=0,5 \Omega$	± 15	350	600	$T_j=25^\circ C$ $T_j=125^\circ C$	18 35			μC
Peak rate of fall of recovery current	$di(rec)max/dt$					$T_j=25^\circ C$ $T_j=125^\circ C$	2050 827			A/ μs
Reverse recovered energy	E_{rec}					$T_j=25^\circ C$ $T_j=125^\circ C$	3 7			mWs
Thermal resistance chip to heatsink per chip	R_{thJH}	100um preapplied PCM						0,27		K/W
Thermal resistance chip to heatsink per chip	R_{thJH}	100um grease 1W/mK						0,31		K/W
neutral point IGBT (T2, T3) (T2 , T3)										
Gate emitter threshold voltage	$V_{GE(th)}$	$V_{CE}=V_{GE}$			0,0032	$T_j=25^\circ C$ $T_j=125^\circ C$	5,1	5,8	6,4	V
Collector-emitter saturation voltage	$V_{CE(sat)}$		15		400	$T_j=25^\circ C$ $T_j=125^\circ C$		1,60 1,86		V
Collector-emitter cut-off incl FWD	I_{CES}		0	650		$T_j=25^\circ C$ $T_j=125^\circ C$			0,1	mA
Gate-emitter leakage current	I_{GES}		20	0		$T_j=25^\circ C$ $T_j=125^\circ C$			1500	nA
Integrated Gate resistor	R_{gint}							1		Ω
Turn-on delay time	$t_{d(on)}$					$T_j=25^\circ C$ $T_j=125^\circ C$	209 213			
Rise time	t_r					$T_j=25^\circ C$ $T_j=125^\circ C$	44 49			
Turn-off delay time	$t_{d(off)}$	$R_{goff}=1 \Omega$ $R_{gon}=1 \Omega$	± 15	350	600	$T_j=25^\circ C$ $T_j=125^\circ C$	250 265			ns
Fall time	t_f					$T_j=25^\circ C$ $T_j=125^\circ C$	79 106			
Turn-on energy loss per pulse	E_{on}					$T_j=25^\circ C$ $T_j=125^\circ C$	6 9			mWs
Turn-off energy loss per pulse	E_{off}					$T_j=25^\circ C$ $T_j=125^\circ C$	21 28			
Input capacitance	C_{ies}					$T_j=25^\circ C$		24640		
Output capacitance	C_{oss}	$f=1MHz$	0	25		$T_j=25^\circ C$		1536		pF
Reverse transfer capacitance	C_{rss}							732		
Gate charge	Q_{Gate}		15	480	400	$T_j=25^\circ C$		2507		nC
Thermal resistance chip to heatsink per chip	R_{thJH}	100um preapplied PCM						0,15		K/W
Thermal resistance chip to heatsink per chip	R_{thJH}	100um grease 1W/mK						0,17		K/W

Characteristic Values

Parameter	Symbol	Conditions					Value			Unit
			V_{GE} [V] or V_{GS} [V]	V_r [V] or V_{CE} [V] or V_{DS} [V]	I_c [A] or I_r [A] or I_D [A]	T_j	Min	Typ	Max	
half bridge FWD (D1, D4) (D1 , D4)										
FWD forward voltage	V_F				400	$T_j=25^\circ C$ $T_j=125^\circ C$		2,19 2,47		V
Reverse leakage current	I_r			1200		$T_j=25^\circ C$ $T_j=125^\circ C$			48	µA
Peak reverse recovery current	I_{RRM}	$R_{gon}=1\ \Omega$	± 15	350	600	$T_j=25^\circ C$ $T_j=125^\circ C$		448 568		A
Reverse recovery time	t_{rr}					$T_j=25^\circ C$ $T_j=125^\circ C$		70 138		ns
Reverse recovered charge	Q_{rr}					$T_j=25^\circ C$ $T_j=125^\circ C$		19 53		µC
Peak rate of fall of recovery current	$di(rec)_{max}/dt$					$T_j=25^\circ C$ $T_j=125^\circ C$		20142 14965		A/µs
Reverse recovery energy	E_{rec}					$T_j=25^\circ C$ $T_j=125^\circ C$		4 13		mWs
Thermal resistance chip to heatsink per chip	R_{thJH}								0,16	K/W
Thermal resistance chip to case per chip	R_{thJH}								0,18	K/W

Thermistor

Rated resistance	R					$T_j=25^\circ C$		22000		Ω
Deviation of R_{100}	$\Delta R/R$	$R_{100}=1486\ \Omega$				$T_j=100^\circ C$	-12		+14	%
Power dissipation	P					$T_j=25^\circ C$		200		mW
Power dissipation constant						$T_j=25^\circ C$		2		mW/K
B-value	$B_{(25/50)}$	Tol. ±3%				$T_j=25^\circ C$		3950		K
B-value	$B_{(25/100)}$	Tol. ±3%				$T_j=25^\circ C$		3996		K
Vincotech NTC Reference									B	

Module Properties

Module inductance (from chips to PCB)	$L_{sCE\ C-PCB}$							5		nH
Module inductance (from PCB to PCB using Intercon board)	$L_{sCE\ PCB-PCB}$							3		nH
Resistance of Intercon boards (from PCB to PCB using Intercon board)	$R_{CC-1+EE}$	$T_c=25^\circ C$, per switch						1,5		mΩ
Mounting torque	M	Screw M4 - mounting according to valid application note FSWB1-4TY-M-*-HI					2		2,2	Nm
Mounting torque	M	Screw M5 - mounting according to valid application note FSWB1-4TY-M-*-HI					4		6	Nm
Terminal connection torque	M	Screw M6 - mounting according to valid application note FSWB1-4TY-M-*-HI					2,5		5	Nm
Weight	G								710	g

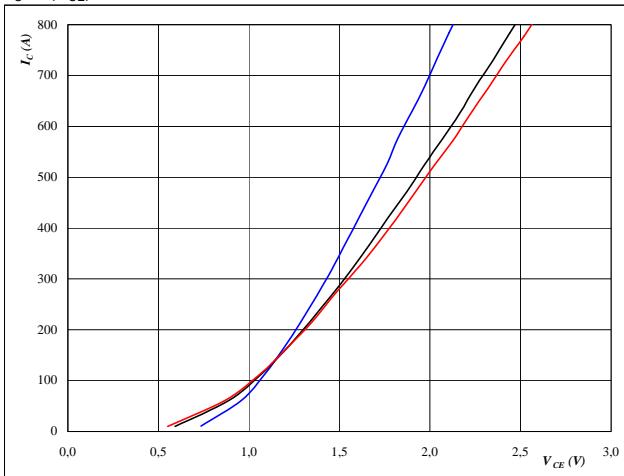
Buck operation

Half Bridge IGBT (T1,T4) and Neutral Point FWD (D2,D3)

Figure 1

Typical output characteristics $V_{GE}=15V$

$$I_C = f(V_{CE})$$



At

$$t_p = 350 \mu s$$

$$T_j = 25/125/150 ^\circ C$$

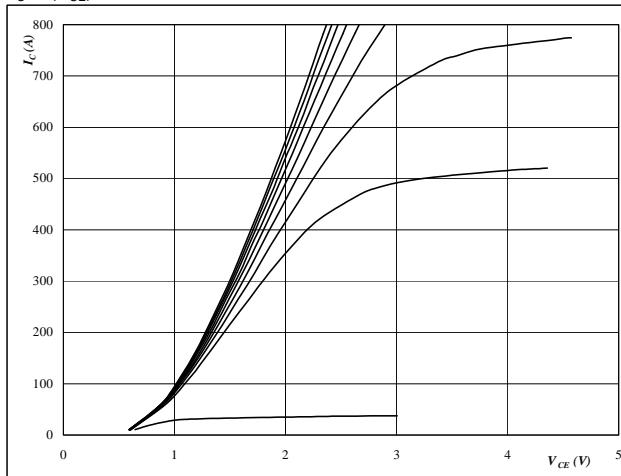
$$V_{GE} = 15 V$$

IGBT

Figure 2

Typical output characteristics

$$I_C = f(V_{CE})$$



At

$$t_p = 350 \mu s$$

$$T_j = 150 ^\circ C$$

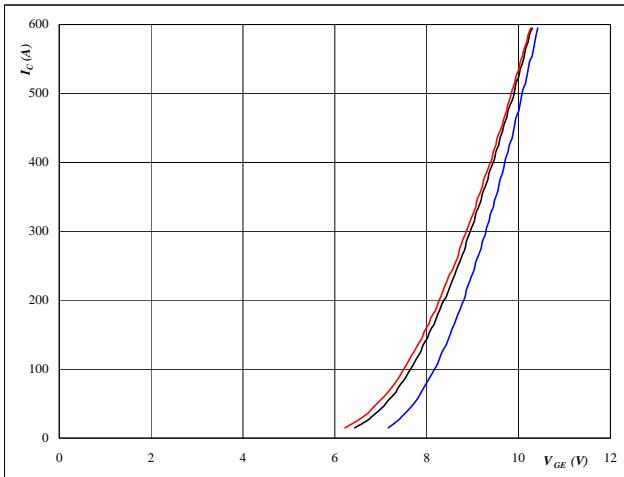
V_{GE} from 7 V to 17 V in steps of 1 V

Figure 3

Typical transfer characteristics

$$I_C = f(V_{GE})$$

IGBT



At

$$t_p = 350 \mu s$$

$$V_{CE} = 350 V$$

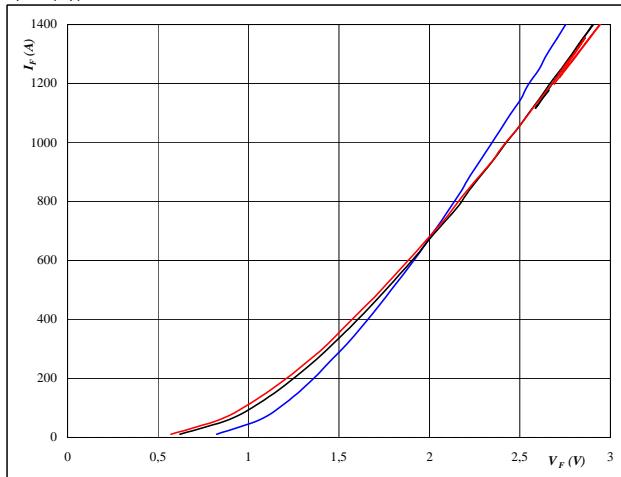
$$T_j = 25/125/150 ^\circ C$$

Figure 4

Typical FWD forward current as a function of forward voltage

$$I_F = f(V_F)$$

FWD



At

$$t_p = 350 \mu s$$

$$T_j = 25/125/150 ^\circ C$$

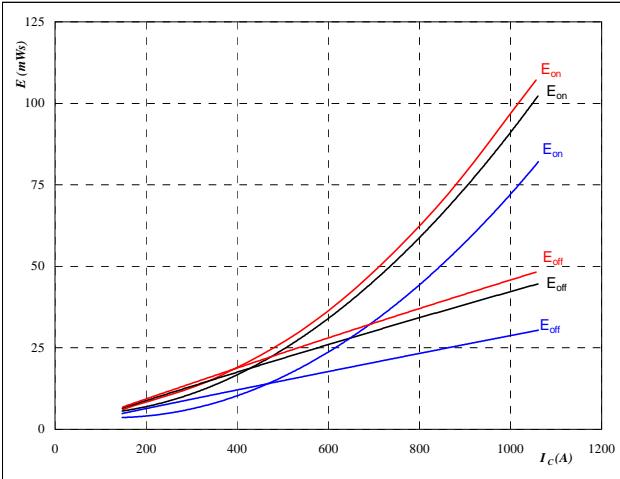
Buck operation

Half Bridge IGBT (T1,T4) and Neutral Point FWD (D2,D3)

Figure 5

Typical switching energy losses
as a function of collector current

$$E = f(I_C)$$



With an inductive load at

$$T_j = 25/125/150 \text{ }^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

$$R_{gon} = 0,5 \text{ } \Omega$$

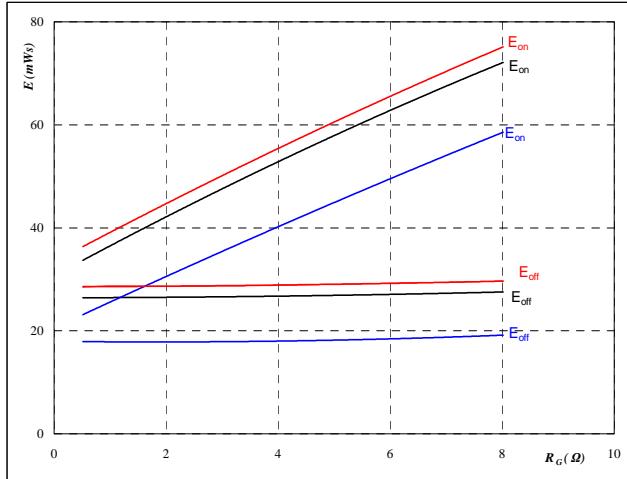
$$R_{goff} = 0,5 \text{ } \Omega$$

IGBT

Figure 6

Typical switching energy losses
as a function of gate resistor

$$E = f(R_G)$$



With an inductive load at

$$T_j = 25/125/150 \text{ }^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

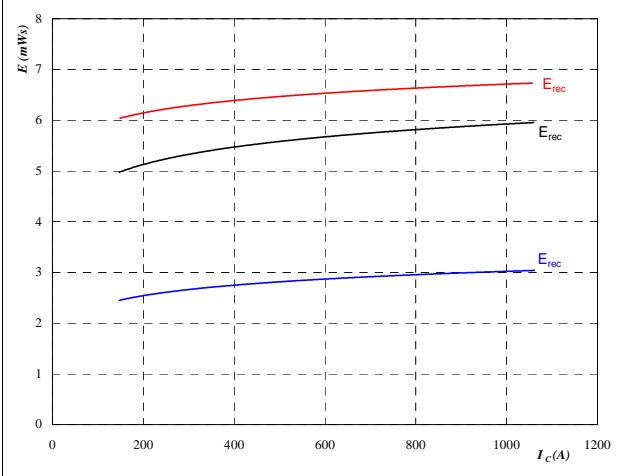
$$V_{GE} = \pm 15 \text{ V}$$

$$I_C = 601 \text{ A}$$

Figure 7

Typical reverse recovery energy loss
as a function of collector current

$$E_{rec} = f(I_C)$$



With an inductive load at

$$T_j = 25/125/150 \text{ }^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

