

flowMNPC 1

1200 V / 160 A

Features

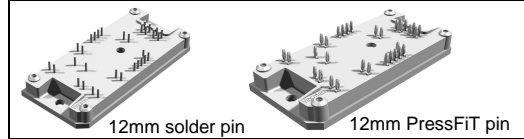
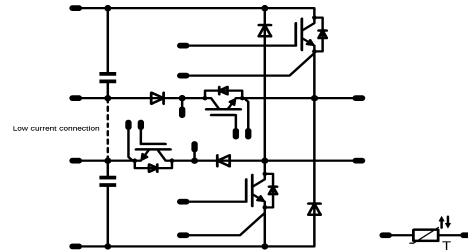
- mixed voltage NPC topology
- reactive power capability
- low inductance layout
- Split output
- Common collector neutral connection

Target Applications

- solar inverter
- UPS
- Active frontend

Types

- 10-FY12NMA160SH-M420F
- 10-PY12NMA160SH-M420FY

flow1 12mm housing

Schematic


Maximum Ratings

 T_j=25°C, unless otherwise specified

Parameter	Symbol	Condition	Value	Unit
Halfbridge IGBT Inverse Diode (D1, D4)				
Repetitive peak reverse voltage	V _{RRM}		1200	V
DC forward current	I _F	T _j =T _{jmax} T _n =80°C T _c =80°C	14 19	A
Repetitive peak forward current	I _{FRM}	t _p =10ms	14	A
Power dissipation per Diode	P _{tot}	T _n =80°C T _c =80°C	31 47	W
Maximum Junction Temperature	T _{jmax}		150	°C

Halfbridge IGBT (T1, T4)

Collector-emitter break down voltage	V _{CE}		1200	V
DC collector current	I _C	T _j =T _{jmax} T _n =80°C T _c =80°C	116 156	A
Repetitive peak collector current	I _{Cpulse}	t _p limited by T _{jmax}	640	A
Power dissipation per IGBT	P _{tot}	T _j =T _{jmax} T _n =80°C T _c =80°C	260 394	W
Gate-emitter peak voltage	V _{GE}		±20	V
Short circuit ratings	t _{SC} V _{CC}	T _j ≤150°C V _{GE} =15V	10 600	μs V
Maximum Junction Temperature	T _{jmax}		175	°C

Maximum Ratings

 $T_j=25^{\circ}\text{C}$, unless otherwise specified

Parameter	Symbol	Condition	Value	Unit
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NP Diode (D7, D8)

Peak Repetitive Reverse Voltage	V_{RRM}	$T_j=25^{\circ}\text{C}$	600	V
DC forward current	I_F	$T_j=T_{jmax}$	$T_h=80^{\circ}\text{C}$ 90 $T_c=80^{\circ}\text{C}$	A
Repetitive peak forward current	I_{FRM}	t_p limited by T_{jmax}	240	A
Power dissipation per Diode	P_{tot}	$T_j=T_{jmax}$	$T_h=80^{\circ}\text{C}$ 101 $T_c=80^{\circ}\text{C}$	W
Maximum Junction Temperature	T_{jmax}		150	$^{\circ}\text{C}$

NP IGBT (T2, T3)

Collector-emitter break down voltage	V_{CE}		600	V
DC collector current	I_C	$T_j=T_{jmax}$	$T_h=80^{\circ}\text{C}$ 83 $T_c=80^{\circ}\text{C}$	A
Repetitive peak collector current	I_{cpulse}	t_p limited by T_{jmax}	300	A
Power dissipation per IGBT	P_{tot}	$T_j=T_{jmax}$	$T_h=80^{\circ}\text{C}$ 142 $T_c=80^{\circ}\text{C}$	W
Gate-emitter peak voltage	V_{GE}		± 20	V
Short circuit ratings	t_{SC} V_{CC}	$T_j \leq 150^{\circ}\text{C}$ $V_{GE}=15\text{V}$	6 360	μs V
Maximum Junction Temperature	T_{jmax}		175	$^{\circ}\text{C}$

NP Inverse Diode (D2, D3)

Peak Repetitive Reverse Voltage	V_{RRM}	$T_c=25^{\circ}\text{C}$	600	V
DC forward current	I_F	$T_j=T_{jmax}$	$T_h=80^{\circ}\text{C}$ 18 $T_c=80^{\circ}\text{C}$	A
Repetitive peak forward current	I_{FRM}	t_p limited by T_{jmax}	30	A
Power dissipation per Diode	P_{tot}	$T_j=T_{jmax}$	$T_h=80^{\circ}\text{C}$ 31 $T_c=80^{\circ}\text{C}$	W
Maximum Junction Temperature	T_{jmax}		150	$^{\circ}\text{C}$

Halfbridge Diode (D5, D6)

Peak Repetitive Reverse Voltage	V_{RRM}	$T_j=25^{\circ}\text{C}$	1200	V
DC forward current	I_F	$T_j=T_{jmax}$	$T_h=80^{\circ}\text{C}$ 50 $T_c=80^{\circ}\text{C}$	A
Repetitive peak forward current	I_{FRM}	t_p limited by T_{jmax}	120	A
Power dissipation per Diode	P_{tot}	$T_j=T_{jmax}$	$T_h=80^{\circ}\text{C}$ 92 $T_c=80^{\circ}\text{C}$	W
Maximum Junction Temperature	T_{jmax}		150	$^{\circ}\text{C}$

DC link Capacitor (C1, C2)

Max.DC voltage	V_{MAX}	$T_c=25^{\circ}\text{C}$	630	V
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Maximum Ratings

T_j=25°C, unless otherwise specified

Parameter	Symbol	Condition	Value	Unit
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Thermal Properties

Storage temperature	T _{stg}		-40...+125	°C
Operation temperature under switching condition	T _{op}		-40...+(T _{jmax} - 25)	°C

Insulation Properties

Insulation voltage	V _{is}	t=2s DC voltage	4000	V
Creepage distance			min 12,7	mm
Clearance			min 12,7	mm
Comparative tracking index	CTI		>200	

Characteristic Values

Parameter	Symbol	Conditions					Value			Unit
		V_{GE} [V] or V_{GS} [V]	V_r [V] or V_{CE} [V] or V_{OS} [V]	I_c [A] or I_f [A] or I_b [A]	T_j	Min	Typ	Max		
Halfbridge IGBT Inverse Diode (D1, D4)										
Forward voltage	V_f				7	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$	1	1,97 1,65	2,7	V
Thermal resistance chip to heatsink per chip	R_{thJH}	Thermal grease thickness \leq 50um						2,24		K/W
Thermal resistance chip to case per chip	R_{thJC}	$\lambda = 1 \text{ W/mK}$						1,48		
Halfbridge IGBT (T1, T4)										
Gate emitter threshold voltage	$V_{GE(th)}$	$V_{CE}=V_{GE}$			0,004	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$	5	5,8	6,5	V
Collector-emitter saturation voltage	$V_{CE(sat)}$		15		160	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$	1	2,02 2,37	2,5	V
Collector-emitter cut-off current incl. Diode	I_{CES}		0	1200		$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$			1	mA
Gate-emitter leakage current	I_{GES}		20	0		$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$			480	nA
Integrated Gate resistor	R_{gint}							none		Ω
Turn-on delay time	$t_{d(ON)}$					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		133 135		ns
Rise time	t_r					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		20 23		
Turn-off delay time	$t_{d(OFF)}$	$R_{goff}=4 \Omega$	± 15	350	100	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		225 276		
Fall time	t_f	$R_{gon}=4 \Omega$				$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		38 64		
Turn-on energy loss per pulse	E_{on}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		1,80 3,18		
Turn-off energy loss per pulse	E_{off}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		2,52 4,03		
Input capacitance	C_{ies}							9320		pF
Output capacitance	C_{oss}	$f=1\text{MHz}$	0	25		$T_j=25^\circ\text{C}$		600		
Reverse transfer capacitance	C_{rss}							520		
Gate charge	Q_{Gate}		15	960	160	$T_j=25^\circ\text{C}$		740		nC
Thermal resistance chip to heatsink per chip	R_{thJH}	Thermal grease thickness \leq 50um						0,37		K/W
Thermal resistance chip to case per chip	R_{thJC}	$\lambda = 1 \text{ W/mK}$						0,24		
*additional value stands for built-in capacitor										
NP Diode (D7, D8)										
Diode forward voltage	V_F				120	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$	1,4	1,47 1,29	2	V
Peak reverse recovery current	I_{RRM}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		127 151		A
Reverse recovery time	t_{rr}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		40 81		ns
Reverse recovered charge	Q_{rr}	$R_{gon}=4 \Omega$	± 15	350	100	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		3,02 7,13		μC
Peak rate of fall of recovery current	$di(\text{rec})_{\text{max}}/dt$					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		12386 3767		A/ μs
Reverse recovered energy	E_{rec}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		0,31 1,01		mWs
Thermal resistance chip to heatsink per chip	R_{thJH}	Thermal grease thickness \leq 50um						1,05		K/W
Thermal resistance chip to case per chip	R_{thJC}	$\lambda = 1 \text{ W/mK}$						0,69		

Characteristic Values

Parameter	Symbol	Conditions					Value			Unit
		V_{GE} [V] or V_{GS} [V]	V_r [V] or V_{CE} [V] or V_{OS} [V]	I_c [A] or I_f [A] or I_b [A]	T_j	Min	Typ	Max		
NP IGBT (T2, T3)										
Gate emitter threshold voltage	$V_{GE(th)}$	$V_{CE}=V_{GE}$			0,0016	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$	5	5,8	6,5	V
Collector-emitter saturation voltage	$V_{CE(sat)}$		15		100	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$	1,05	1,58 1,8	1,85	V
Collector-emitter cut-off incl diode	I_{CES}		0	600		$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$			0,0052	mA
Gate-emitter leakage current	I_{GES}		20	0		$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$			1200	nA
Integrated Gate resistor	R_{gint}							none		Ω
Turn-on delay time	$t_{d(on)}$					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		103 103		ns
Rise time	t_r					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		16,8 19,2		
Turn-off delay time	$t_{d(off)}$	$R_{goff}=4\ \Omega$ $R_{gon}=4\ \Omega$	± 15	350	100	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		158 179		
Fall time	t_f					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		44 64		
Turn-on energy loss per pulse	E_{on}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		1,06 1,52		
Turn-off energy loss per pulse	E_{off}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		2,48 3,32		
Input capacitance	C_{ies}							6280		pF
Output capacitance	C_{oss}	$f=1\text{MHz}$	15	480	100	$T_j=25^\circ\text{C}$		400		
Reverse transfer capacitance	C_{rss}							186		
Gate charge	Q_{Gate}					$T_j=25^\circ\text{C}$		620		nC
Thermal resistance chip to heatsink per chip	$R_{th,JH}$	Thermal grease thickness $\leq 50\mu\text{m}$						1,01		K/W
Thermal resistance chip to case per chip	$R_{th,JC}$	$\lambda = 1\ \text{W/mK}$						0,67		
NP Inverse Diode (D2, D3)										
Diode forward voltage	V_F				15	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$	1,00	1,61 1,57	2,15	V
Thermal resistance chip to heatsink per chip	$R_{th,JH}$	Thermal grease thickness $\leq 50\mu\text{m}$						3,43		K/W
Coupled thermal resistance inverter transistor-diode	$R_{th,JC}$	$\lambda = 1\ \text{W/mK}$						2,27		
Halfbridge Diode (D5, D6)										
Diode forward voltage	V_F				60	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$	1,50	2,47 2,11	3,40	V
Reverse leakage current	I_r			1200		$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$			200	μA
Peak reverse recovery current	I_{RRM}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		107 142		A
Reverse recovery time	t_{rr}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		51 69		ns
Reverse recovered charge	Q_{rr}	$R_{gon}=4\ \Omega$	± 15	350	100	$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		6,24 12,71		μC
Peak rate of fall of recovery current	$di(\text{rec})_{\text{max}}/dt$					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		5985 2890		A/ μs
Reverse recovery energy	E_{rec}					$T_j=25^\circ\text{C}$ $T_j=125^\circ\text{C}$		1,71 3,61		mWs
Thermal resistance chip to heatsink per chip	$R_{th,JH}$	Thermal grease thickness $\leq 50\mu\text{m}$						1,15		K/W
Thermal resistance chip to case per chip	$R_{th,JC}$	$\lambda = 1\ \text{W/mK}$						0,76		
DC link Capacitor (C1, C2)										
C value	C	DC+ to Neutral and DC- to Neutral						100		nF
Thermistor										
Rated resistance	R					$T=25^\circ\text{C}$		22000		Ω
Deviation of R25	$\Delta R/R$	$R_{100}=1486\ \Omega$				$T=25^\circ\text{C}$	-5		+5	%
Power dissipation	P					$T=25^\circ\text{C}$		200		mW
Power dissipation constant						$T_j=25^\circ\text{C}$		2		mW/K
B-value	$B_{(25/50)}$	Tol. $\pm 3\%$				$T_j=25^\circ\text{C}$		3950		K
B-value	$B_{(25/100)}$	Tol. $\pm 3\%$				$T_j=25^\circ\text{C}$		3996		K
Vincotech NTC Reference									B	
Module Properties										
Thermal resistance, case to heatsink	$R_{th,CH}$	per module $\lambda_{\text{Paste}}=1\text{W}/(\text{m}\cdot\text{K})/\lambda_{\text{grease}}=1\text{W}/(\text{m}\cdot\text{K})$						tbd.		K/W
Module stray inductance	L_{sCE}	V23990-P-M107- \cdot -31						5		nH
Chip module lead resistance, terminals -chip	R_{CC1+EE}	$T_C=25^\circ\text{C}$, per switch						tbd.		m Ω
Mounting torque	M	Screw M4 - mounting according to valid application note Flow1-4TY-P \cdot -HI for PressFIT, V23990-P-M101- \cdot -31 for SolderPin						2	2,2	Nm
Weight	G							42,28		g

Half bridge (T1, T4 / D7, D8)

half bridge IGBT and Neutral Point FWD

Figure 1 IGBT

Typical output characteristics

$I_C = f(V_{CE})$

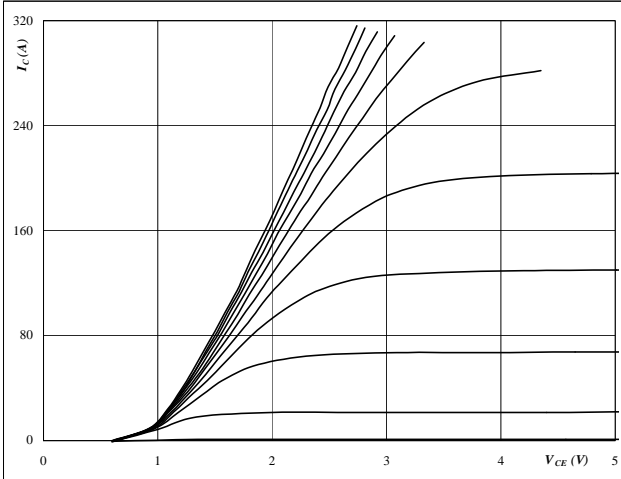

At
 $t_p = 250 \mu s$
 $T_j = 25 \text{ } ^\circ C$
 V_{GE} from 7 V to 17 V in steps of 1 V

Figure 2 IGBT

Typical output characteristics

$I_C = f(V_{CE})$

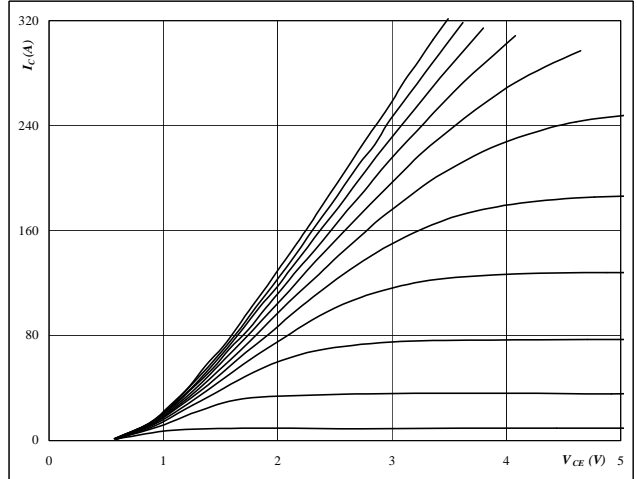
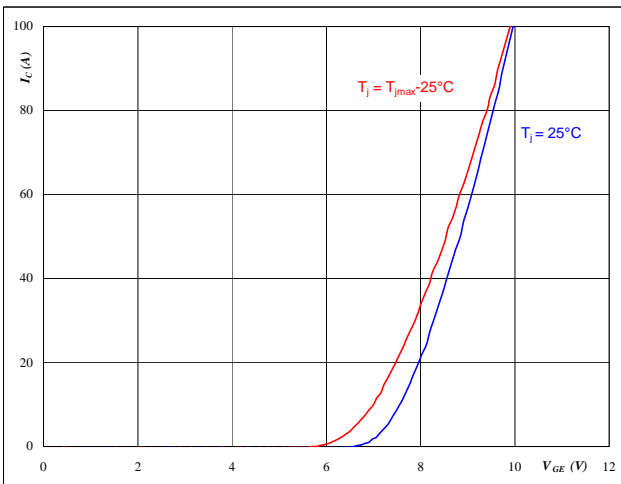

At
 $t_p = 250 \mu s$
 $T_j = 125 \text{ } ^\circ C$
 V_{GE} from 7 V to 17 V in steps of 1 V

Figure 3 IGBT

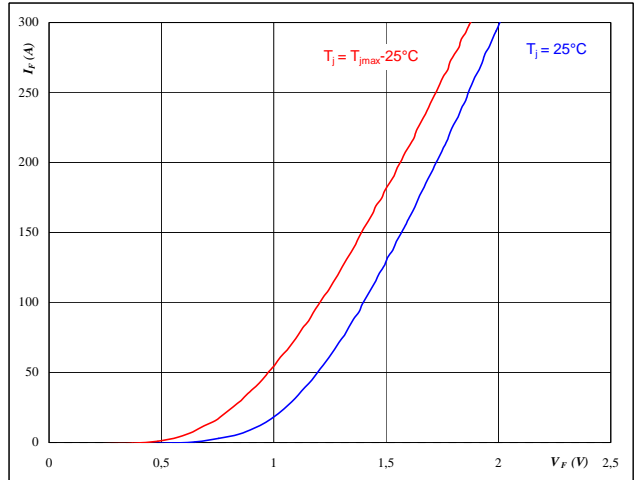
Typical transfer characteristics

$I_C = f(V_{GE})$


At
 $t_p = 250 \mu s$
 $V_{CE} = 10 \text{ V}$
 $T_j = 25/150 \text{ } ^\circ C$
Figure 4 NP FWD

Typical FWD forward current as a function of forward voltage

$I_F = f(V_F)$


At
 $t_p = 250 \mu s$
 $T_j = 25/150 \text{ } ^\circ C$

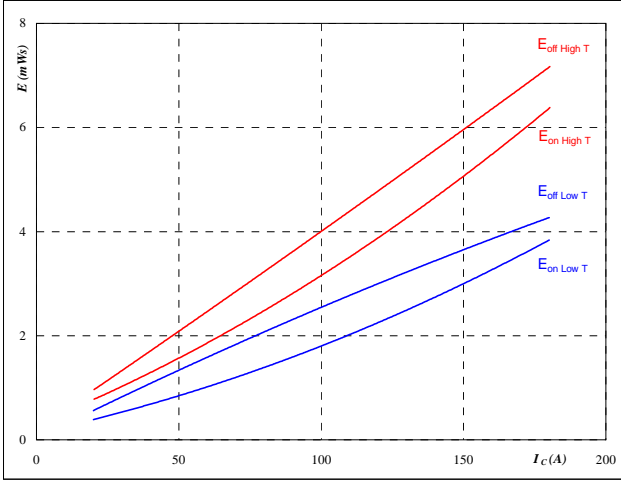
Half bridge (T1, T4 / D7, D8)

half bridge IGBT and Neutral Point FWD

Figure 5 IGBT

Typical switching energy losses
as a function of collector current

$$E = f(I_C)$$



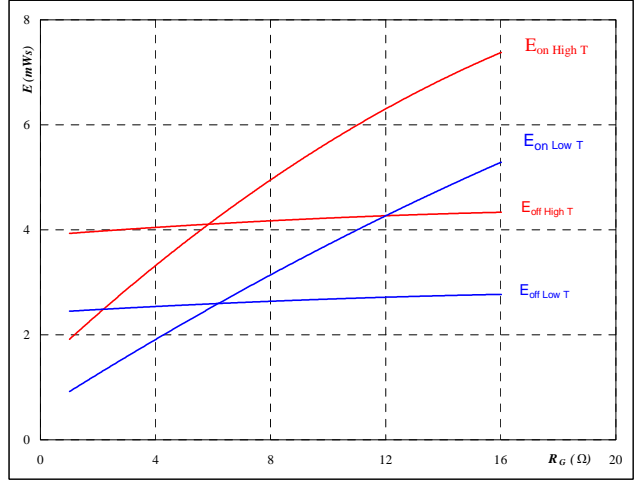
With an inductive load at

$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	4	Ω
$R_{goff} =$	4	Ω

Figure 6 IGBT

Typical switching energy losses
as a function of gate resistor

$$E = f(R_G)$$



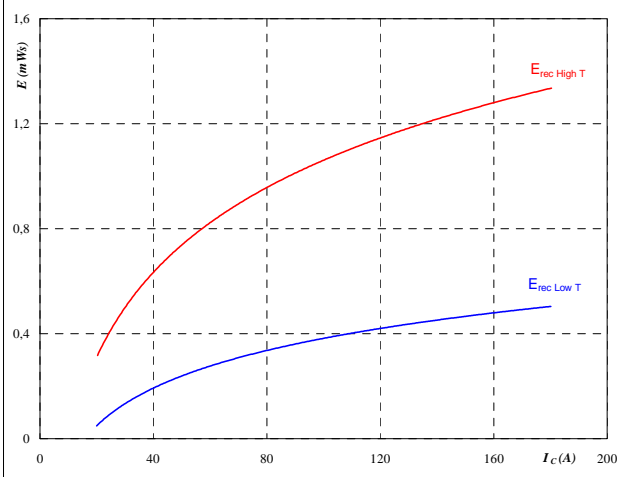
With an inductive load at

$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$I_C =$	100	A

Figure 7 NP FWD

Typical reverse recovery energy loss
as a function of collector current

$$E_{rec} = f(I_C)$$



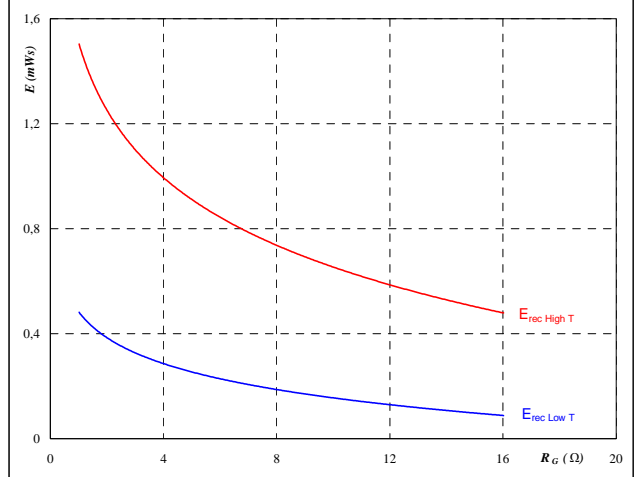
With an inductive load at

$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	4	Ω

Figure 8 NP FWD

Typical reverse recovery energy loss
as a function of gate resistor

$$E_{rec} = f(R_G)$$



With an inductive load at

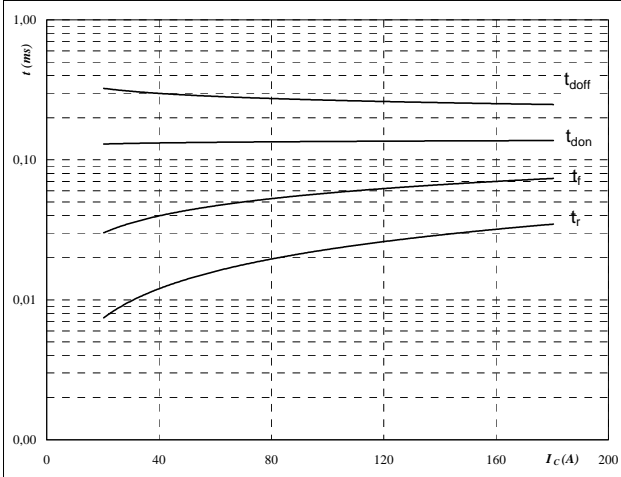
$T_J =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$I_C =$	100	A