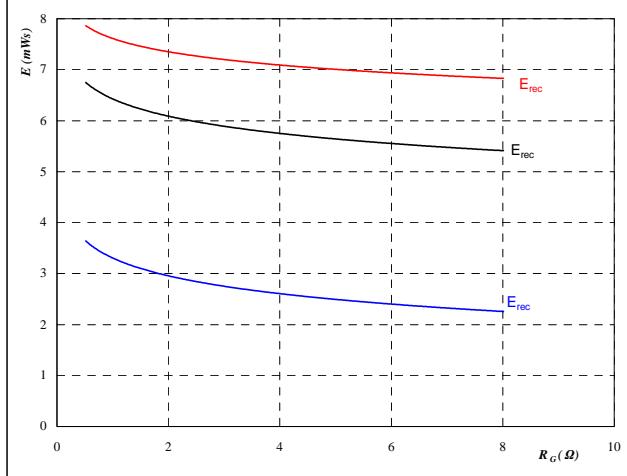
$$R_{gon} = 0,5 \text{ } \Omega$$

FWD

Figure 8

Typical reverse recovery energy loss
as a function of gate resistor

$$E_{rec} = f(R_G)$$



With an inductive load at

$$T_j = 25/125/150 \text{ }^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

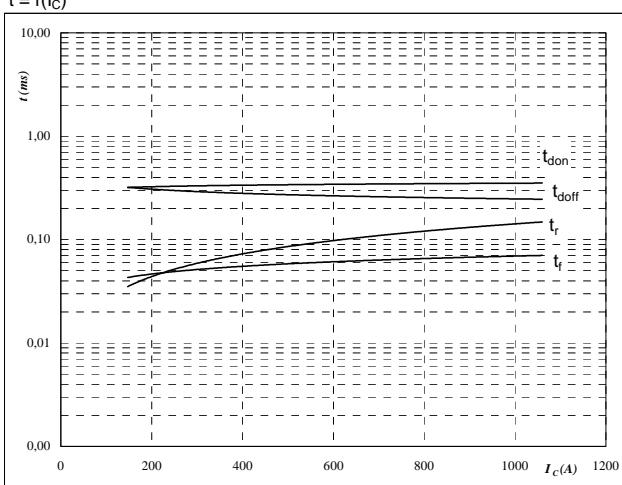
$$I_C = 601 \text{ A}$$

Buck operation

Half Bridge IGBT (T1,T4) and Neutral Point FWD (D2,D3)

Figure 9

Typical switching times as a function of collector current
 $t = f(I_C)$



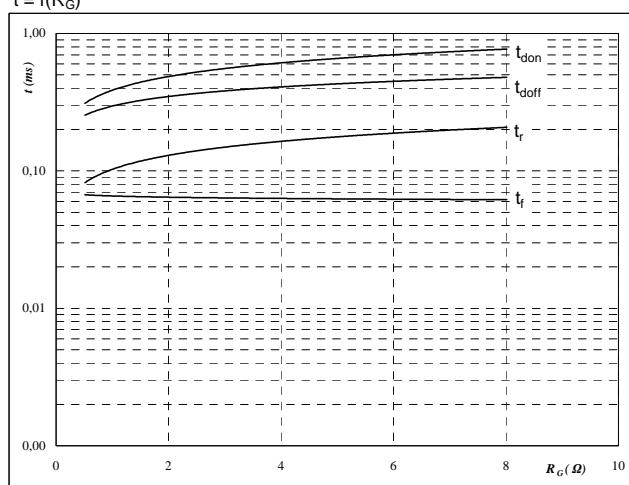
With an inductive load at

$T_j = 125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 0,5 \text{ } \Omega$
 $R_{goff} = 0,5 \text{ } \Omega$

IGBT

Figure 10

Typical switching times as a function of gate resistor
 $t = f(R_G)$

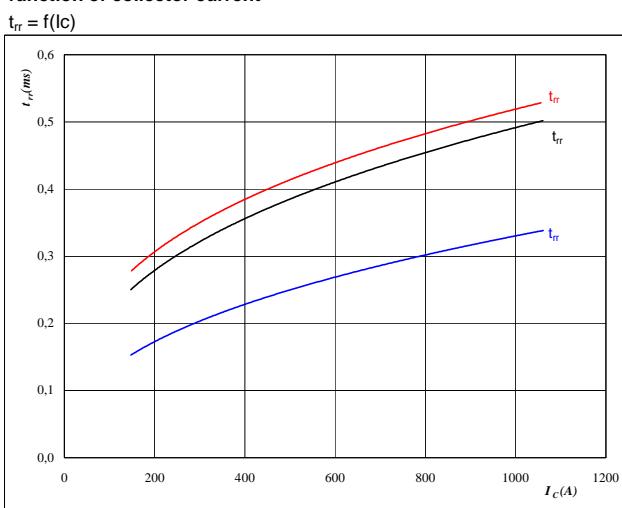


With an inductive load at

$T_j = 125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $I_C = 601 \text{ A}$

Figure 11

Typical reverse recovery time as a function of collector current
 $t_{rr} = f(I_C)$



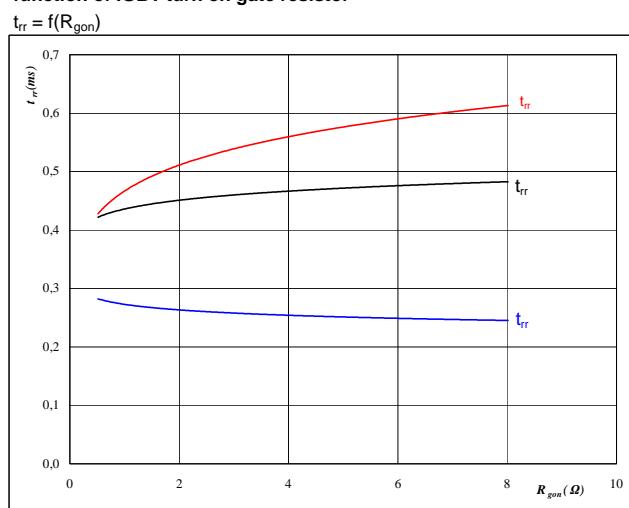
At

$T_j = 25/125/150 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 0,5 \text{ } \Omega$

FWD

Figure 12

Typical reverse recovery time as a function of IGBT turn on gate resistor
 $t_{rr} = f(R_{gon})$



At

$T_j = 25/125/150 \text{ } ^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 601 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$

Buck operation

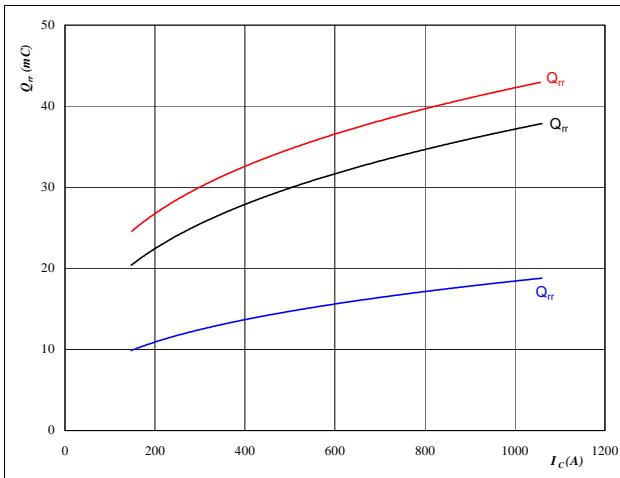
Half Bridge IGBT (T1,T4) and Neutral Point FWD (D2,D3)

Figure 13

Typical reverse recovery charge as a function of collector current

$$Q_{rr} = f(I_C)$$

FWD



At

$$T_j = 25/125/150 \text{ }^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

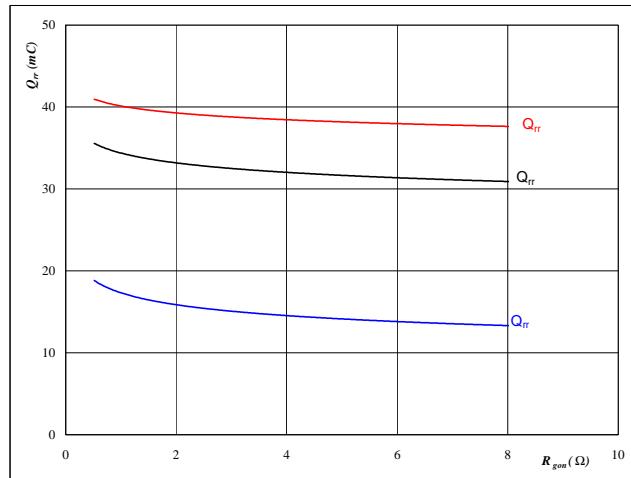
$$R_{gon} = 0,5 \text{ } \Omega$$

Figure 14

Typical reverse recovery charge as a function of IGBT turn on gate resistor

$$Q_{rr} = f(R_{gon})$$

FWD



At

$$T_j = 25/125/150 \text{ }^\circ\text{C}$$

$$V_R = 350 \text{ V}$$

$$I_F = 601 \text{ A}$$

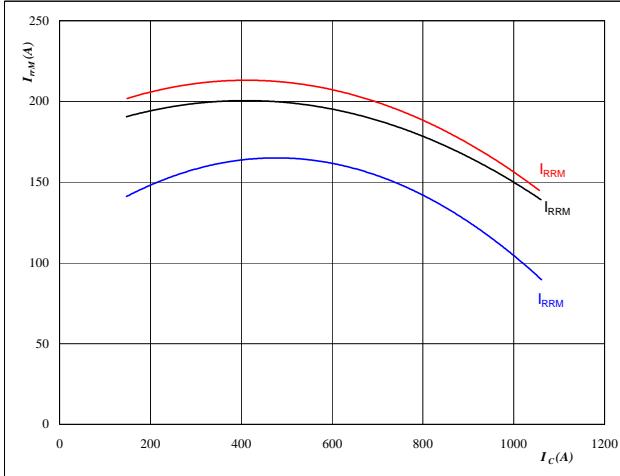
$$V_{GE} = \pm 15 \text{ V}$$

Figure 15

Typical reverse recovery current as a function of collector current

$$I_{RRM} = f(I_C)$$

FWD



At

$$T_j = 25/125/150 \text{ }^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

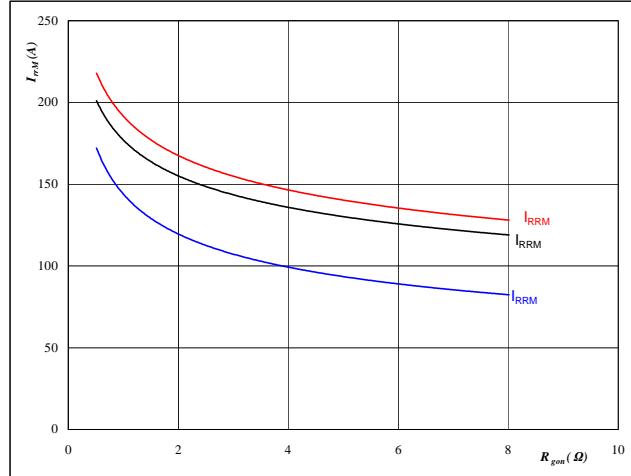
$$R_{gon} = 0,5 \text{ } \Omega$$

Figure 16

Typical reverse recovery current as a function of IGBT turn on gate resistor

$$I_{RRM} = f(R_{gon})$$

FWD



At

$$T_j = 25/125/150 \text{ }^\circ\text{C}$$

$$V_R = 350 \text{ V}$$

$$I_F = 601 \text{ A}$$

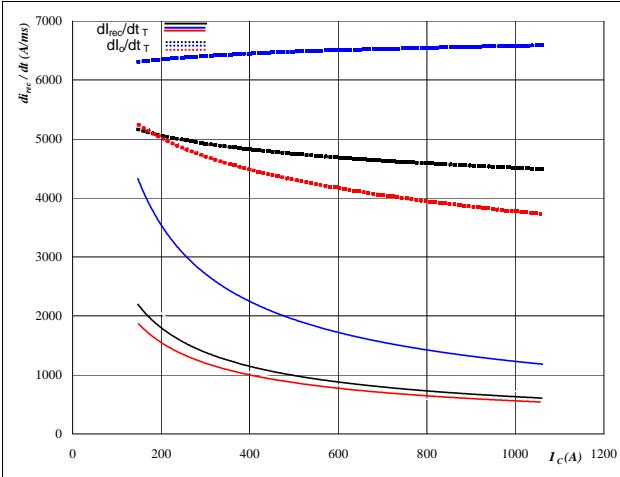
$$V_{GE} = \pm 15 \text{ V}$$

Buck operation

Half Bridge IGBT (T1,T4) and Neutral Point FWD (D2,D3)

Figure 17

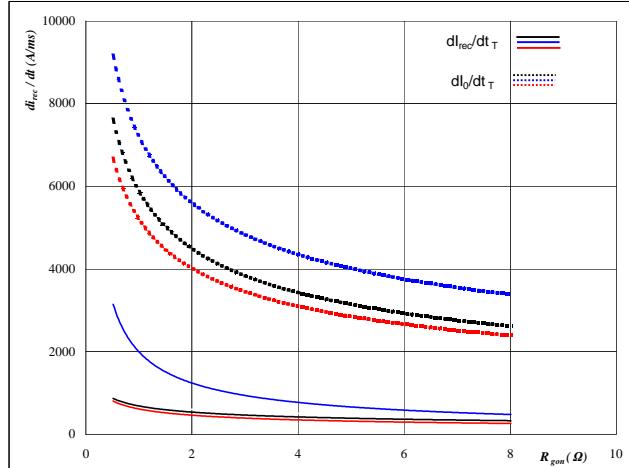
Typical rate of fall of forward
and reverse recovery current as a
function of collector current
 $dI_0/dt, dI_{rec}/dt = f(I_C)$