Half bridge (T1, T4 / D7, D8)

half bridge IGBT and Neutral Point FWD

Figure 9 IGBT

Typical switching times as a function of collector current
 $t = f(I_C)$

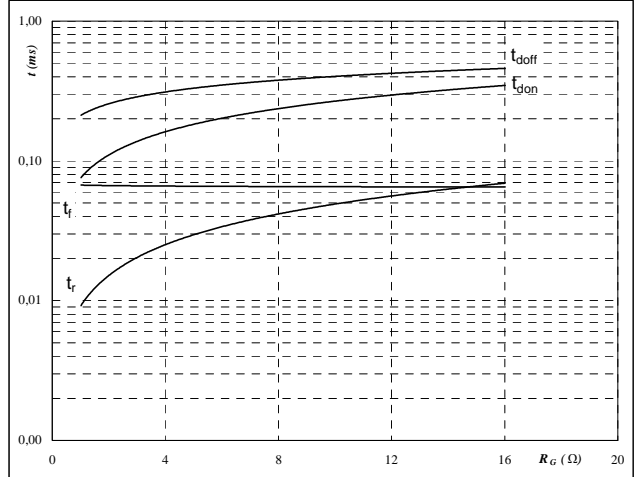


With an inductive load at

$T_J = 125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 4 \text{ } \Omega$
 $R_{goff} = 4 \text{ } \Omega$

Figure 10 IGBT

Typical switching times as a function of gate resistor
 $t = f(R_G)$

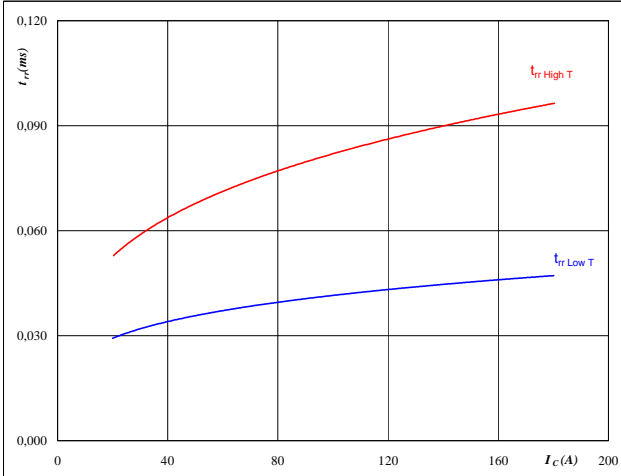


With an inductive load at

$T_J = 125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $I_C = 100 \text{ A}$

Figure 11 NP FWD

Typical reverse recovery time as a function of collector current
 $t_{rr} = f(I_C)$

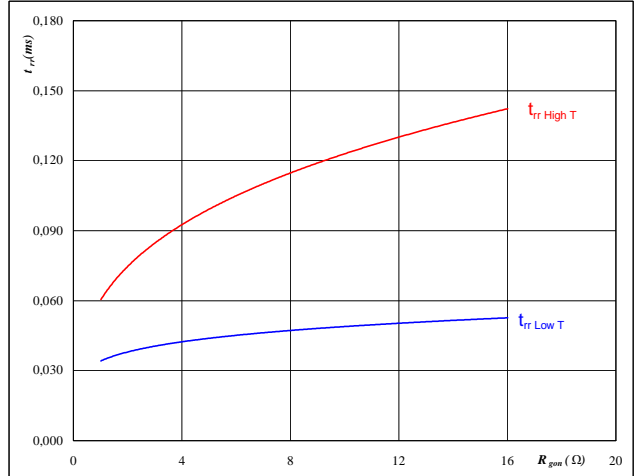


At

$T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 4 \text{ } \Omega$

Figure 12 NP FWD

Typical reverse recovery time as a function of IGBT turn on gate resistor
 $t_{rr} = f(R_{gon})$



At

$T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 100 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$

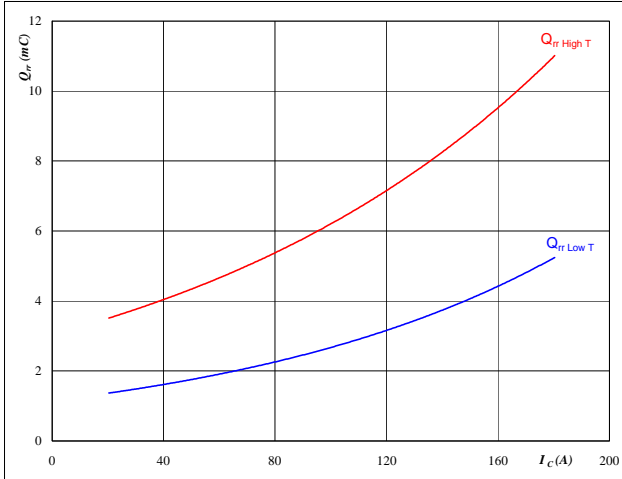
Half bridge (T1, T4 / D7, D8)

half bridge IGBT and Neutral Point FWD

Figure 13 NP FWD

Typical reverse recovery charge as a function of collector current

$$Q_{rr} = f(I_c)$$



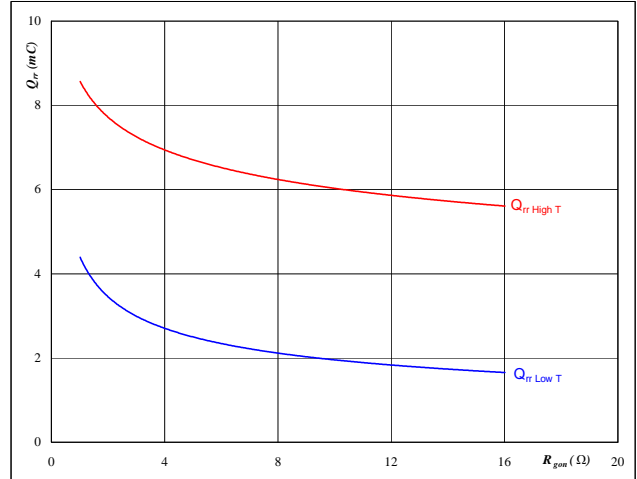
At

$T_j =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	4	Ω

Figure 14 NP FWD

Typical reverse recovery charge as a function of JFET turn on gate resistor

$$Q_{rr} = f(R_{gon})$$



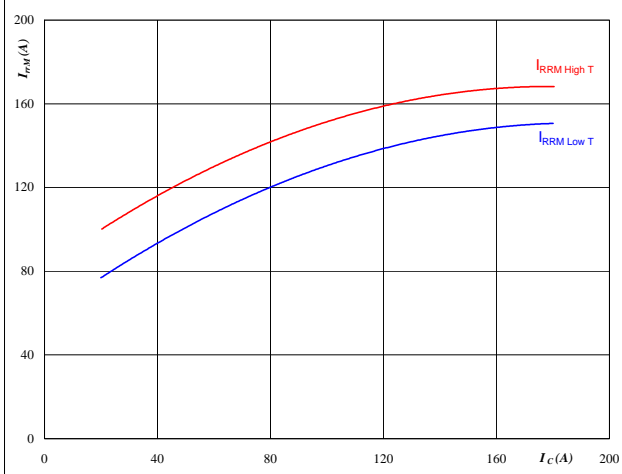
At

$T_j =$	25/125	°C
$V_R =$	350	V
$I_F =$	100	A
$V_{GE} =$	±15	V

Figure 15 NP FWD

Typical reverse recovery current as a function of collector current

$$I_{RRM} = f(I_c)$$



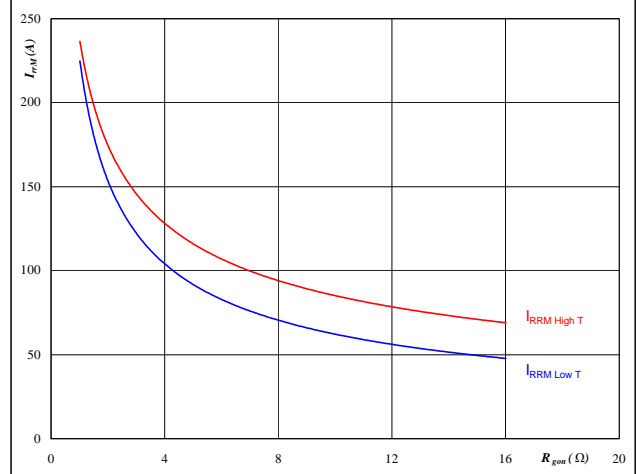
At

$T_j =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	4	Ω

Figure 16 NP FWD

Typical reverse recovery current as a function of JFET turn on gate resistor

$$I_{RRM} = f(R_{gon})$$



At

$T_j =$	25/125	°C
$V_R =$	350	V
$I_F =$	100	A
$V_{GE} =$	±15	V

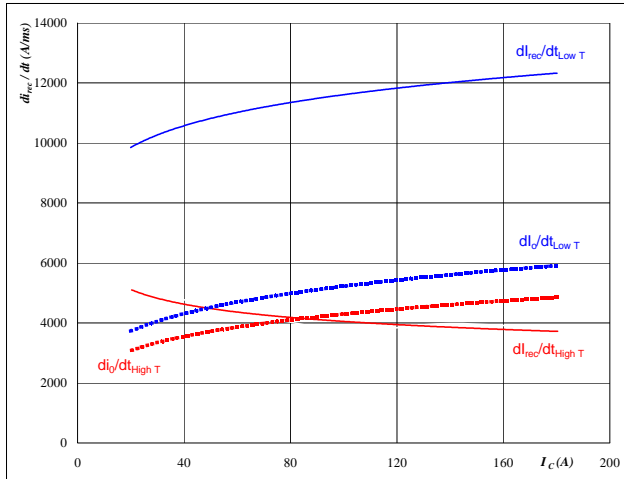
Half bridge (T1, T4 / D7, D8)

half bridge IGBT and Neutral Point FWD

Figure 17 NP FWD

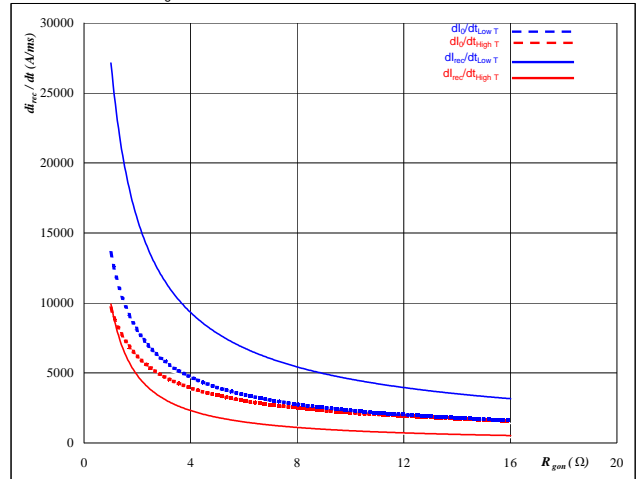
Typical rate of fall of forward and reverse recovery current as a function of collector current

$$di_o/dt, di_{rec}/dt = f(I_c)$$


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 4 \text{ } \Omega$
Figure 18 NP FWD

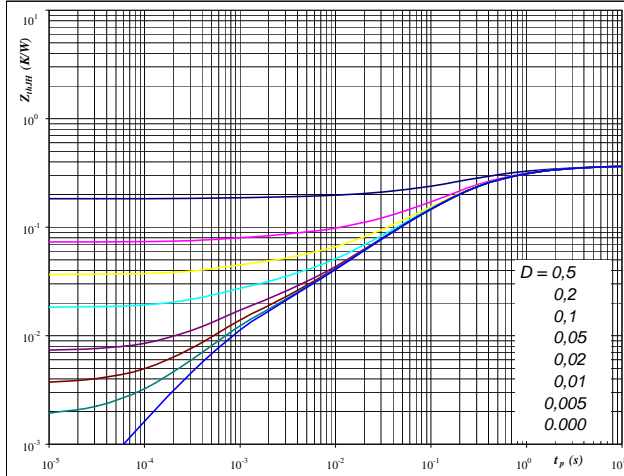
Typical rate of fall of forward and reverse recovery current as a function of JFET turn on gate resistor

$$di_o/dt, di_{rec}/dt = f(R_{gon})$$


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 100 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$
Figure 19 IGBT

IGBT transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$


At
 $D = t_p / T$
 $R_{thJH} = 0,37 \text{ K/W}$

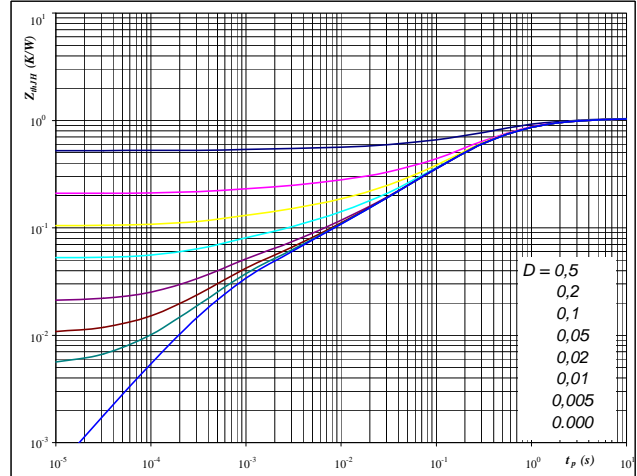
IGBT thermal model values

R (C/W)	Tau (s)
0,06	2,4E+00
0,15	4,0E-01
0,12	1,0E-01
0,03	1,3E-02
0,01	8,4E-04

Figure 20 NP FWD

FWD transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$


At
 $D = t_p / T$
 $R_{thJH} = 1,05 \text{ K/W}$

FWD thermal model values

R (C/W)	Tau (s)
0,05	7,4E+00
0,27	1,3E+00
0,55	2,7E-01
0,11	4,0E-02
0,04	5,1E-03
0,03	6,0E-04