FWD

Figure 18

Typical rate of fall of forward
and reverse recovery current as a
function of IGBT turn on gate resistor
 $dI_0/dt, dI_{rec}/dt = f(R_{gon})$



FWD

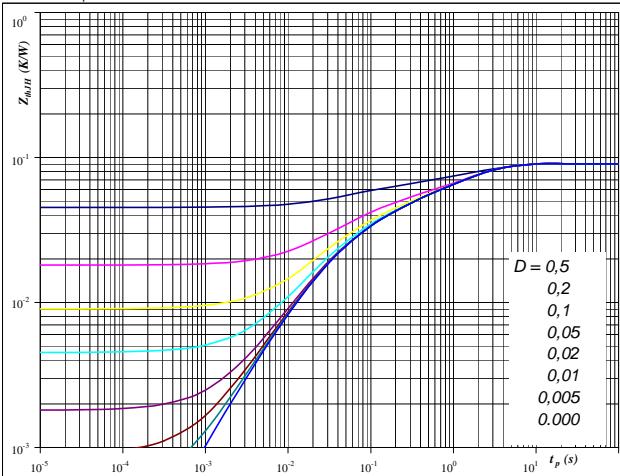
At

$T_j = 25/125/150^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 1,0 \Omega$

Figure 19

IGBT transient thermal impedance
as a function of pulse width

$$Z_{thJH} = f(t_p)$$



IGBT

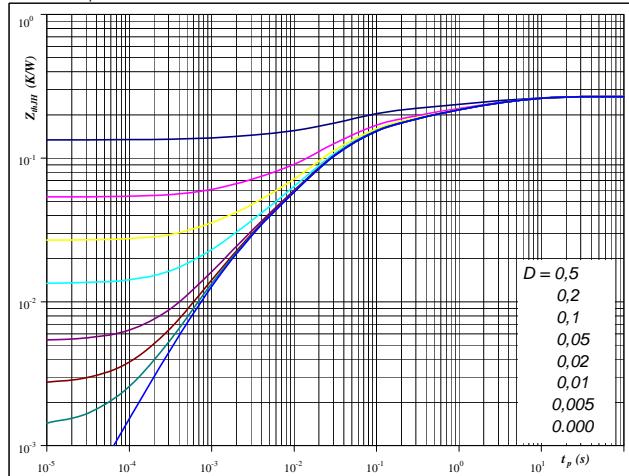
At

$T_j = 25/125/150^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 601 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$

Figure 20

FWD transient thermal impedance
as a function of pulse width

$$Z_{thJH} = f(t_p)$$



FWD

At

$D = t_p / T$
Preapplied PCM Thermal grease
 $R_{thJH} = 0,09 \text{ K/W}$ $R_{thJH} = 0,11 \text{ K/W}$

IGBT thermal model values

100um preapplied PCM 100um grease 1W/mK (P12)

R (C/W)	Tau (s)	R (C/W)	Tau (s)
0,042	1,916	0,051	1,916
0,024	0,234	0,030	0,234
0,023	0,035	0,028	0,035
0,002	0,006	0,002	0,006

At

$D = t_p / T$
Preapplied PCM Thermal grease
 $R_{thJH} = 0,27 \text{ K/W}$ $R_{thJH} = 0,31 \text{ K/W}$

FWD thermal model values

100um preapplied PCM 100um grease 1W/mK (P12)

R (C/W)	Tau (s)	R (C/W)	Tau (s)
0,040	5,632	0,047	5,632
0,044	1,073	0,051	1,073
0,044	0,202	0,051	0,202
0,087	0,041	0,100	0,041
0,038	0,012	0,044	0,012

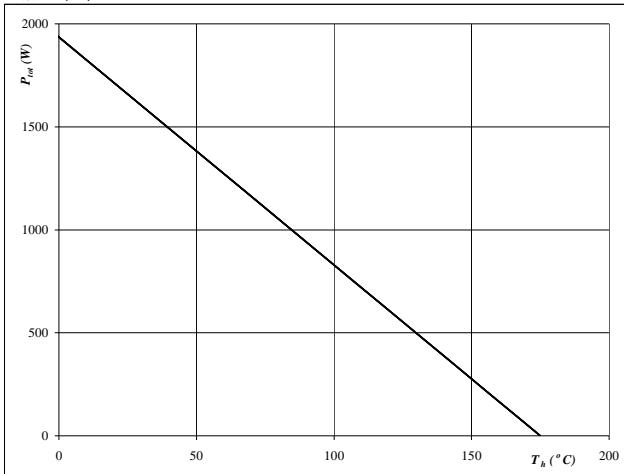
Buck operation

Half Bridge IGBT (T1,T4) and Neutral Point FWD (D2,D3)

Figure 21

Power dissipation as a function of heatsink temperature

$$P_{\text{tot}} = f(T_h)$$



At

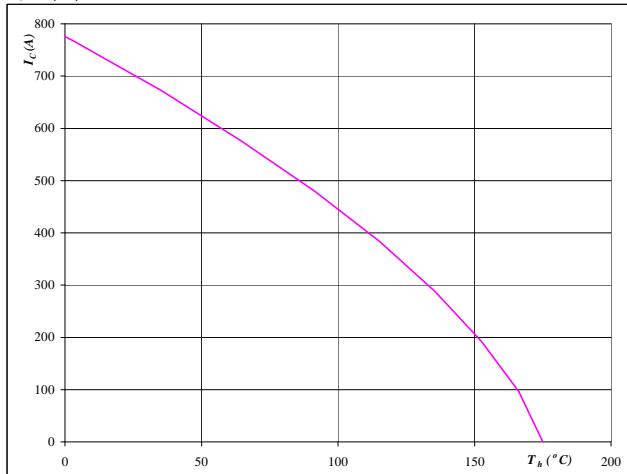
$$T_j = 175 \text{ } ^\circ\text{C}$$

IGBT

Figure 22

Collector current as a function of heatsink temperature

$$I_C = f(T_h)$$



At

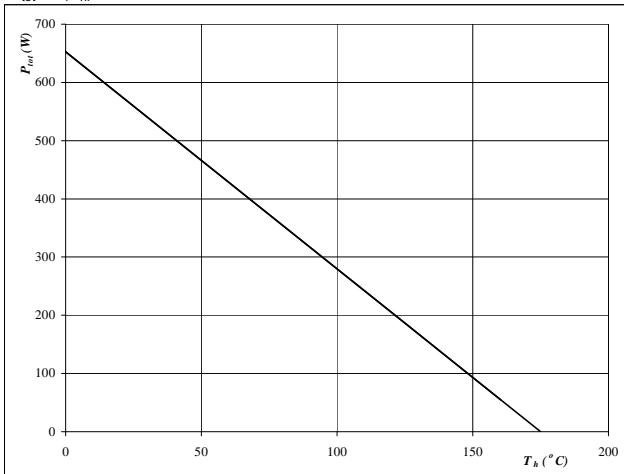
$$T_j = 175 \text{ } ^\circ\text{C}$$

$$V_{GE} = 15 \text{ V}$$

Figure 23

Power dissipation as a function of heatsink temperature

$$P_{\text{tot}} = f(T_h)$$



At

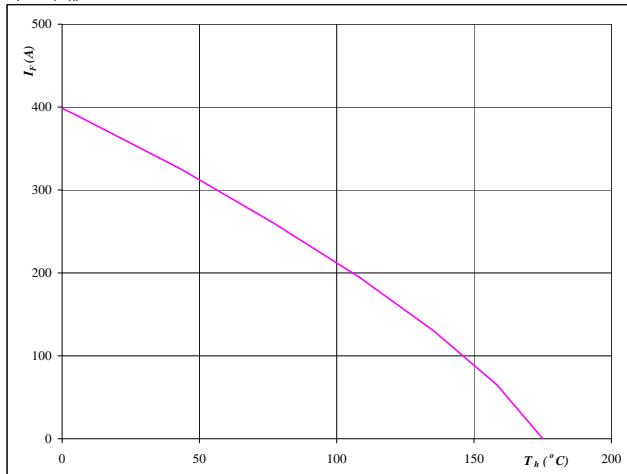
$$T_j = 175 \text{ } ^\circ\text{C}$$

FWD

Figure 24

Forward current as a function of heatsink temperature

$$I_F = f(T_h)$$



At

$$T_j = 175 \text{ } ^\circ\text{C}$$

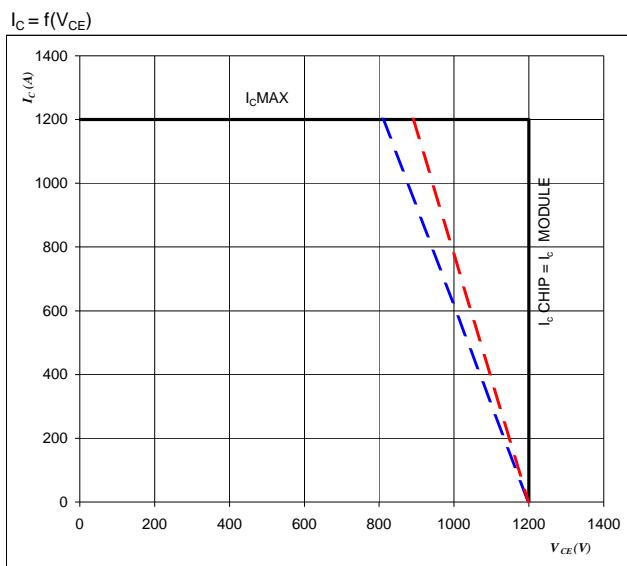
Buck operation

Half Bridge IGBT (T1,T4) and Neutral Point FWD (D2,D3)

Figure 21

IGBT

Reverse bias safe operating area



At

$T_j = 25, 150 \text{ } ^\circ\text{C}$

$U_{ccminus} = U_{ccplus} = U_{cc}/2$

$V_{GE} = \pm 15 \text{ V}$

$R_{gon} = 1 \Omega$

Switching mode: 3 level cont
 2 level dashed

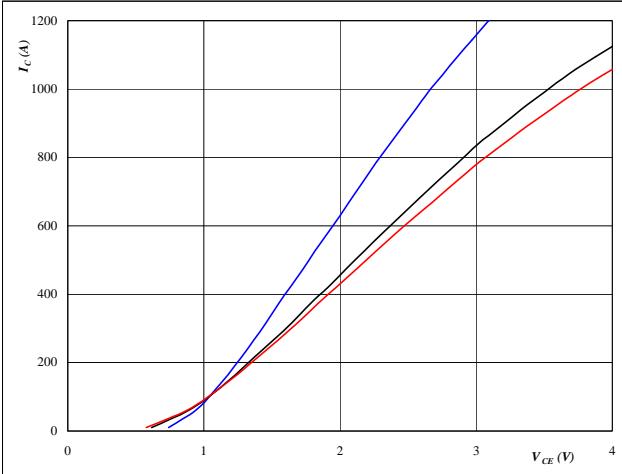
Boost operation

neutral point IGBT (T2,T3) and half bridge FWD (D1,D4)