Half bridge (T1, T4 / D7, D8)

half bridge IGBT and Neutral Point FWD

Figure 21 IGBT

Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$


At
 $T_j = 175$ °C

Figure 22 IGBT

Collector current as a function of heatsink temperature

$$I_C = f(T_h)$$

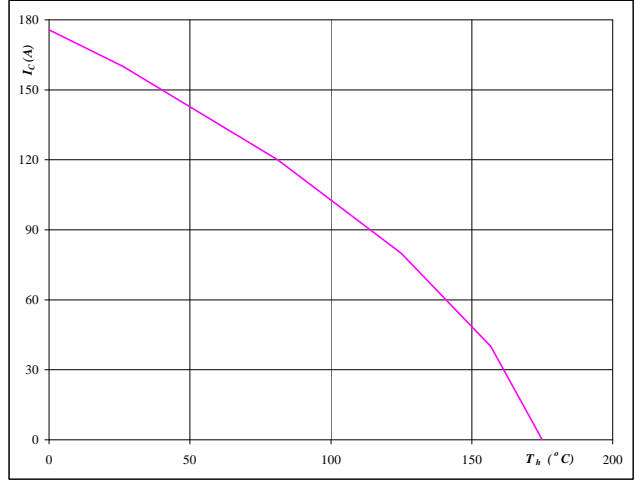

At
 $T_j = 175$ °C
 $V_{GE} = 15$ V

Figure 23 NP FWD

Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$

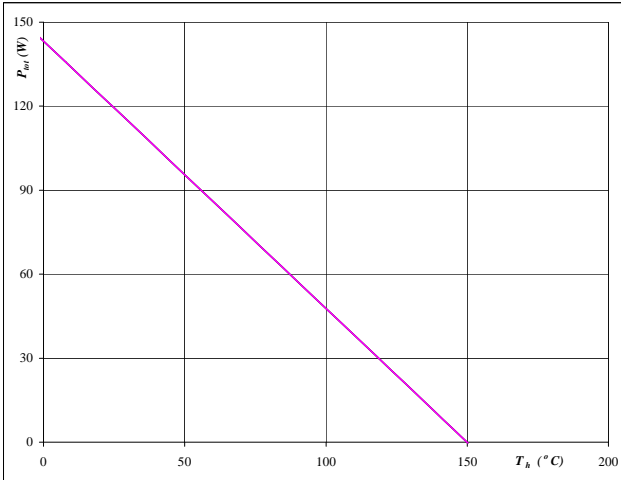
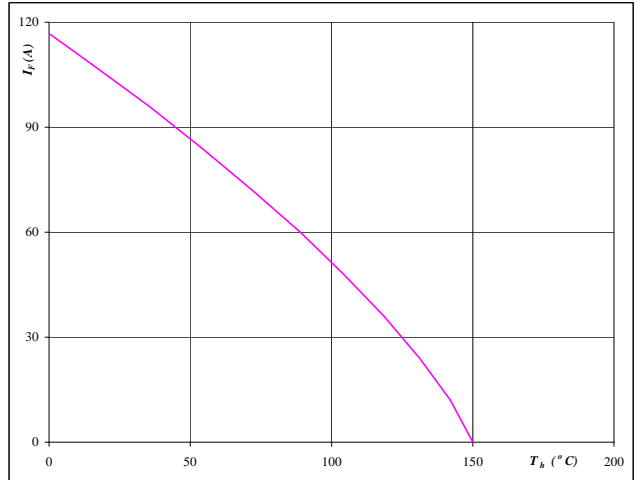

At
 $T_j = 150$ °C

Figure 24 NP FWD

Forward current as a function of heatsink temperature

$$I_F = f(T_h)$$


At
 $T_j = 150$ °C

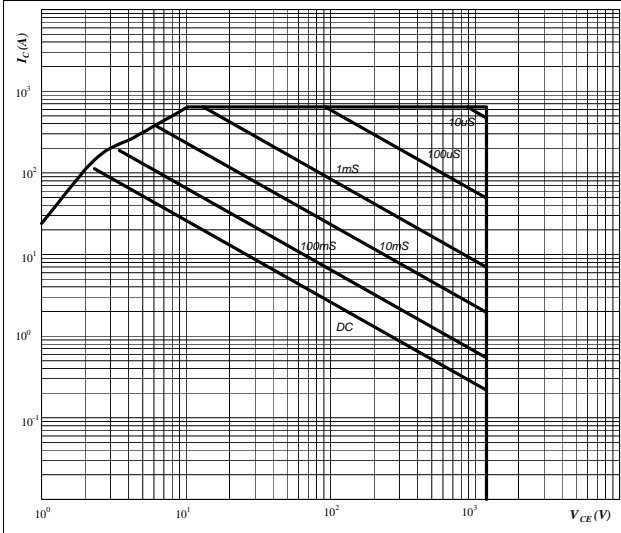
Half bridge (T1, T4 / D7, D8)

half bridge IGBT and Neutral Point FWD

Figure 25 IGBT

Safe operating area as a function of collector-emitter voltage

$$I_C = f(V_{CE})$$

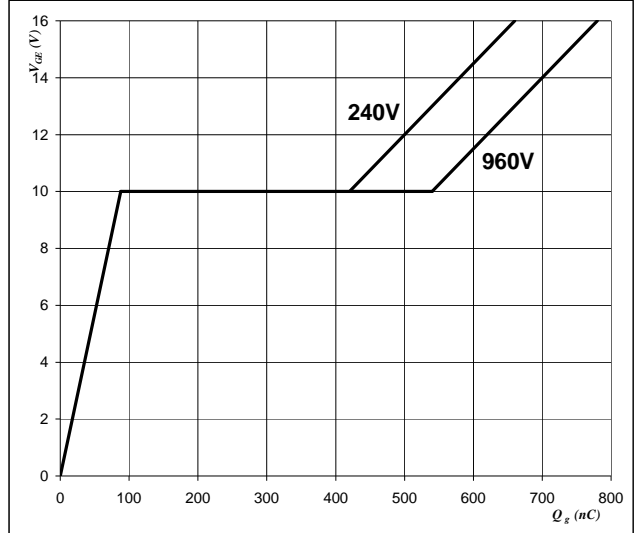


At
 D = single pulse
 Th = 80 °C
 V_{GE} = 0 V
 T_j = T_{jmax} °C

Figure 26 IGBT

Gate voltage vs Gate charge

$$V_{GE} = f(Q_g)$$

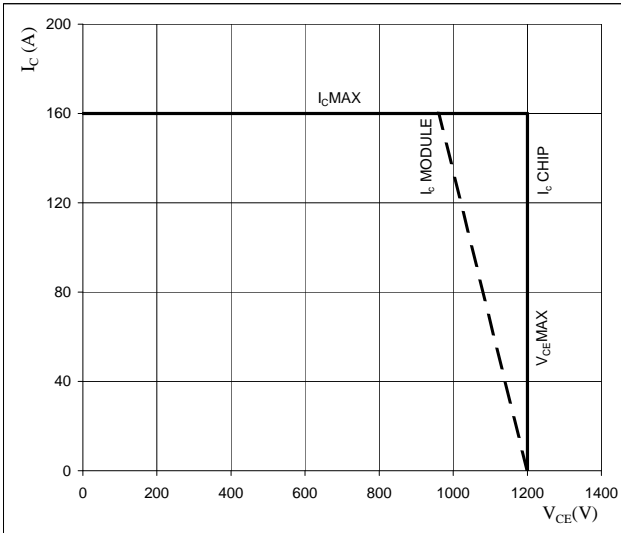


At
 I_D = 20 A
 V_{DS} = 600 V
 T_j = 25 °C

Figure 27 IGBT

Reverse bias safe operating area

$$I_C = f(V_{CE})$$



At
 T_j = T_{jmax}-25 °C
 U_{ocminus} = U_{ccplus}
 Switching mode : 3 level switching

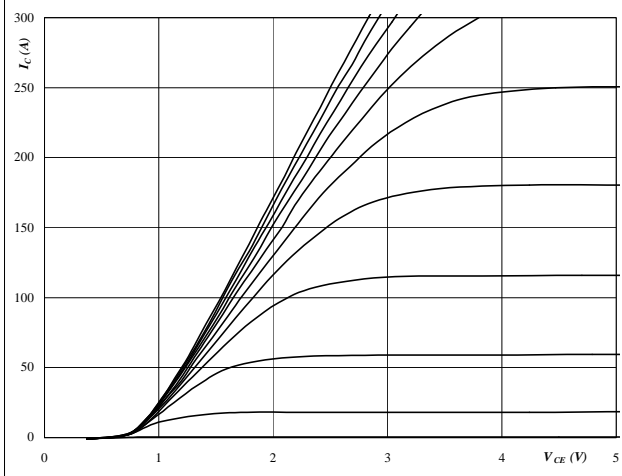
Neutral Point IGBT (T2, T3 / D5, D6)

neutral point IGBT and half bridge FWD

Figure 1 NP IGBT

Typical output characteristics

$I_C = f(V_{CE})$

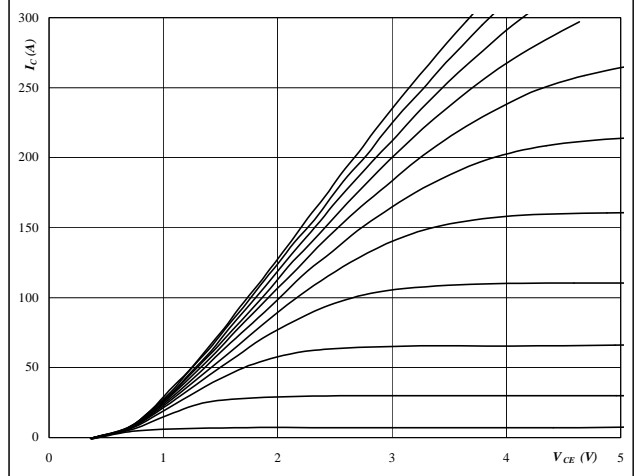


At
 $t_p = 250 \mu s$
 $T_j = 25 \text{ } ^\circ C$
 V_{GE} from 7 V to 17 V in steps of 1 V

Figure 2 NP IGBT

Typical output characteristics

$I_C = f(V_{CE})$

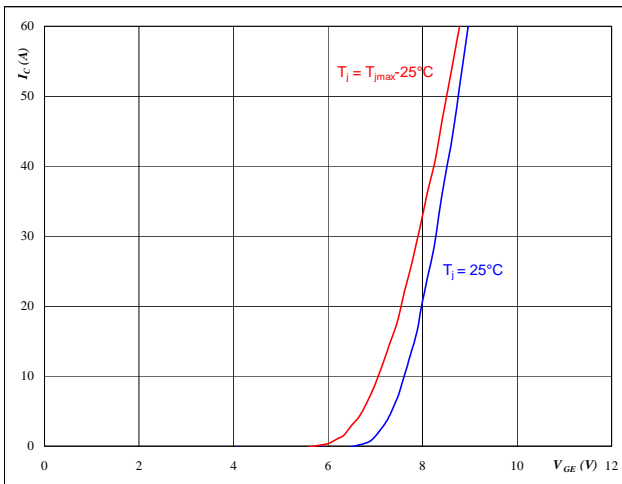


At
 $t_p = 250 \mu s$
 $T_j = 150 \text{ } ^\circ C$
 V_{GE} from 7 V to 17 V in steps of 1 V

Figure 3 NP IGBT

Typical transfer characteristics

$I_C = f(V_{GE})$

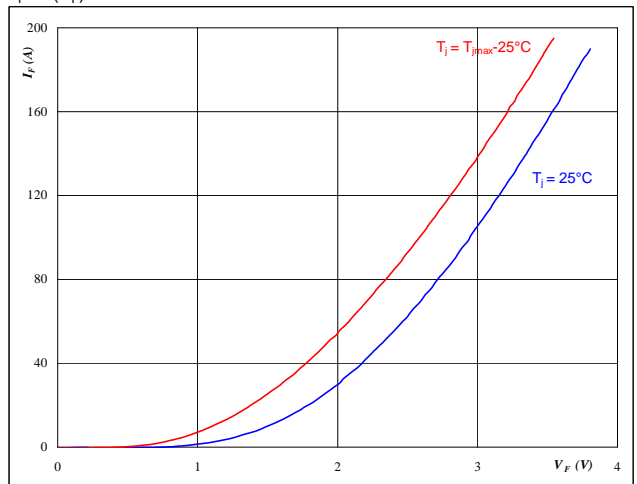


At
 $t_p = 250 \mu s$
 $V_{CE} = 10 \text{ V}$
 $T_j = 25/150 \text{ } ^\circ C$

Figure 4 FWD

Typical FWD forward current as a function of forward voltage

$I_F = f(V_F)$



At
 $t_p = 250 \mu s$
 $T_j = 25/150 \text{ } ^\circ C$

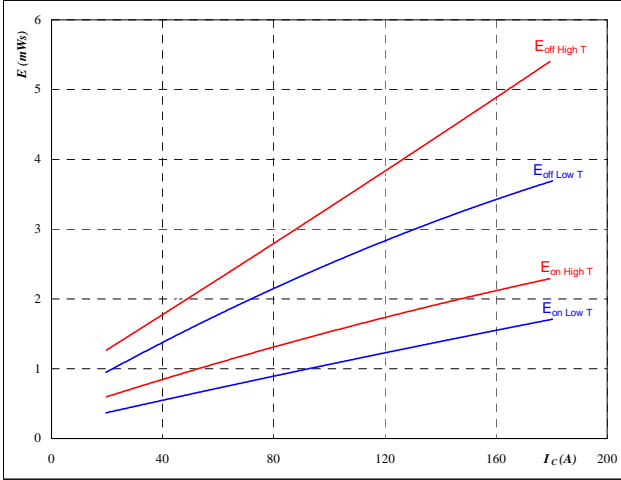
Neutral Point IGBT (T2, T3 / D5, D6)

neutral point IGBT and half bridge FWD

Figure 5 NP IGBT

Typical switching energy losses
as a function of collector current

$$E = f(I_C)$$

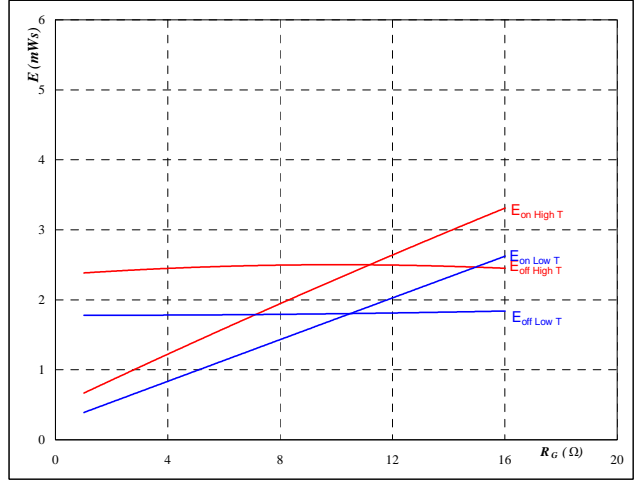


With an inductive load at

 $T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 4 \text{ } \Omega$
 $R_{goff} = 4 \text{ } \Omega$
Figure 6 NP IGBT

Typical switching energy losses
as a function of gate resistor

$$E = f(R_G)$$

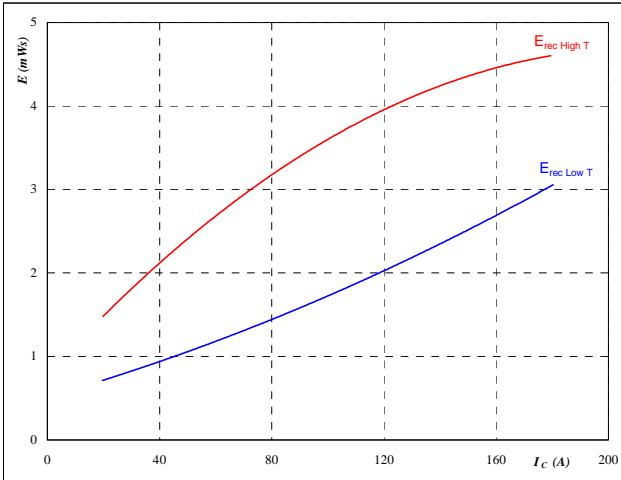


With an inductive load at

 $T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $I_C = 60 \text{ A}$
Figure 7 FWD

Typical reverse recovery energy loss
as a function of collector current

$$E_{rec} = f(I_C)$$

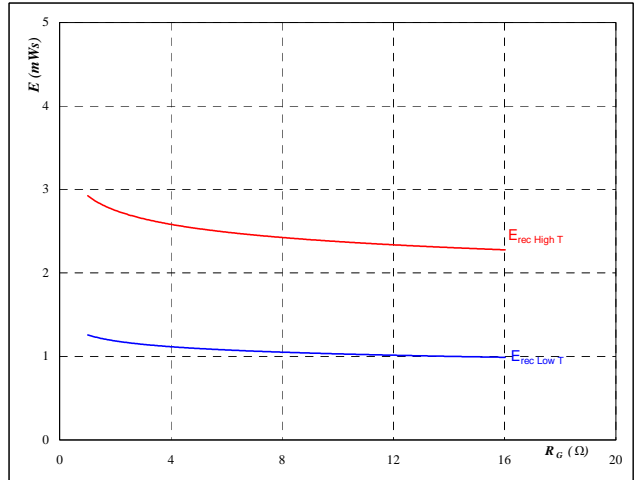


With an inductive load at

 $T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 4 \text{ } \Omega$
Figure 8 FWD

Typical reverse recovery energy loss
as a function of gate resistor

$$E_{rec} = f(R_G)$$



With an inductive load at

 $T_J = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $I_C = 60 \text{ A}$

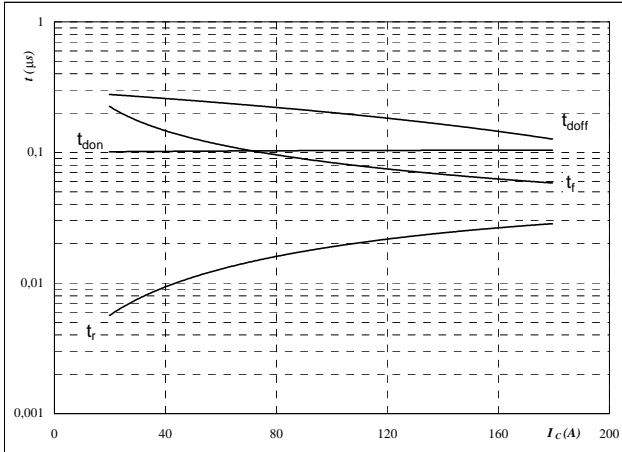
Neutral Point IGBT (T2, T3 / D5, D6)

neutral point IGBT and half bridge FWD

Figure 9 NP IGBT

Typical switching times as a function of collector current

$$t = f(I_c)$$



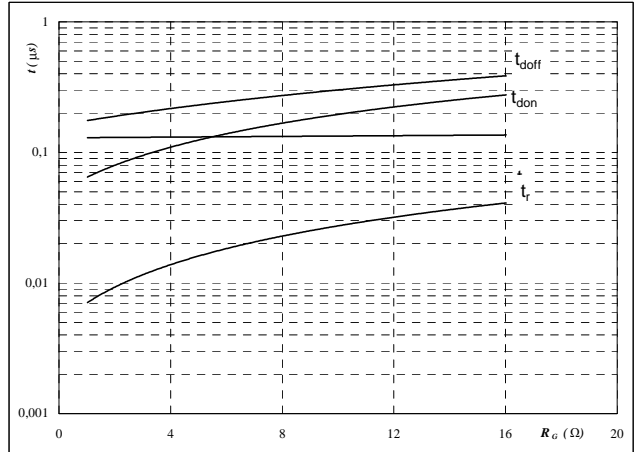
With an inductive load at

$T_j =$	125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	4	Ω
$R_{goff} =$	4	Ω

Figure 10 NP IGBT

Typical switching times as a function of gate resistor

$$t = f(R_G)$$



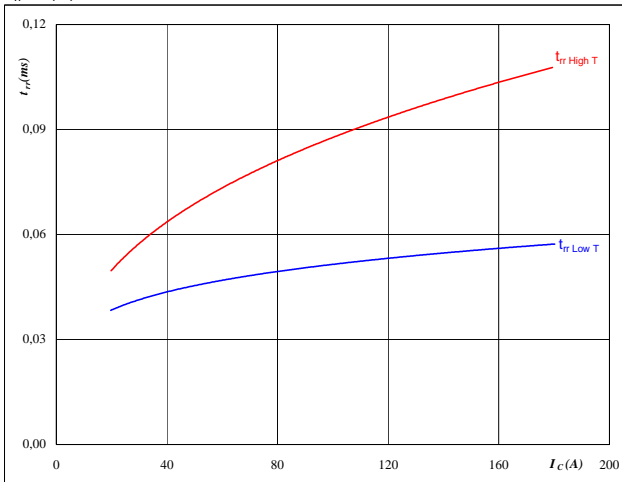
With an inductive load at

$T_j =$	125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$I_C =$	60	A

Figure 11 FWD

Typical reverse recovery time as a function of collector current

$$t_{rr} = f(I_c)$$

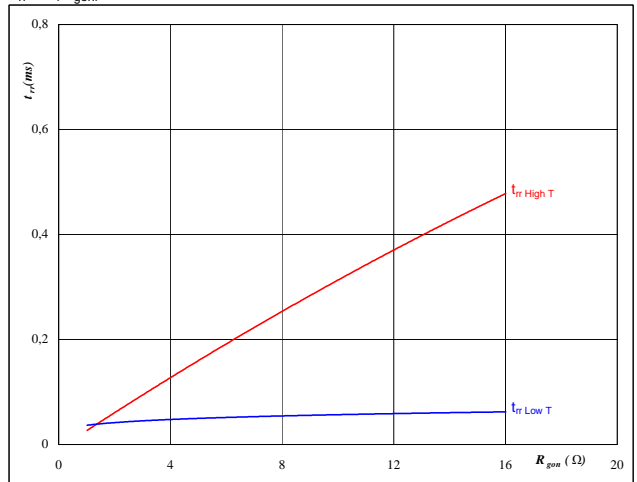

At

$T_j =$	25/125	°C
$V_{CE} =$	350	V
$V_{GE} =$	±15	V
$R_{gon} =$	4,0	Ω

Figure 12 FWD

Typical reverse recovery time as a function of IGBT turn on gate resistor

$$t_{rr} = f(R_{gon})$$


At

$T_j =$	25/125	°C
$V_R =$	350	V
$I_F =$	60	A
$V_{GE} =$	±15	V

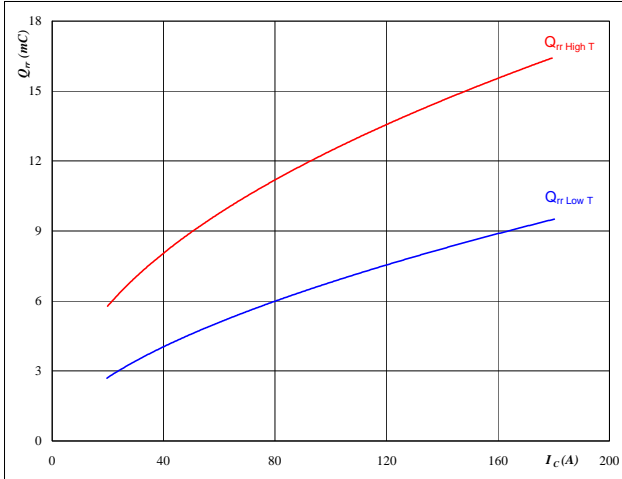
Neutral Point IGBT (T2, T3 / D5, D6)

neutral point IGBT and half bridge FWD

Figure 13 FWD

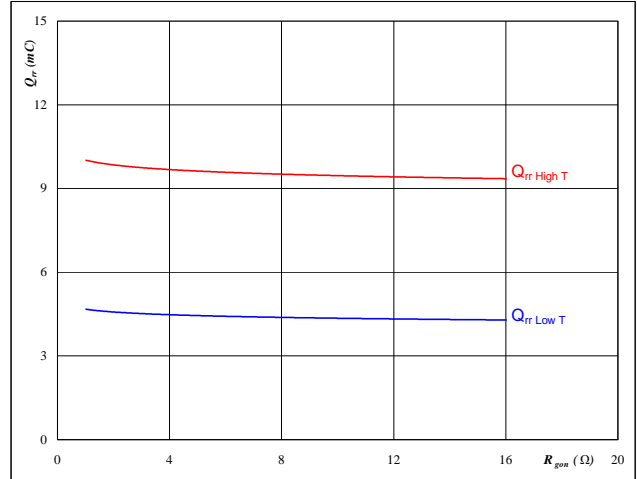
Typical reverse recovery charge as a function of collector current

$$Q_{rr} = f(I_C)$$


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 4,0 \text{ } \Omega$
Figure 14 FWD