Figure 1

Typical output characteristics $V_{GE}=15V$

$$I_C = f(V_{CE})$$



IGBT

At

$$t_p = 350 \mu s$$

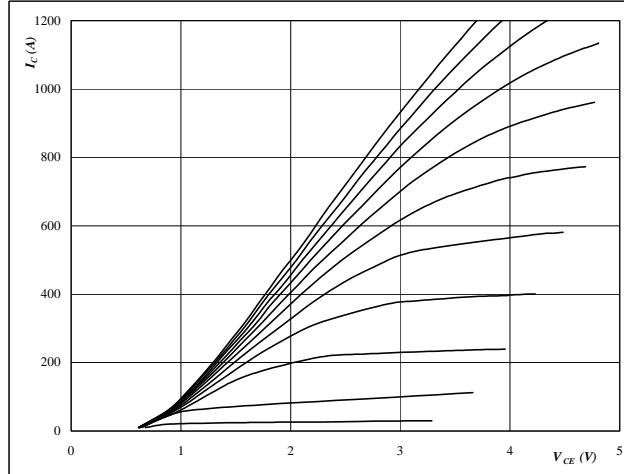
$$T_j = 25/125/150 ^\circ C$$

$$V_{GE} = 15 V$$

Figure 2

Typical output characteristics

$$I_C = f(V_{CE})$$



IGBT

At

$$t_p = 350 \mu s$$

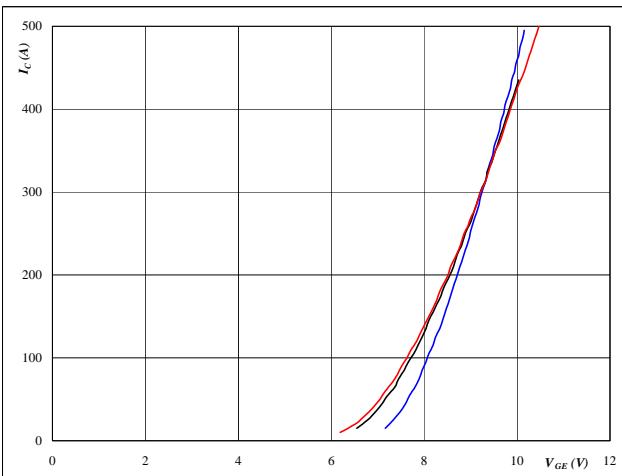
$$T_j = 151 ^\circ C$$

V_{GE} from 7 V to 17 V in steps of 1 V

Figure 3

Typical transfer characteristics

$$I_C = f(V_{GE})$$



IGBT

At

$$t_p = 350 \mu s$$

$$V_{CE} = 350 V$$

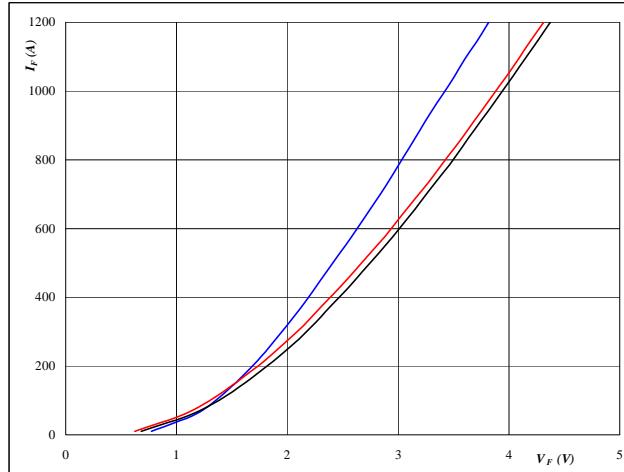
$$T_j = 25/125/150 ^\circ C$$

Figure 4

Typical FWD forward current as

a function of forward voltage

$$I_F = f(V_F)$$



FWD

At

$$t_p = 350 \mu s$$

$$T_j = 25/125/150 ^\circ C$$

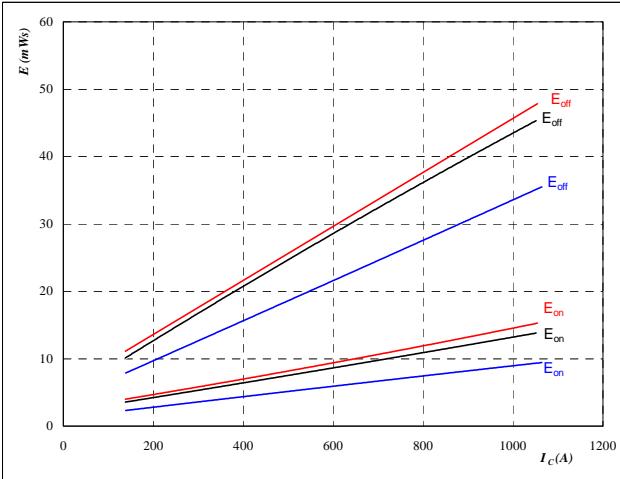
Boost operation

neutral point IGBT (T2,T3) and half bridge FWD (D1,D4)

Figure 5

Typical switching energy losses
as a function of collector current

$$E = f(I_C)$$



With an inductive load at

$$T_j = 25/125/150 \text{ } ^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

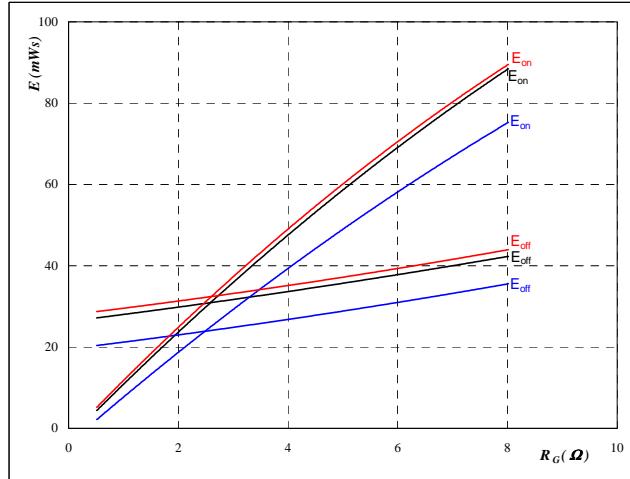
$$R_{gon} = 1,0 \text{ } \Omega$$

$$R_{goff} = 1 \text{ } \Omega$$

Figure 6

Typical switching energy losses
as a function of gate resistor

$$E = f(R_G)$$



With an inductive load at

$$T_j = 25/125/150 \text{ } ^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

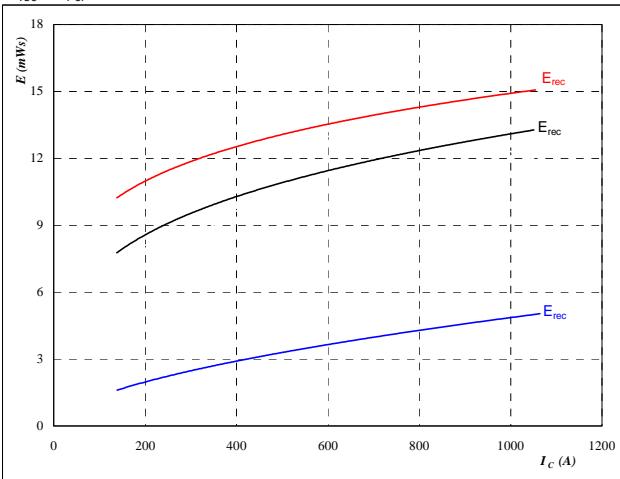
$$V_{GE} = \pm 15 \text{ V}$$

$$I_C = 600 \text{ A}$$

Figure 7

Typical reverse recovery energy loss
as a function of collector current

$$E_{rec} = f(I_C)$$



With an inductive load at

$$T_j = 25/125/150 \text{ } ^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

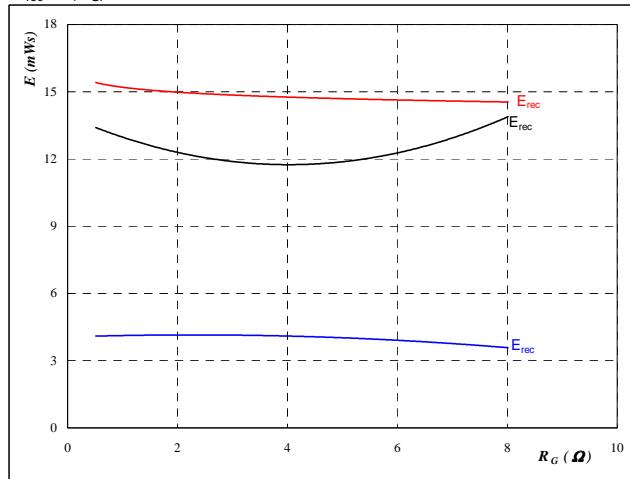
$$V_{GE} = \pm 15 \text{ V}$$

$$R_{gon} = 1 \text{ } \Omega$$

Figure 8

Typical reverse recovery energy loss
as a function of gate resistor

$$E_{rec} = f(R_G)$$



With an inductive load at

$$T_j = 25/125/150 \text{ } ^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

$$I_C = 600 \text{ A}$$

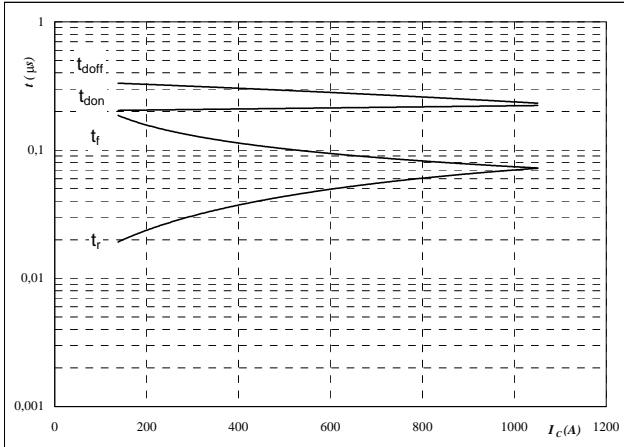
Boost operation

neutral point IGBT (T2,T3) and half bridge FWD (D1,D4)

Figure 9

Typical switching times as a function of collector current

$$t = f(I_C)$$



With an inductive load at

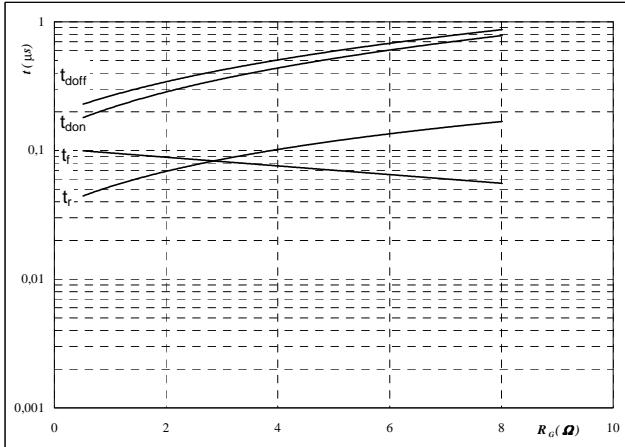
$$\begin{aligned} T_j &= 126 \quad ^\circ\text{C} \\ V_{CE} &= 350 \quad \text{V} \\ V_{GE} &= \pm 15 \quad \text{V} \\ R_{gon} &= 1 \quad \Omega \\ R_{goff} &= 1 \quad \Omega \end{aligned}$$

IGBT

Figure 10

Typical switching times as a function of gate resistor

$$t = f(R_G)$$



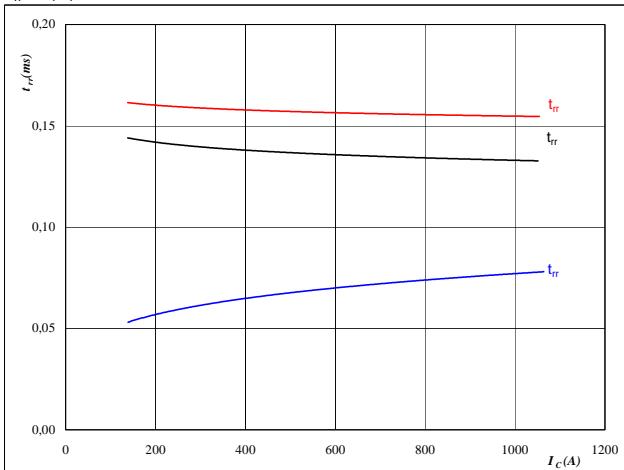
With an inductive load at

$$\begin{aligned} T_j &= 126 \quad ^\circ\text{C} \\ V_{CE} &= 350 \quad \text{V} \\ V_{GE} &= \pm 15 \quad \text{V} \\ I_C &= 600 \quad \text{A} \end{aligned}$$

Figure 11

Typical reverse recovery time as a function of collector current

$$t_{rr} = f(I_C)$$



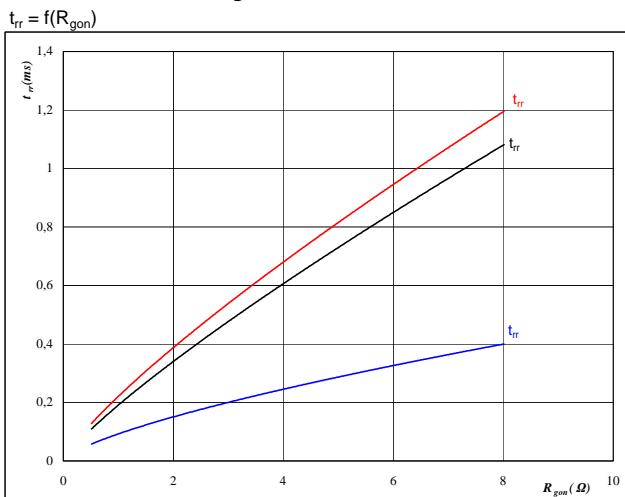
At

$$\begin{aligned} T_j &= 25/125/150 \quad ^\circ\text{C} \\ V_{CE} &= 350 \quad \text{V} \\ V_{GE} &= \pm 15 \quad \text{V} \\ R_{gon} &= 1 \quad \Omega \end{aligned}$$

FWD

Typical reverse recovery time as a function of IGBT turn on gate resistor

$$t_{rr} = f(R_{gon})$$



At

$$\begin{aligned} T_j &= 25/125/150 \quad ^\circ\text{C} \\ V_R &= 350 \quad \text{V} \\ I_F &= 600 \quad \text{A} \\ V_{GE} &= \pm 15 \quad \text{V} \end{aligned}$$

Boost operation

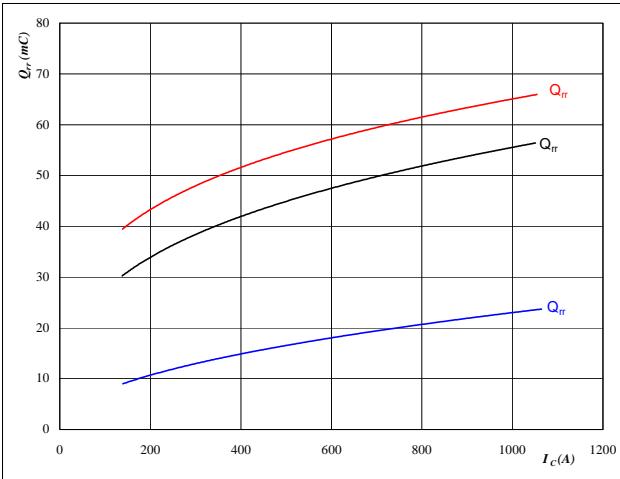
neutral point IGBT (T2,T3) and half bridge FWD (D1,D4)

Figure 13

Typical reverse recovery charge as a function of collector current

$$Q_{rr} = f(I_C)$$

FWD



At

$$T_j = 25/125/150 \text{ } ^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

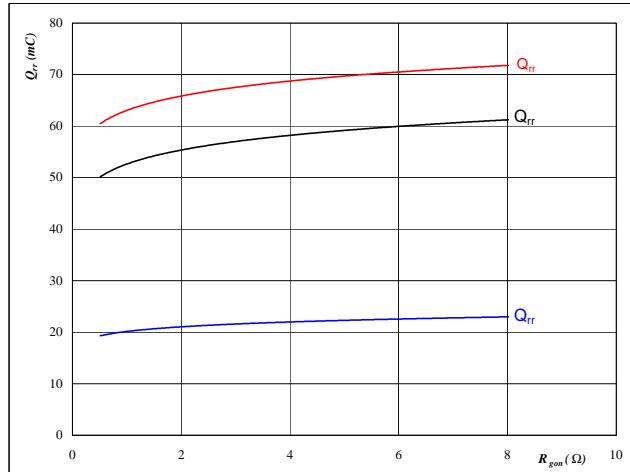
$$R_{gon} = 1 \Omega$$

Figure 14

Typical reverse recovery charge as a function of IGBT turn on gate resistor

$$Q_{rr} = f(R_{gon})$$

FWD



At

$$T_j = 25/125/150 \text{ } ^\circ\text{C}$$

$$V_R = 350 \text{ V}$$

$$I_F = 600 \text{ A}$$

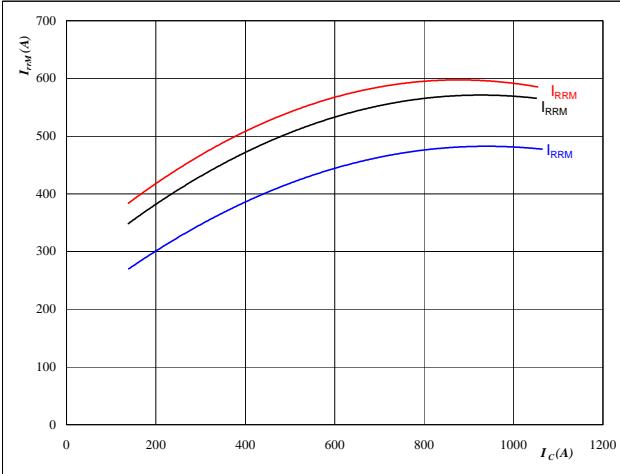
$$V_{GE} = \pm 15 \text{ V}$$

Figure 15

Typical reverse recovery current as a function of collector current

$$I_{RRM} = f(I_C)$$

FWD



At

$$T_j = 25/125/150 \text{ } ^\circ\text{C}$$

$$V_{CE} = 350 \text{ V}$$

$$V_{GE} = \pm 15 \text{ V}$$

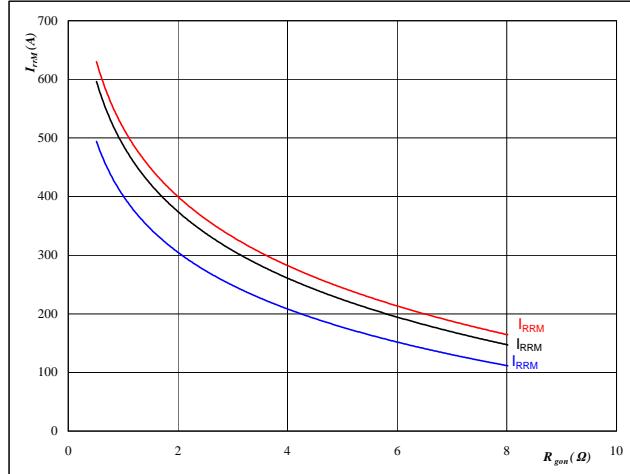
$$R_{gon} = 1 \Omega$$

Figure 16

Typical reverse recovery current as a function of IGBT turn on gate resistor

$$I_{RRM} = f(R_{gon})$$

FWD



At

$$T_j = 25/125/150 \text{ } ^\circ\text{C}$$

$$V_R = 350 \text{ V}$$

$$I_F = 600 \text{ A}$$

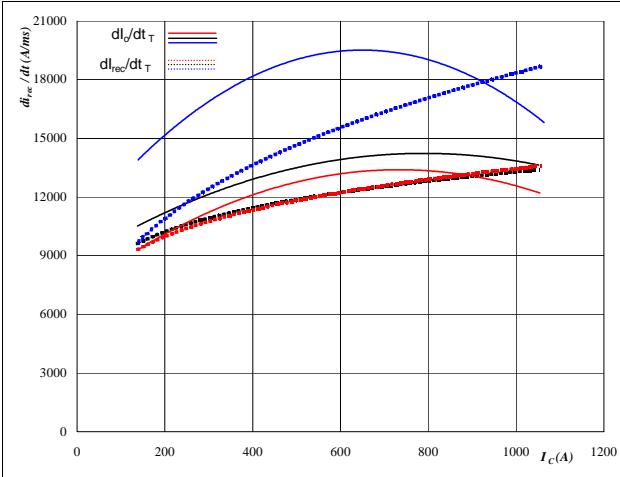
$$V_{GE} = \pm 15 \text{ V}$$

Boost operation

neutral point IGBT (T2,T3) and half bridge FWD (D1,D4)

Figure 17

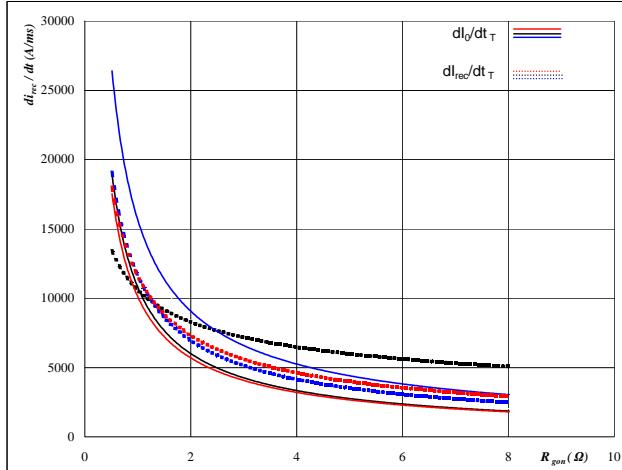
Typical rate of fall of forward
and reverse recovery current as a
function of collector current
 $dl_0/dt, dl_{rec}/dt = f(I_c)$


At

$T_j = 25/125/150 \text{ }^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 1 \Omega$

FWD
Figure 18

Typical rate of fall of forward
and reverse recovery current as a
function of IGBT turn on gate resistor
 $dl_0/dt, dl_{rec}/dt = f(R_{gon})$

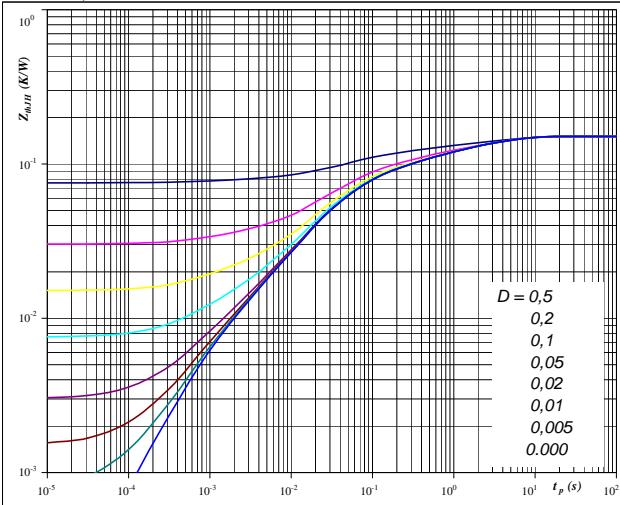

At

$T_j = 25/125/150 \text{ }^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 600 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$

Figure 19

IGBT transient thermal impedance
as a function of pulse width

$$Z_{thJH} = f(t_p)$$

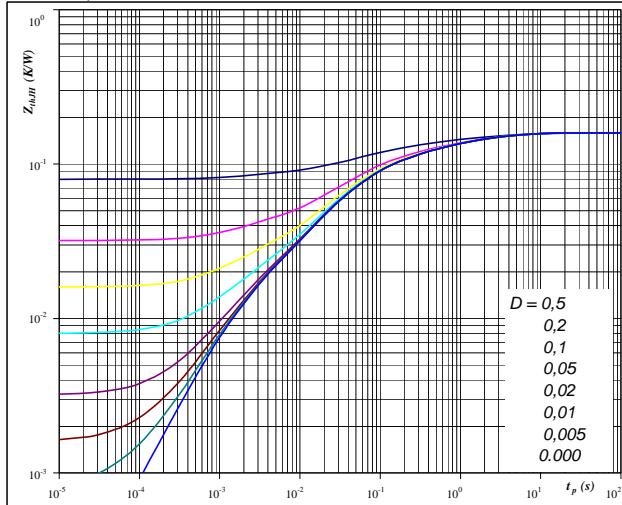

At

$D = t_p / T$
 Preapplied PCM $R_{thJH} = 0,15 \text{ K/W}$ Thermal grease $R_{thJH} = 0,17 \text{ K/W}$

IGBT
Figure 20

FWD transient thermal impedance
as a function of pulse width

$$Z_{thJH} = f(t_p)$$


At

$D = t_p / T$
 Preapplied PCM $R_{thJH} = 0,16 \text{ K/W}$ Thermal grease $R_{thJH} = 0,18 \text{ K/W}$

FWD thermal model values

100um preapplied PCM		100um grease 1W/mK (P12)	
R (C/W)	Tau (s)	R (C/W)	Tau (s)
0,029	2,074	0,033	2,074
0,027	0,416	0,030	0,416
0,030	0,086	0,034	0,086
0,048	0,018	0,054	0,018
0,011	0,005	0,012	0,005

100um preapplied PCM		100um grease 1W/mK (P12)	
R (C/W)	Tau (s)	R (C/W)	Tau (s)
0,015	5,007	0,0165	5,007
0,026	1,172	0,0296	1,172
0,033	0,251	0,0377	0,251
0,052	0,054	0,0590	0,054
0,022	0,015	0,0252	0,015

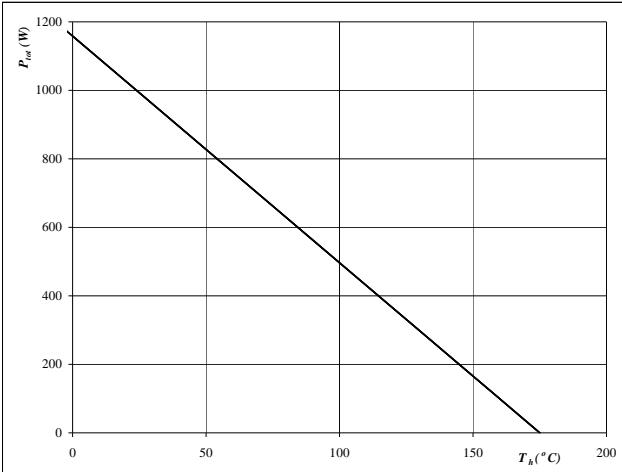
Boost operation

neutral point IGBT (T2,T3) and half bridge FWD (D1,D4)