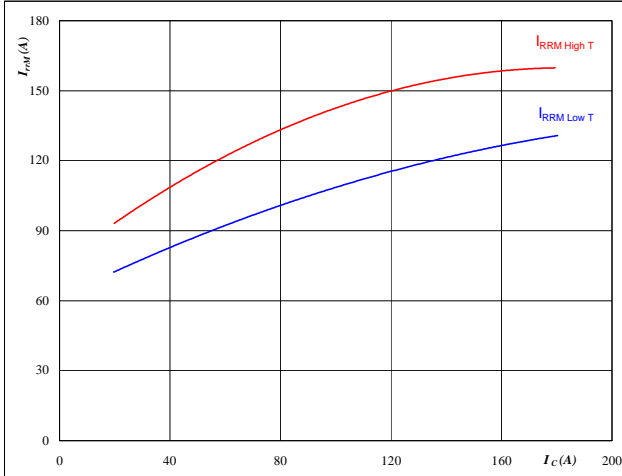
Typical reverse recovery charge as a function of IGBT turn on gate resistor

$$Q_{rr} = f(R_{gon})$$


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 60 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$
Figure 15 FWD

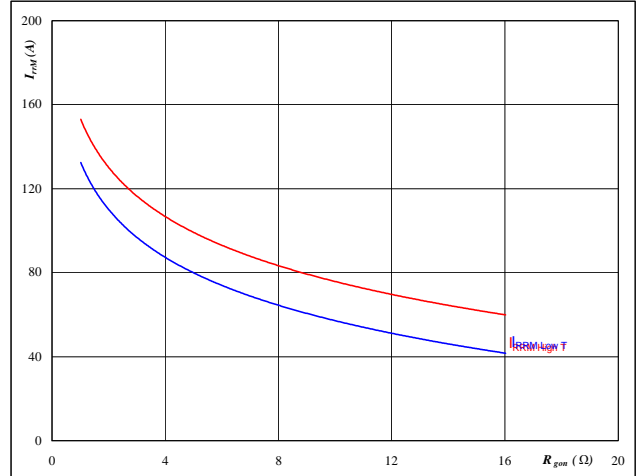
Typical reverse recovery current as a function of collector current

$$I_{RRM} = f(I_C)$$


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 4,0 \text{ } \Omega$
Figure 16 FWD

Typical reverse recovery current as a function of IGBT turn on gate resistor

$$I_{RRM} = f(R_{gon})$$

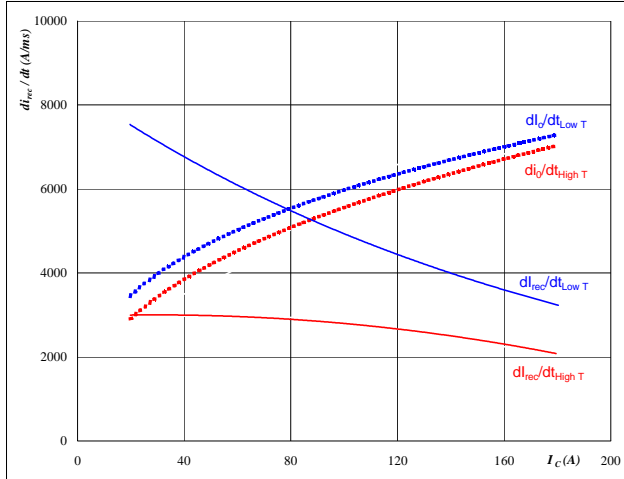

At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 60 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$

Neutral Point IGBT (T2, T3 / D5, D6)

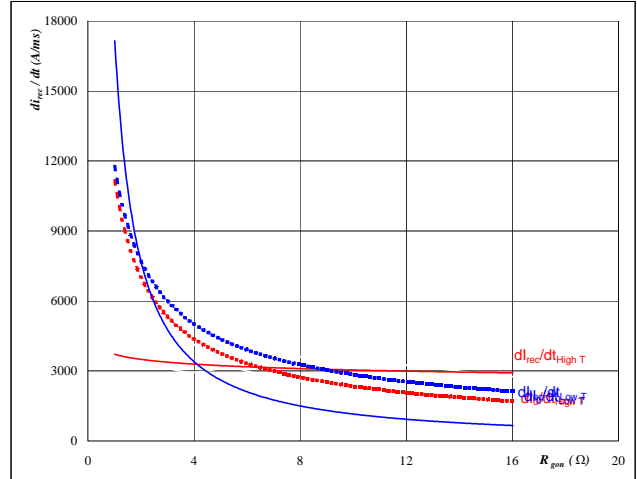
neutral point IGBT and half bridge FWD

Figure 17 FWD
Typical rate of fall of forward and reverse recovery current as a function of collector current

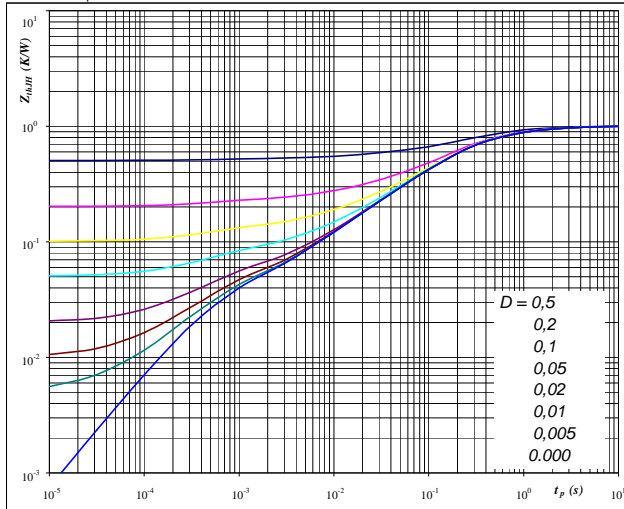
$$di_o/dt, di_{rec}/dt = f(I_c)$$


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_{CE} = 350 \text{ V}$
 $V_{GE} = \pm 15 \text{ V}$
 $R_{gon} = 4,0 \text{ } \Omega$
Figure 18 FWD
Typical rate of fall of forward and reverse recovery current as a function of IGBT turn on gate resistor

$$di_o/dt, di_{rec}/dt = f(R_{gon})$$


At
 $T_j = 25/125 \text{ } ^\circ\text{C}$
 $V_R = 350 \text{ V}$
 $I_F = 60 \text{ A}$
 $V_{GE} = \pm 15 \text{ V}$
Figure 19 NP IGBT
IGBT transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$

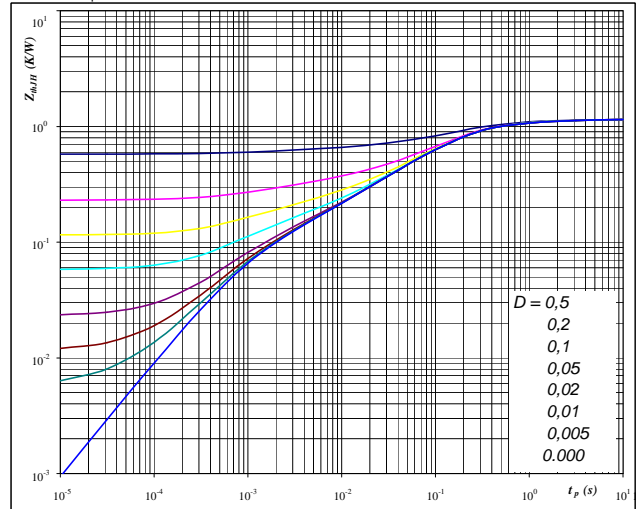

At
 $D = t_p / T$
 $R_{thJH} = 1,01 \text{ K/W}$

IGBT thermal model values

R (C/W)	Tau (s)
0,05	6,49
0,16	1,27
0,52	0,25
0,18	0,07
0,07	0,01

Figure 20 FWD
FWD transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$


At
 $D = t_p / T$
 $R_{thJH} = 1,15 \text{ K/W}$

FWD thermal model values

R (C/W)	Tau (s)
0,05	4,90
0,13	0,82
0,59	0,18
0,22	0,05
0,10	0,01

Neutral Point IGBT (T2, T3 / D5, D6)

neutral point IGBT and half bridge FWD

Figure 21 NP IGBT

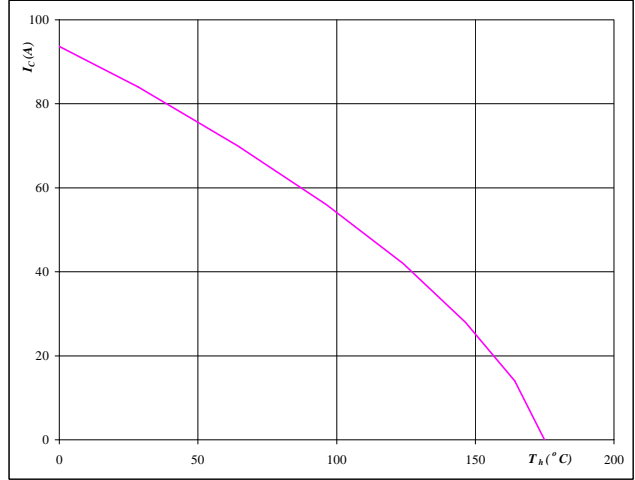
Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$


At
 $T_j = 175 \text{ } ^\circ\text{C}$
Figure 22 NP IGBT

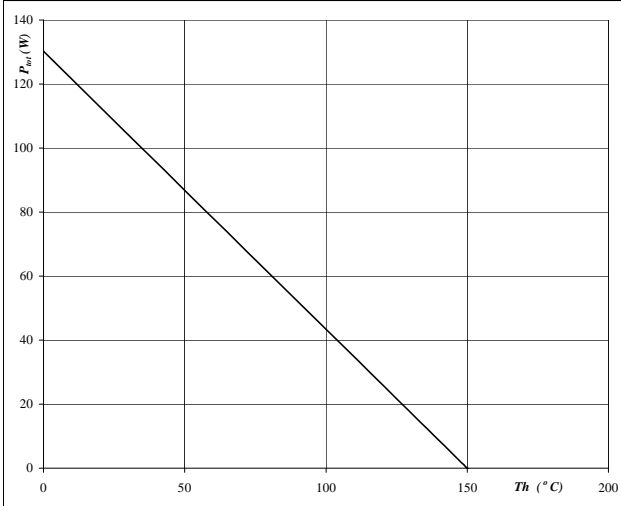
Collector current as a function of heatsink temperature

$$I_C = f(T_h)$$


At
 $T_j = 175 \text{ } ^\circ\text{C}$
 $V_{GE} = 15 \text{ V}$
Figure 23 FWD

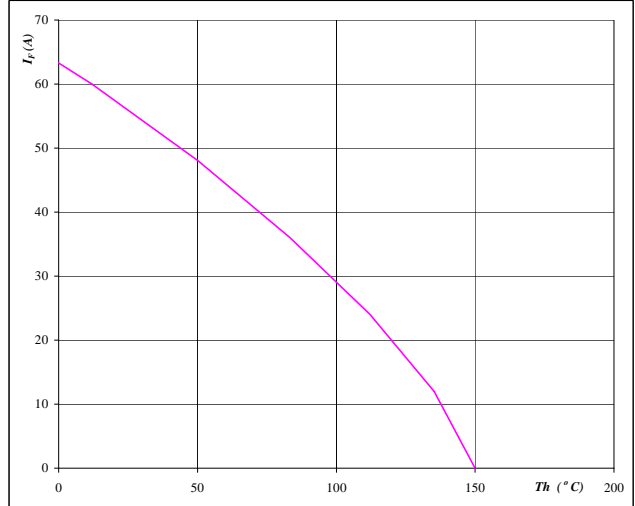
Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$