Figure 21

Power dissipation as a function of heatsink temperature

$$P_{\text{tot}} = f(T_h)$$



At

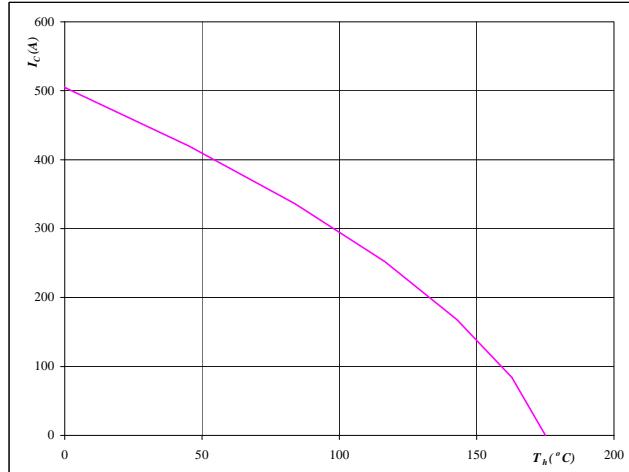
$$T_j = 175 \quad {}^\circ\text{C}$$

IGBT

Figure 22

Collector current as a function of heatsink temperature

$$I_C = f(T_h)$$



At

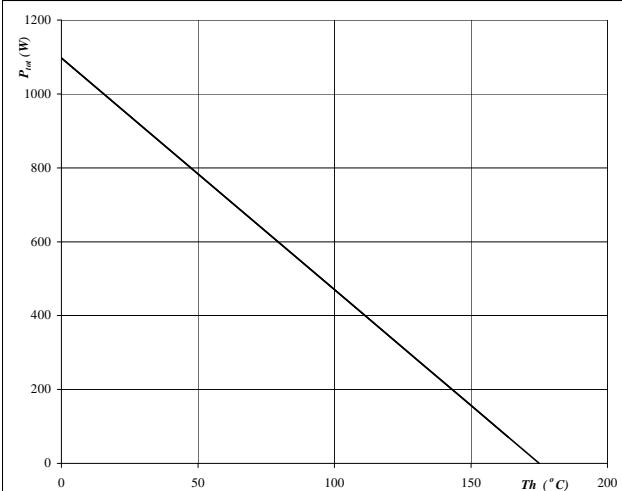
$$T_j = 175 \quad {}^\circ\text{C}$$

$$V_{GE} = 15 \quad \text{V}$$

Figure 23

Power dissipation as a function of heatsink temperature

$$P_{\text{tot}} = f(T_h)$$



At

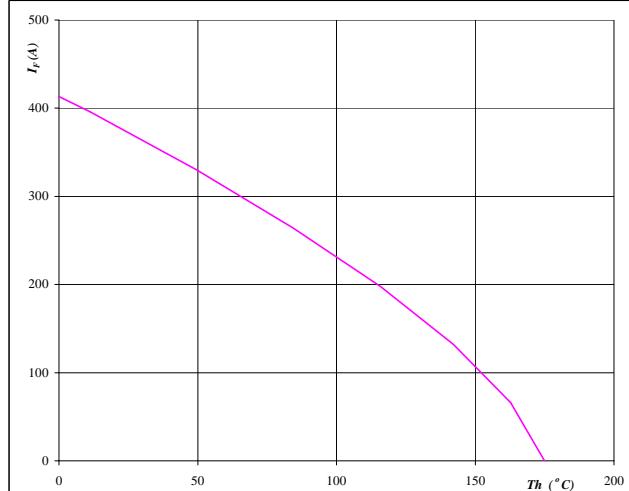
$$T_j = 175 \quad {}^\circ\text{C}$$

FWD

Figure 24

Forward current as a function of heatsink temperature

$$I_F = f(T_h)$$



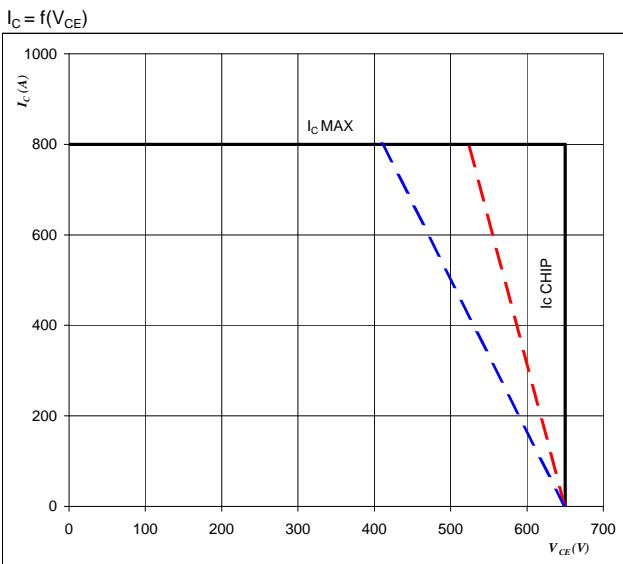
At

$$T_j = 175 \quad {}^\circ\text{C}$$

Boost operation

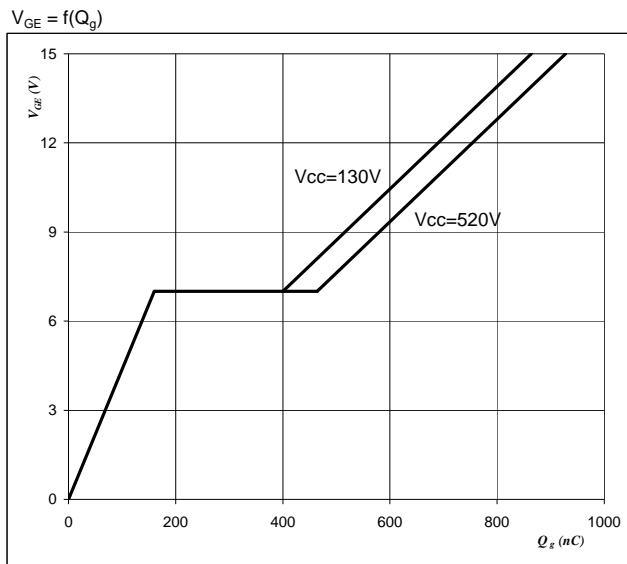
neutral point IGBT (T2,T3) and half bridge FWD (D1,D4)

Figure 25
Reverse bias safe operating area



At
 $T_j = 25 \setminus 150^\circ C$
 $U_{ccminus} = U_{ccplus} = U_{C_c}/2$
 $V_{GE} = \pm 15 V$
 $R_{gon} = 1 \Omega$

Figure 22
Gate voltage vs Gate charge



At
 $I_c = 400 A$

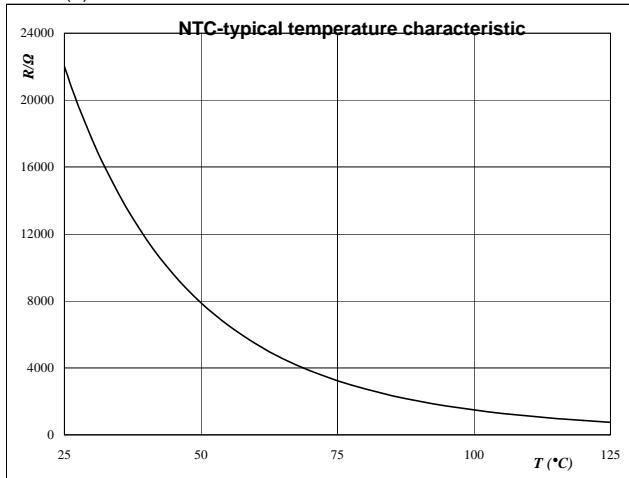
Thermistor

Figure 1

Thermistor

Typical NTC characteristic
as a function of temperature

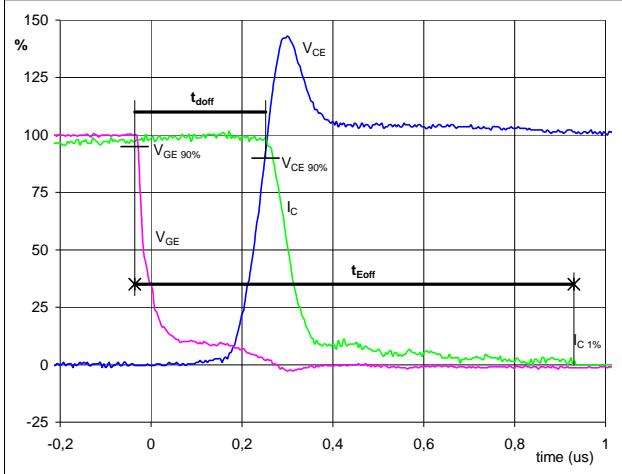
$$R_T = f(T)$$



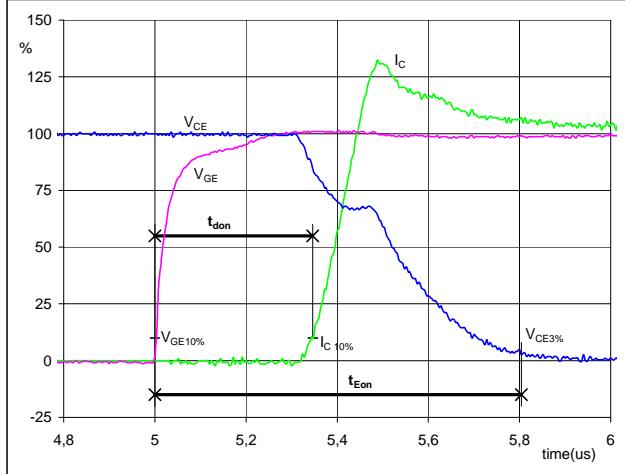
Switching Definitions Half Bridge

General conditions

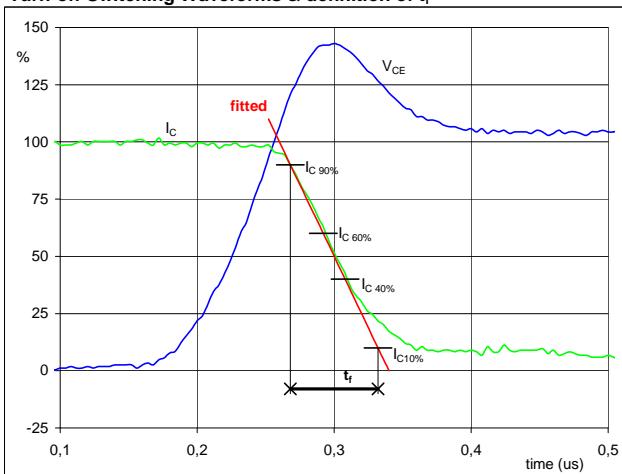
T_j	=	125 °C
R_{gon}	=	0,5 Ω
R_{goff}	=	0,5 Ω

Figure 1
Half Bridge IGBT
Turn-off Switching Waveforms & definition of t_{doff} , t_{Eoff}
 $(t_{Eoff} = \text{integrating time for } E_{off})$


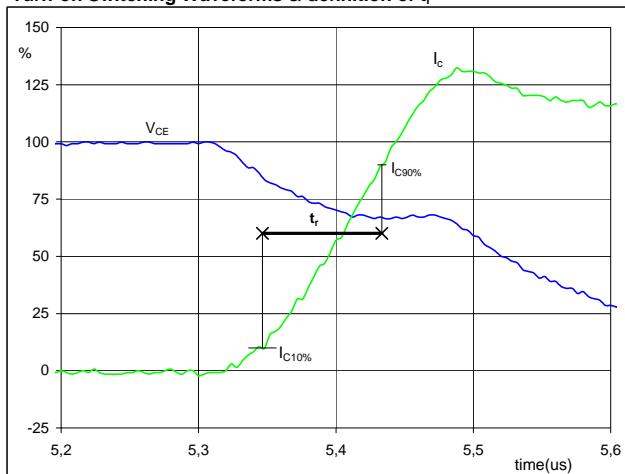
$V_{GE}(0\%) = -15$ V
 $V_{GE}(100\%) = 15$ V
 $V_C(100\%) = 350$ V
 $I_C(100\%) = 599$ A
 $t_{doff} = 0,27$ μs
 $t_{Eoff} = 0,97$ μs

Figure 2
Half Bridge IGBT
Turn-on Switching Waveforms & definition of t_{don} , t_{Eon}
 $(t_{Eon} = \text{integrating time for } E_{on})$


$V_{GE}(0\%) = -15$ V
 $V_{GE}(100\%) = 15$ V
 $V_C(100\%) = 350$ V
 $I_C(100\%) = 599$ A
 $t_{don} = 0,34$ μs
 $t_{Eon} = 0,80$ μs

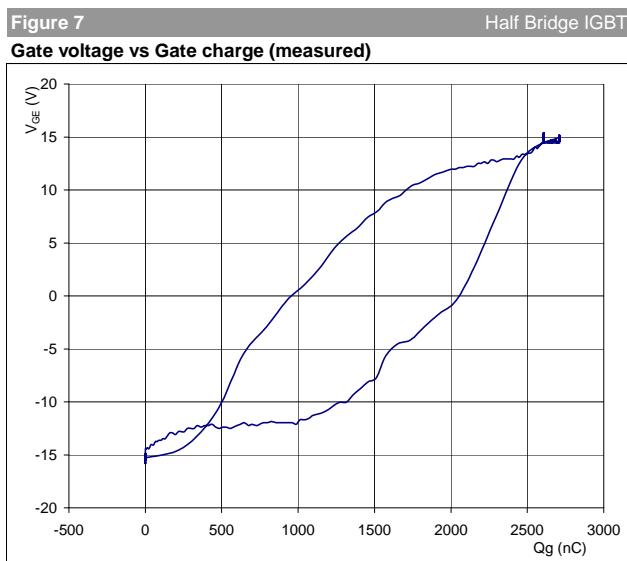
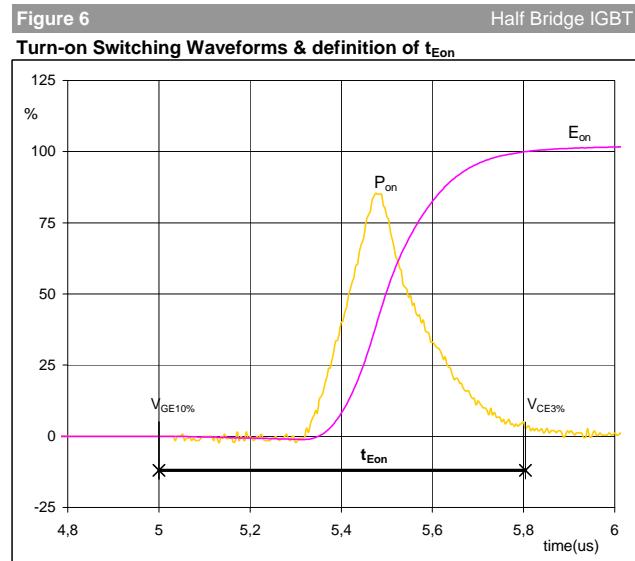
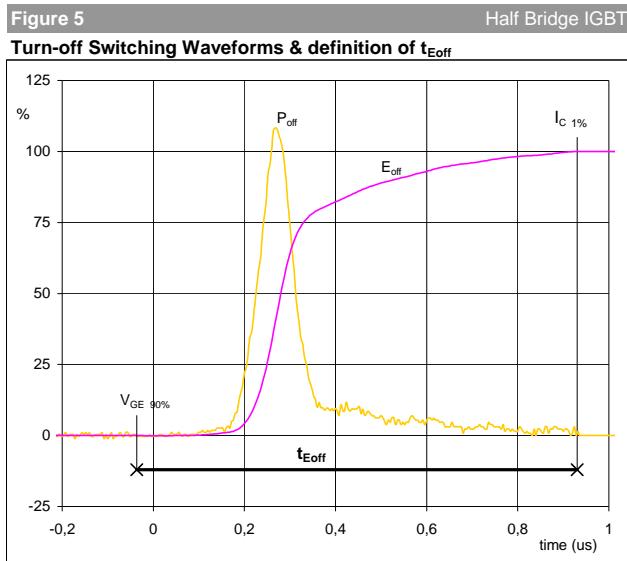
Figure 3
Half Bridge IGBT
Turn-off Switching Waveforms & definition of t_f


$V_C(100\%) = 350$ V
 $I_C(100\%) = 599$ A
 $t_f = 0,07$ μs

Figure 4
Half Bridge IGBT
Turn-on Switching Waveforms & definition of t_r


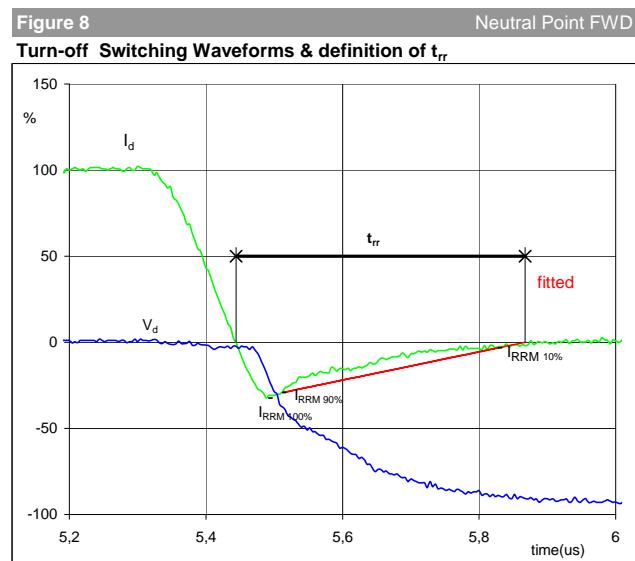
$V_C(100\%) = 350$ V
 $I_C(100\%) = 599$ A
 $t_r = 0,09$ μs

Switching Definitions Half Bridge

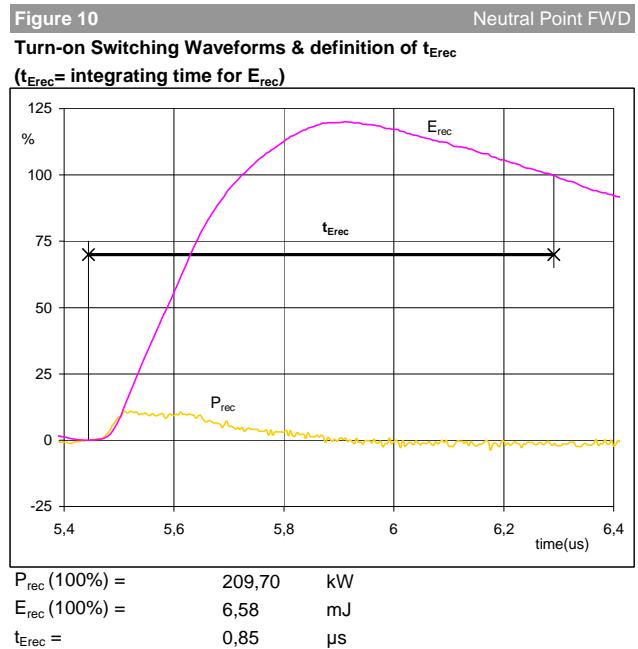
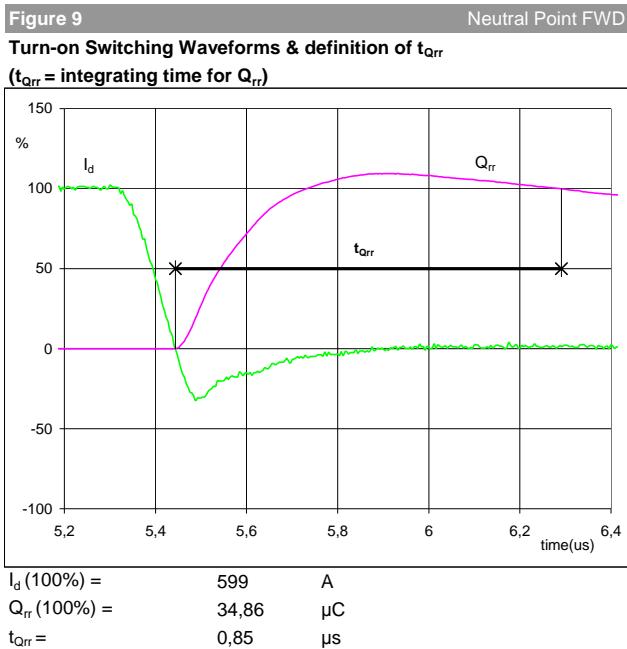


Measured values:

- $V_{GEoff} = -15 \text{ V}$
- $V_{GEon} = 15 \text{ V}$
- $V_C (100\%) = 350 \text{ V}$
- $I_C (100\%) = 599 \text{ A}$
- $Q_g = 2710,20 \text{ nC}$

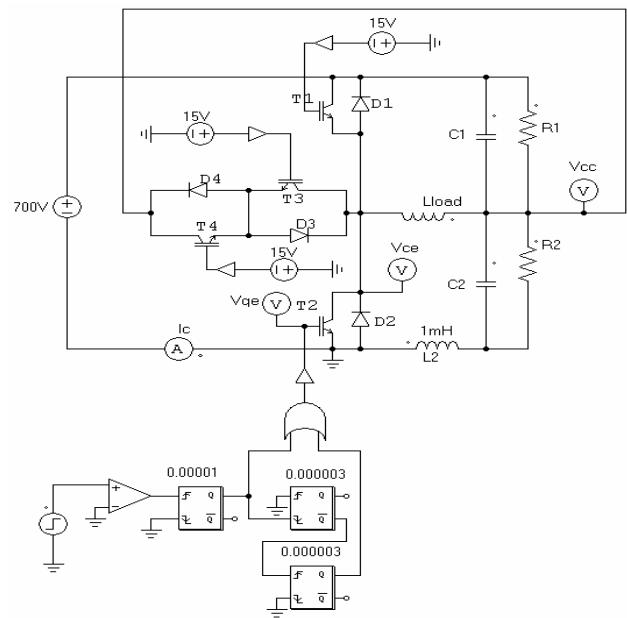


Switching Definitions Half Bridge



Half Bridge switching measurement circuit

Figure 11



Switching Definitions Neutral Point

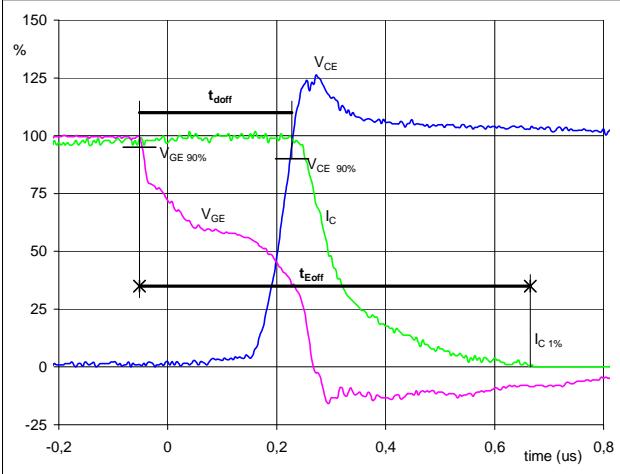
General conditions

T_j	=	125 °C
R_{gon}	=	1 Ω
R_{goff}	=	1 Ω

Figure 1

Neutral Point IGBT

Turn-off Switching Waveforms & definition of t_{doff} , t_{Eoff}
 $(t_{Eoff} = \text{integrating time for } E_{off})$

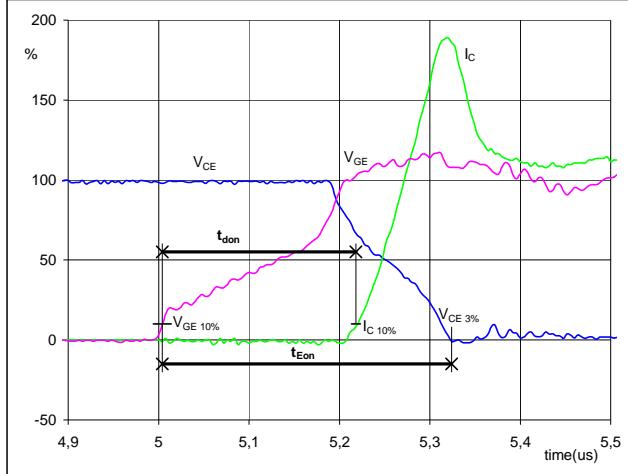


$V_{GE}(0\%) = -15$ V
 $V_{GE}(100\%) = 15$ V
 $V_C(100\%) = 350$ V
 $I_C(100\%) = 601$ A
 $t_{doff} = 0,23$ μs
 $t_{Eoff} = 0,58$ μs

Figure 2

Neutral Point IGBT

Turn-on Switching Waveforms & definition of t_{don} , t_{Eon}
 $(t_{Eon} = \text{integrating time for } E_{on})$

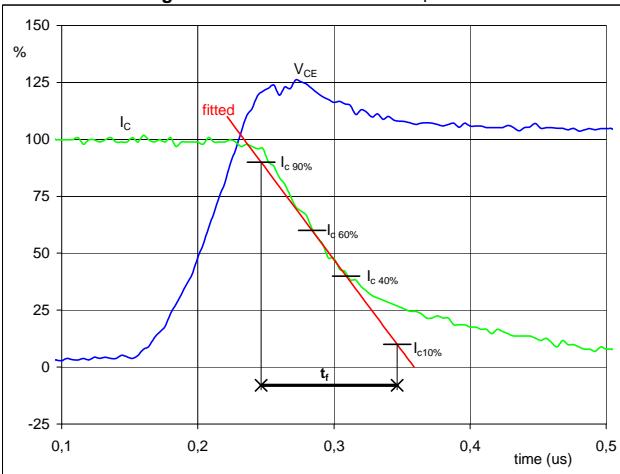


$V_{GE}(0\%) = -15$ V
 $V_{GE}(100\%) = 15$ V
 $V_C(100\%) = 350$ V
 $I_C(100\%) = 601$ A
 $t_{don} = 0,21$ μs
 $t_{Eon} = 0,38$ μs

Figure 3

Neutral Point IGBT

Turn-off Switching Waveforms & definition of t_f

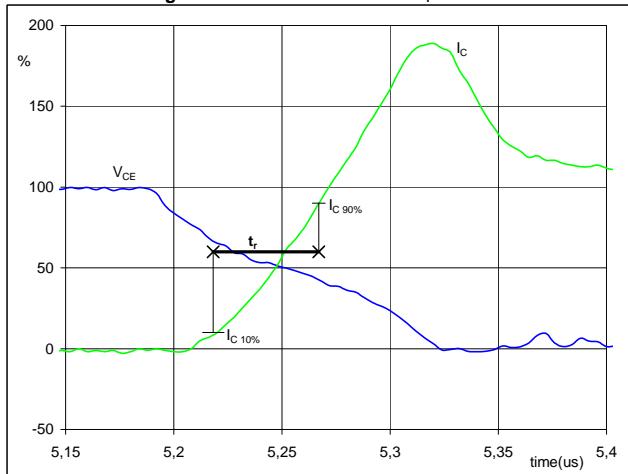


$V_C(100\%) = 350$ V
 $I_C(100\%) = 601$ A
 $t_f = 0,106$ μs

Figure 4

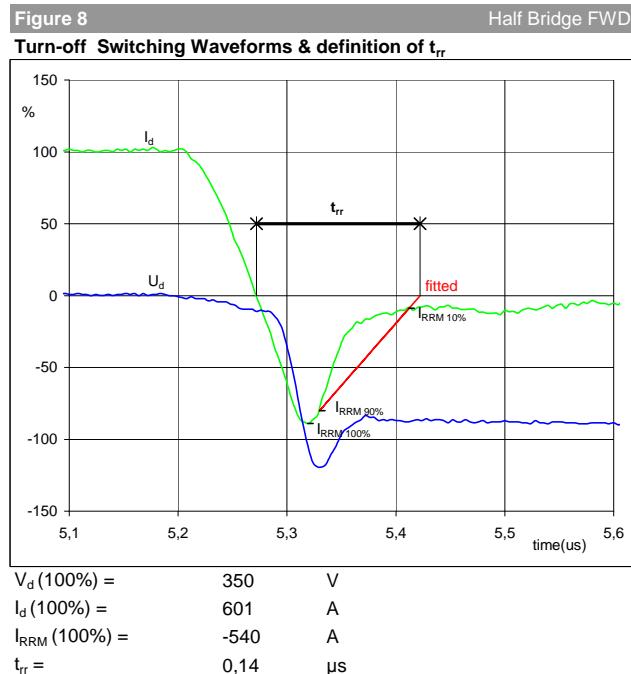
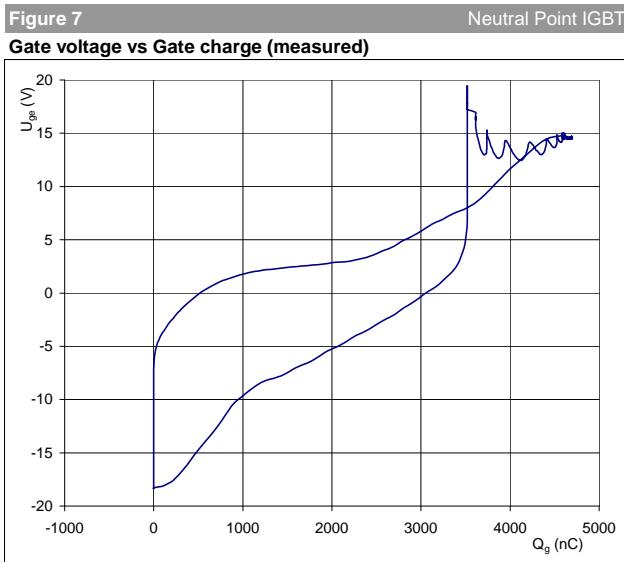
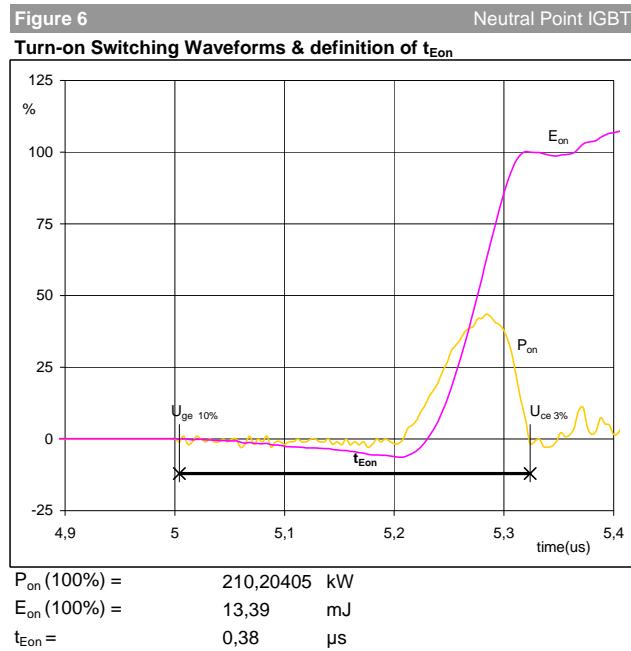
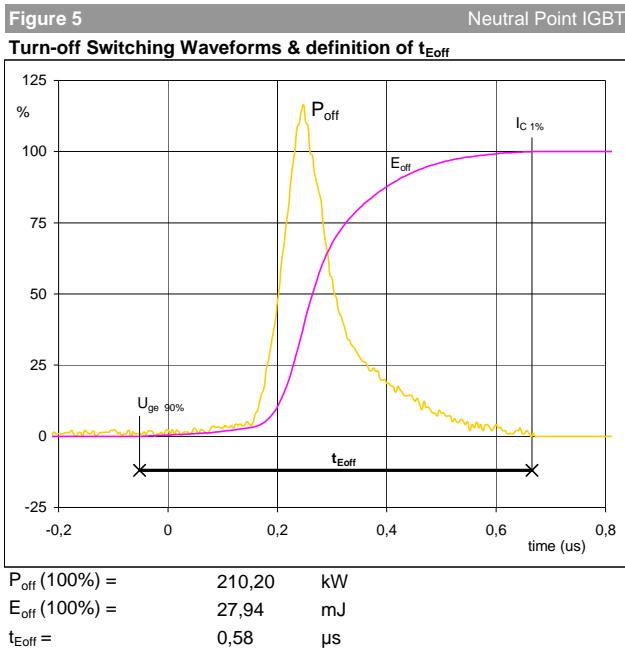
Neutral Point IGBT