At
 $T_j = 150 \text{ } ^\circ\text{C}$
Figure 24 FWD

Forward current as a function of heatsink temperature

$$I_F = f(T_h)$$

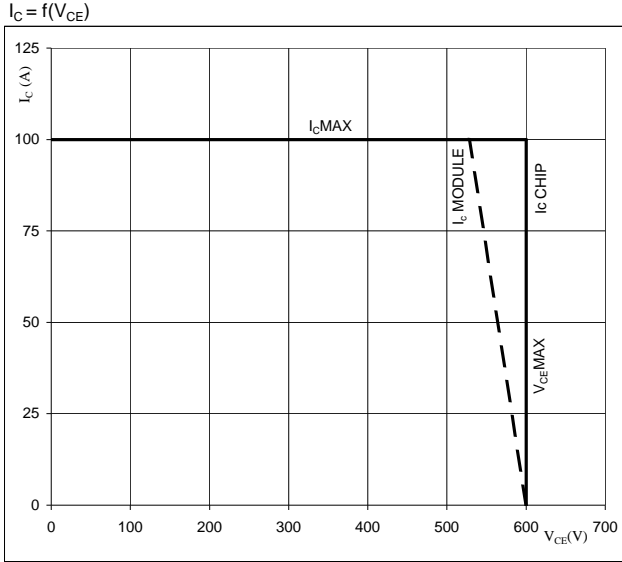

At
 $T_j = 150 \text{ } ^\circ\text{C}$

Neutral Point IGBT (T2, T3 / D5, D6)

neutral point IGBT

Figure 25 NP IGBT

Reverse bias safe operating area



At

$$T_J = T_{j\text{max}} - 25 \quad ^\circ\text{C}$$

$$U_{oc\text{minus}} = U_{cc\text{plus}}$$

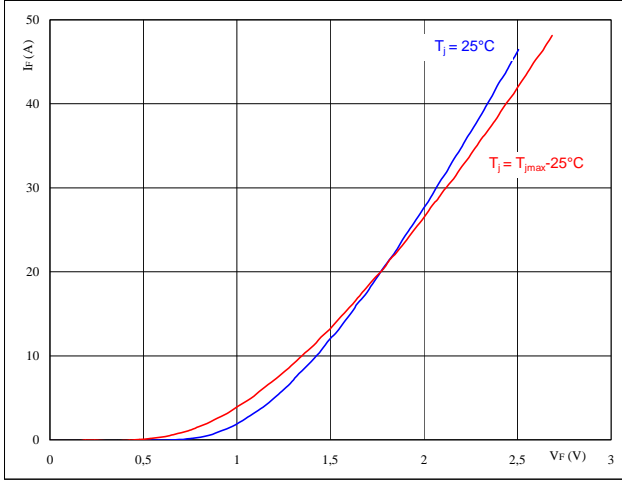
Switching mode : 3 level switching

NP IGBT Inverse Diode (D2, D3)

Figure 25 NP Inverse Diode

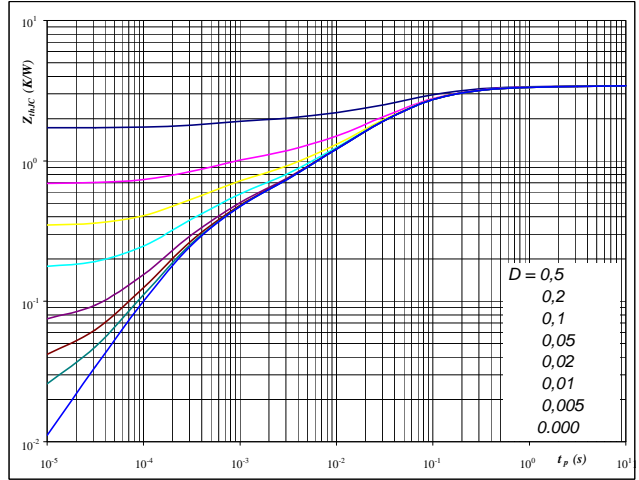
Typical FWD forward current as a function of forward voltage

$$I_F = f(V_F)$$


At
 $t_p = 250 \mu s$
Figure 26 NP Inverse Diode

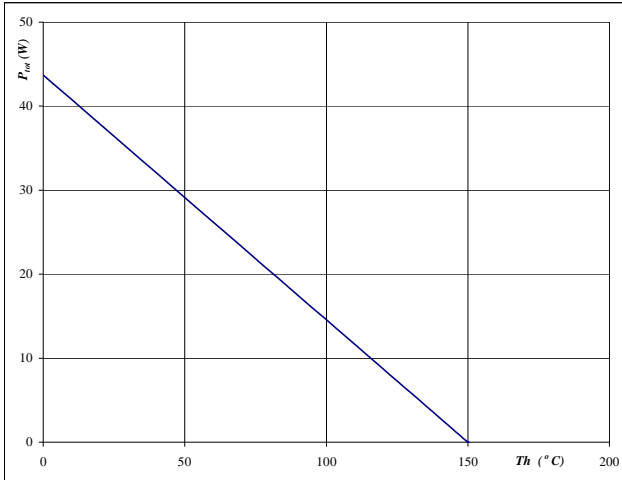
FWD transient thermal impedance as a function of pulse width

$$Z_{thJH} = f(t_p)$$


At
 $D = t_p / T$
 $R_{thJH} = 3,43 \text{ K/W}$
Figure 27 NP Inverse Diode

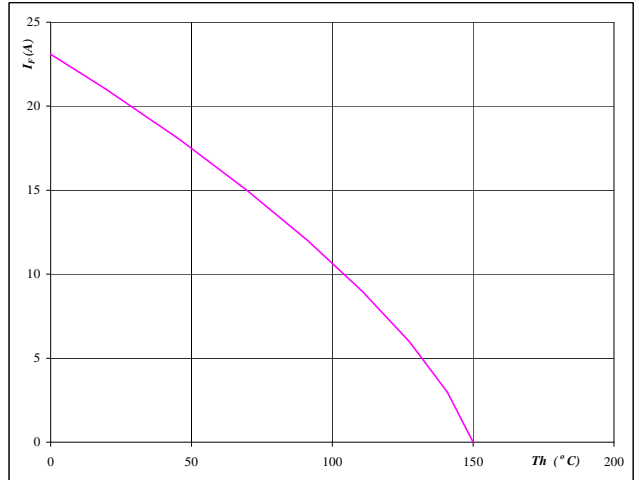
Power dissipation as a function of heatsink temperature

$$P_{tot} = f(T_h)$$

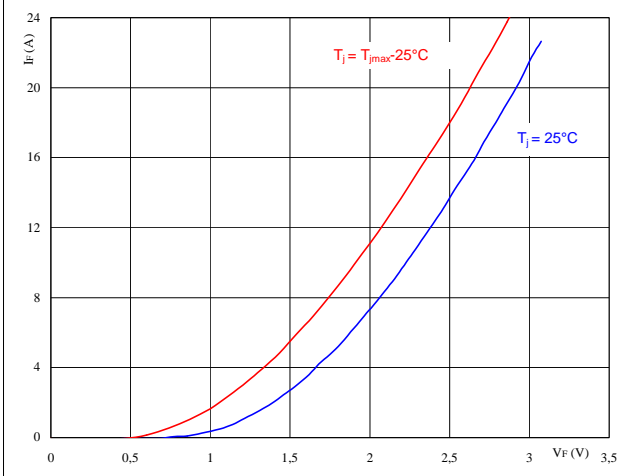

At
 $T_j = 150 \text{ °C}$
Figure 28 NP Inverse Diode

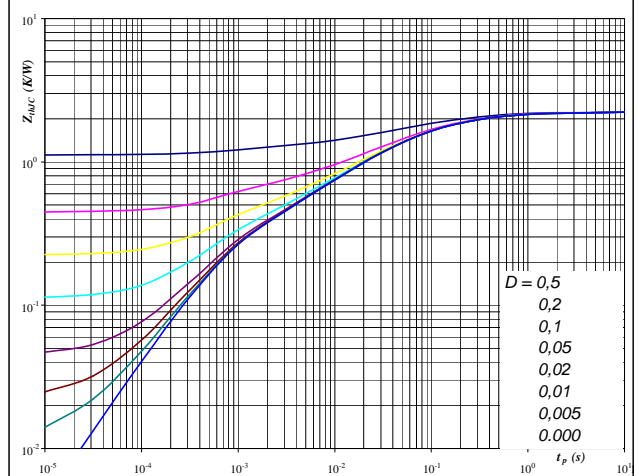
Forward current as a function of heatsink temperature

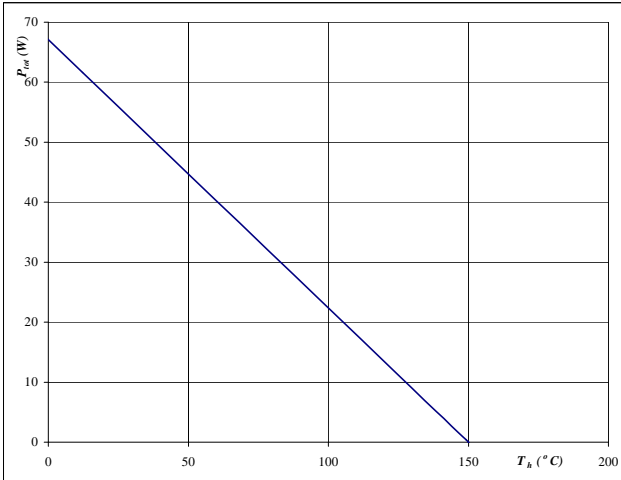
$$I_F = f(T_h)$$

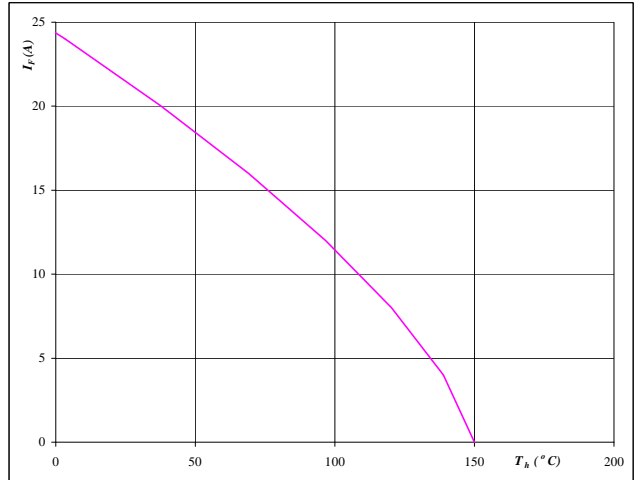

At
 $T_j = 150 \text{ °C}$

Half bridge Inverse Diode (D1, D4)
Figure 1 Halfbridge IGBT Inverse Diode

Typical FWD forward current as a function of forward voltage
 $I_F = f(V_F)$

At
 $t_p = 250 \mu s$
Figure 2 Halfbridge IGBT Inverse Diode

FWD transient thermal impedance as a function of pulse width
 $Z_{thJH} = f(t_p)$

At
 $D = t_p / T$
 $R_{thJH} = 2,235 \text{ K/W}$
Figure 3 Halfbridge IGBT Inverse Diode

Power dissipation as a function of heatsink temperature
 $P_{tot} = f(T_h)$

At
 $T_j = 150 \text{ °C}$
Figure 4 Halfbridge IGBT Inverse Diode

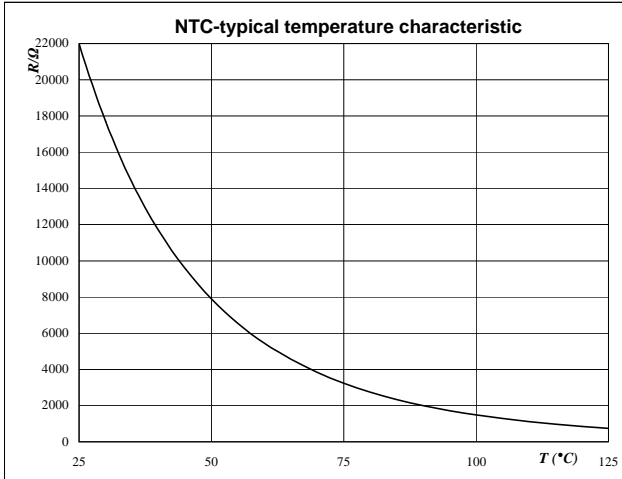
Forward current as a function of heatsink temperature
 $I_F = f(T_h)$

At
 $T_j = 150 \text{ °C}$

Thermistor

Figure 1 Thermistor

Typical NTC characteristic
 as a function of temperature

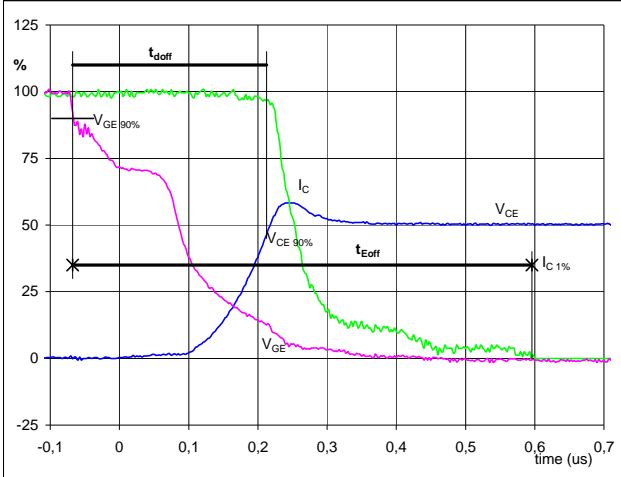
$$R_T = f(T)$$



Switching Definitions half bridge (T1, T4 / D7, D8)