Turn-on Switching Waveforms & definition of t_r

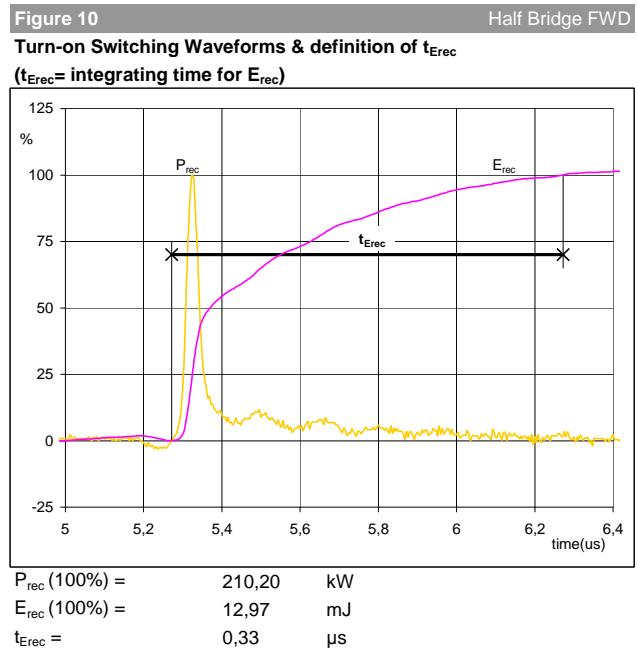
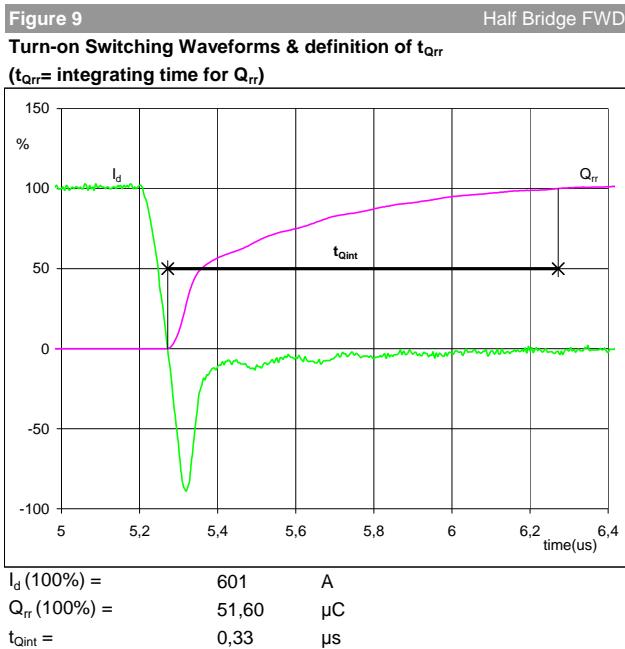


$V_C(100\%) = 350$ V
 $I_C(100\%) = 601$ A
 $t_r = 0,049$ μs

Switching Definitions Neutral Point

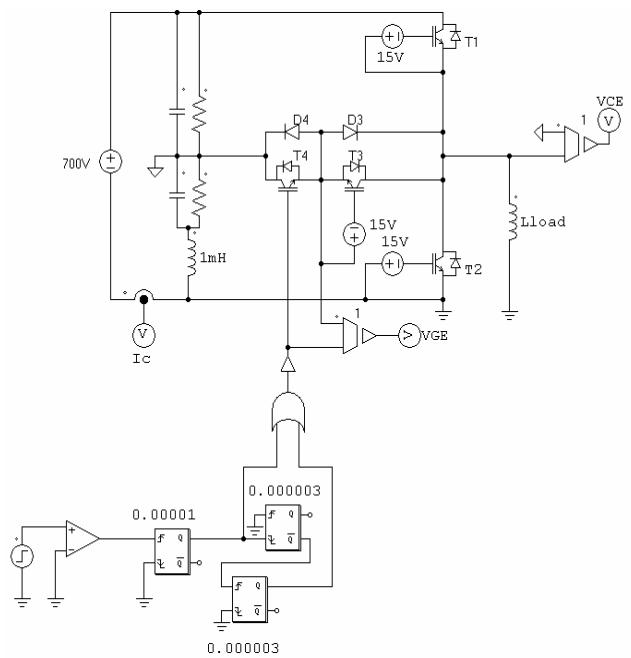


Switching Definitions Neutral Point



Neutral Point switching measurement circuit

Figure 11

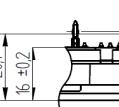


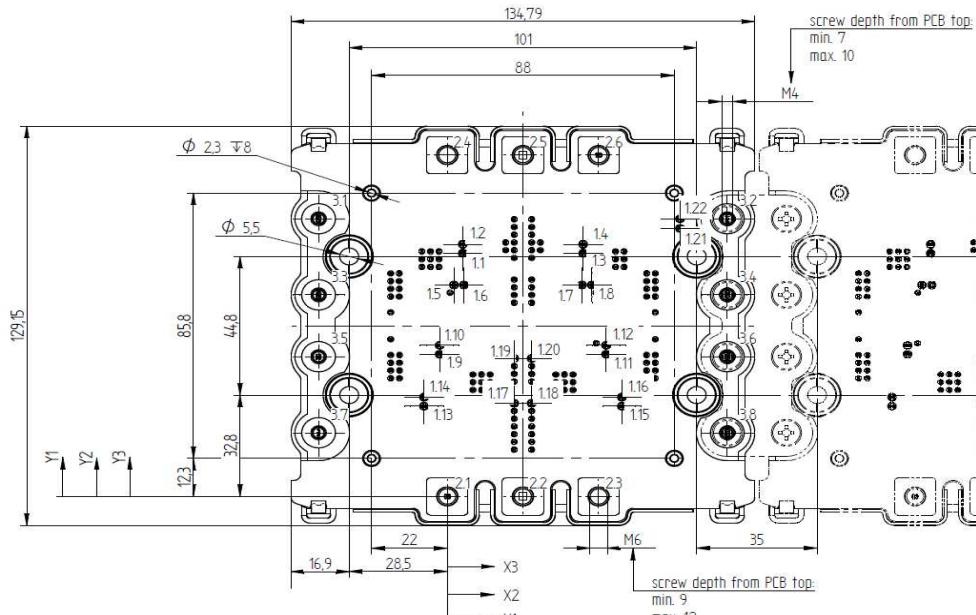
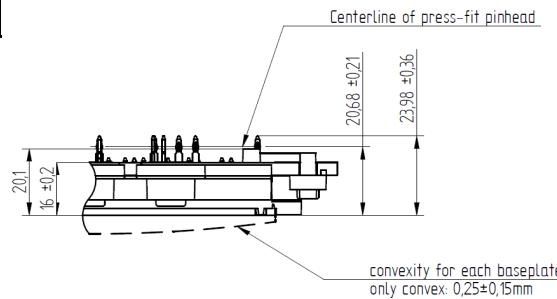
Ordering Code and Marking - Outline - Pinout

Ordering Code & Marking			
Version	Ordering Code	in DataMatrix as	in packaging barcode as
without PCM	70-W212NMA600NB02-M200P62	M200P62	M200P62
with PCM	70-W212NMA600NB02-M200P62-/3/	M200P62	M200P62-/3/

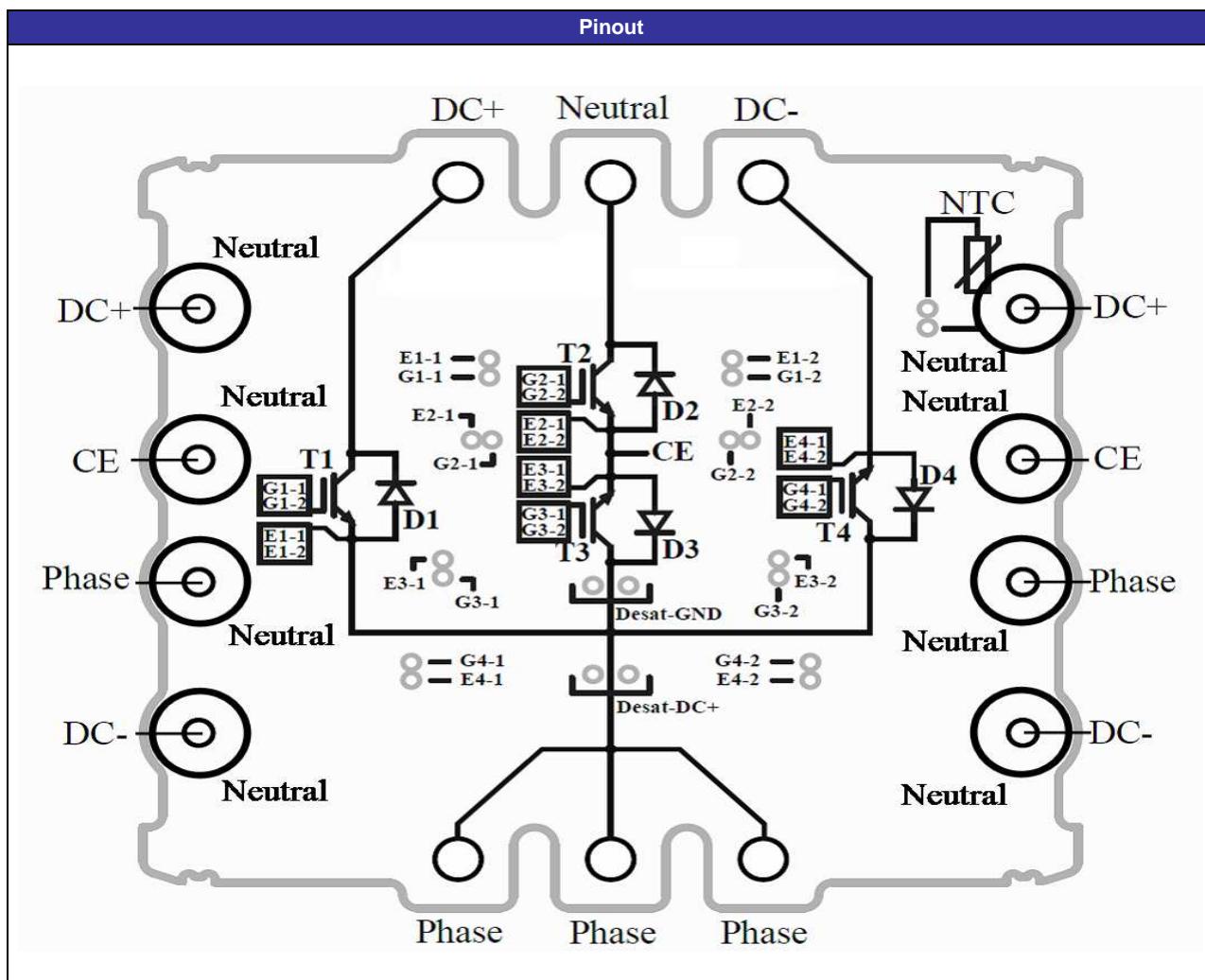
Outline

Driver pins					Low current connections				Power connections			
Pin	X1	Y1	Function	Group	M4 screw	X3	Y3	Function	M6 screw	X2	Y2	Function
1.1	4,5	78,65	G1-1	T1		3,1	-37,4	89,8	DC+	2,1	0	0
1.2	4,5	81,55	E1-1	T1		3,2	81,4	89,8	DC+	2,2	22	0
1.3	39,5	78,65	G1-2	T1		3,3	-37,4	65,2	CE	2,3	44	0
1.4	39,5	81,55	E1-2	T1		3,4	81,4	65,2	CE	2,4	0	110,4
1.5	1,95	68,4	E2-1	T2		3,5	-37,4	45,2	Phase	2,5	22	110,4
1.6	4,85	68,4	G2-1	T2		3,6	81,4	45,2	Phase	2,6	44	110,4
1.7	39,15	68,4	G2-2	T2		3,7	-37,4	20,6	DC-			DC-
1.8	42,05	68,4	E2-2	T2		3,8	81,4	20,6	DC-			
1.9	-2,2	46	G3-1	T3								
1.10	-2,2	48,9	E3-1	T3								
1.11	46,2	46	G3-2	T3								
1.12	46,2	48,9	E3-2	T3								
1.13	-6,75	29,2	E4-1	T4								
1.14	-6,75	32,1	G4-1	T4								
1.15	50,75	29,2	E4-2	T4								
1.16	50,75	32,1	G4-2	T4								
1.17	19,45	30,15	Desat-DC+									
1.18	24,55	30,15	Desat-DC+									
1.19	19,45	44,65	Desat-GND									
1.20	24,55	44,65	Desat-GND									
1.21	67,65	86,7	NTC									
1.22	67,65	89,8	NTC									





Ordering Code and Marking - Outline - Pinout



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2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.