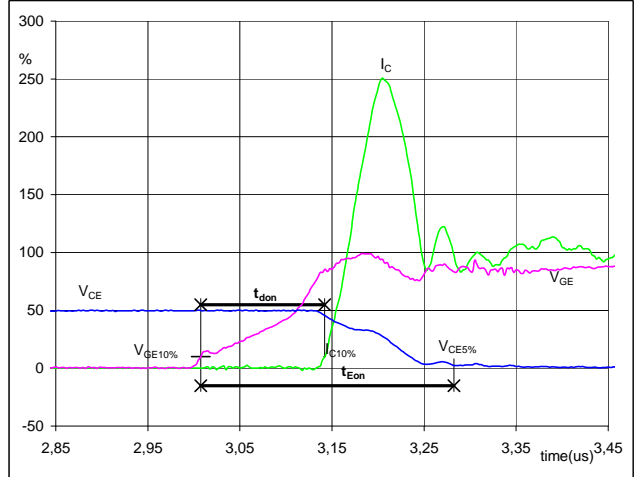
General conditions	
T_j	= 125 °C
R_{gon}	= 4 Ω
R_{goff}	= 4 Ω

Figure 1 half bridge IGBT

Turn-off Switching Waveforms & definition of t_{doff} , t_{Eoff}
 (t_{Eoff} = integrating time for E_{off})


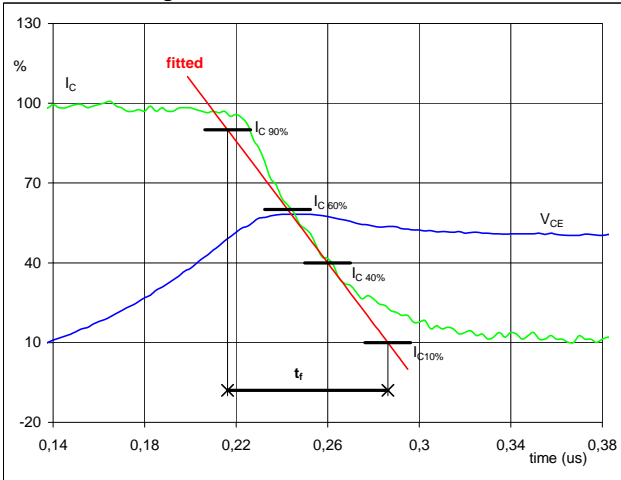
$V_{GE}(0\%) =$	-15	V
$V_{GE}(100\%) =$	15	V
$V_C(100\%) =$	700	V
$I_C(100\%) =$	100	A
$t_{doff} =$	0,28	μ s
$t_{Eoff} =$	0,66	μ s

Figure 2 half bridge IGBT

Turn-on Switching Waveforms & definition of t_{don} , t_{Eon}
 (t_{Eon} = integrating time for E_{on})


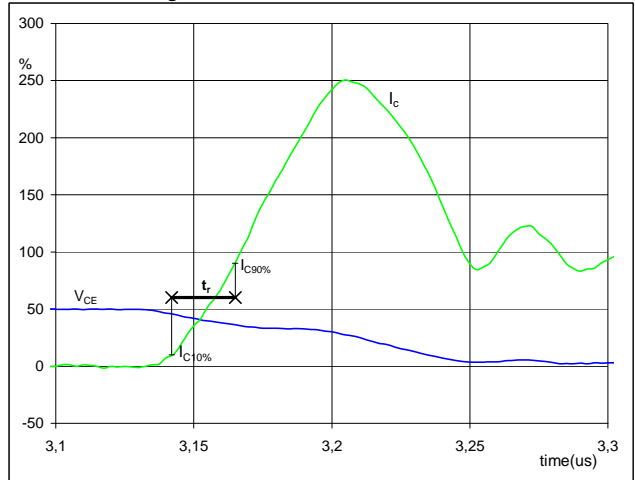
$V_{GE}(0\%) =$	-15	V
$V_{GE}(100\%) =$	15	V
$V_C(100\%) =$	700	V
$I_C(100\%) =$	100	A
$t_{don} =$	0,14	μ s
$t_{Eon} =$	0,27	μ s

Figure 3 half bridge IGBT

Turn-off Switching Waveforms & definition of t_f


$V_C(100\%) =$	700	V
$I_C(100\%) =$	100	A
$t_f =$	0,06	μ s

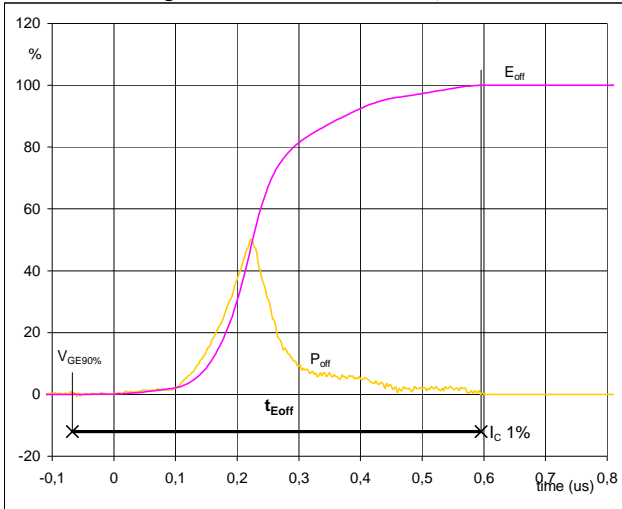
Figure 4 half bridge IGBT

Turn-on Switching Waveforms & definition of t_r


$V_C(100\%) =$	700	V
$I_C(100\%) =$	100	A
$t_r =$	0,02	μ s

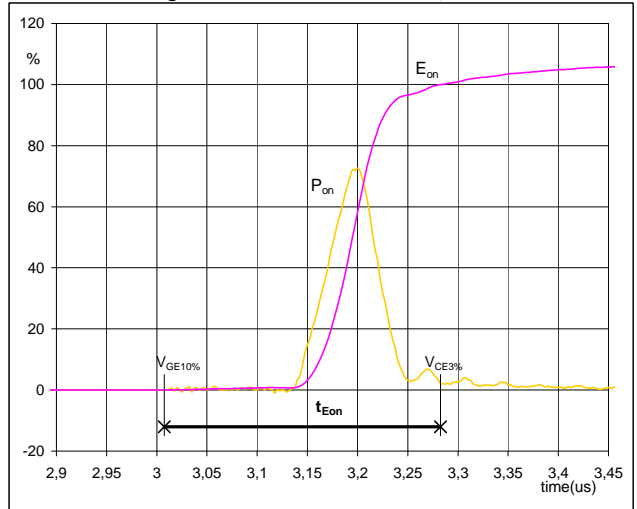
Switching Definitions half bridge (T1, T4 / D7, D8)

Figure 5 half bridge IGBT

Turn-off Switching Waveforms & definition of t_{Eoff}


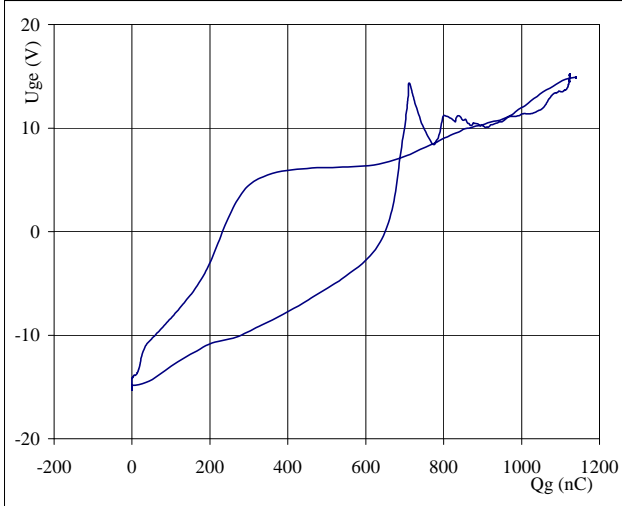
$P_{off} (100\%) =$	70,22	kW
$E_{off} (100\%) =$	4,03	mJ
$t_{Eoff} =$	0,66	μ s

Figure 6 half bridge IGBT

Turn-on Switching Waveforms & definition of t_{Eon}


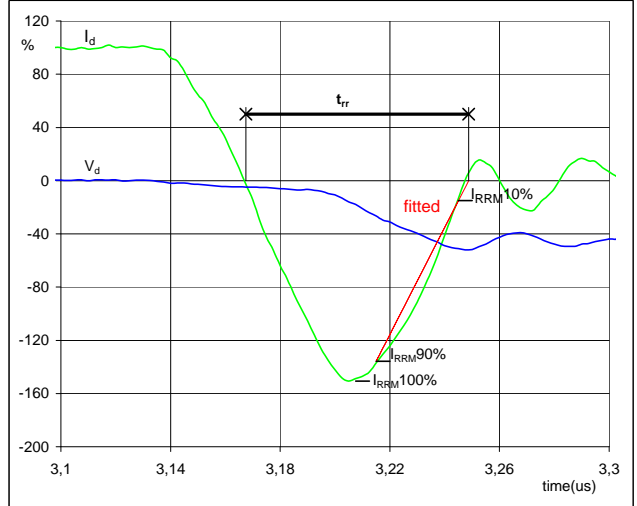
$P_{on} (100\%) =$	70,22	kW
$E_{on} (100\%) =$	3,18	mJ
$t_{Eon} =$	0,27	μ s

Figure 7 half bridge IGBT

Gate voltage vs Gate charge (measured)


$V_{GEoff} =$	-15	V
$V_{GEon} =$	15	V
$V_C (100\%) =$	700	V
$I_C (100\%) =$	100	A
$Q_g =$	1140,19	nC

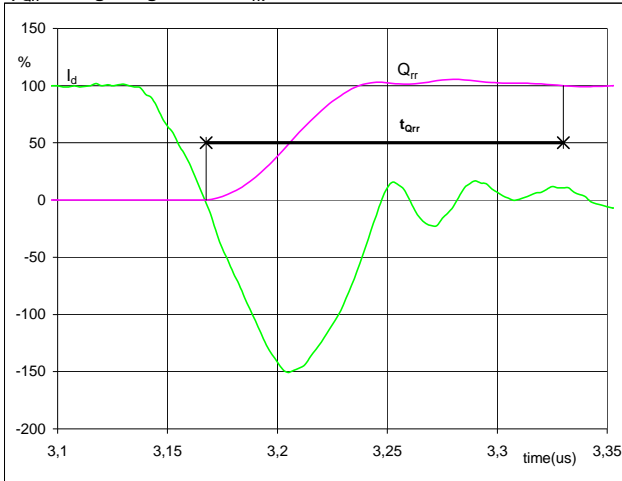
Figure 8 neutral point FWD

Turn-off Switching Waveforms & definition of t_{rr}


$V_d (100\%) =$	700	V
$I_d (100\%) =$	100	A
$I_{RRM} (100\%) =$	-151	A
$t_{rr} =$	0,08	μ s

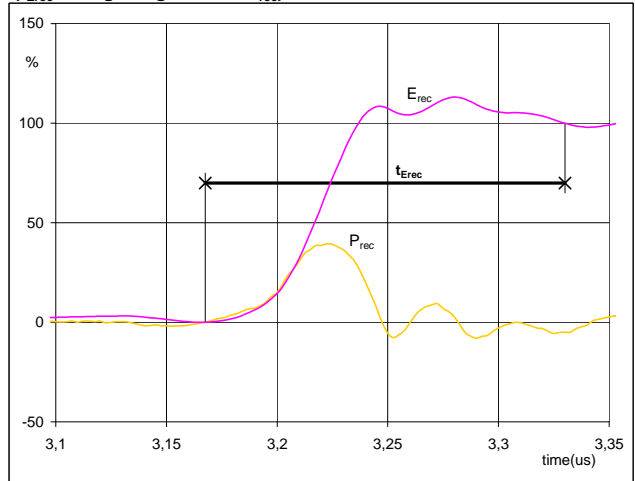
Switching Definitions half bridge (T1, T4 / D7, D8)

Figure 9 neutral point FWD

Turn-on Switching Waveforms & definition of t_{Qrr}
 (t_{Qrr} = integrating time for Q_{rr})


I_d (100%) =	100	A
Q_{rr} (100%) =	7,13	μC
t_{Qrr} =	0,16	μs

Figure 10 neutral point FWD

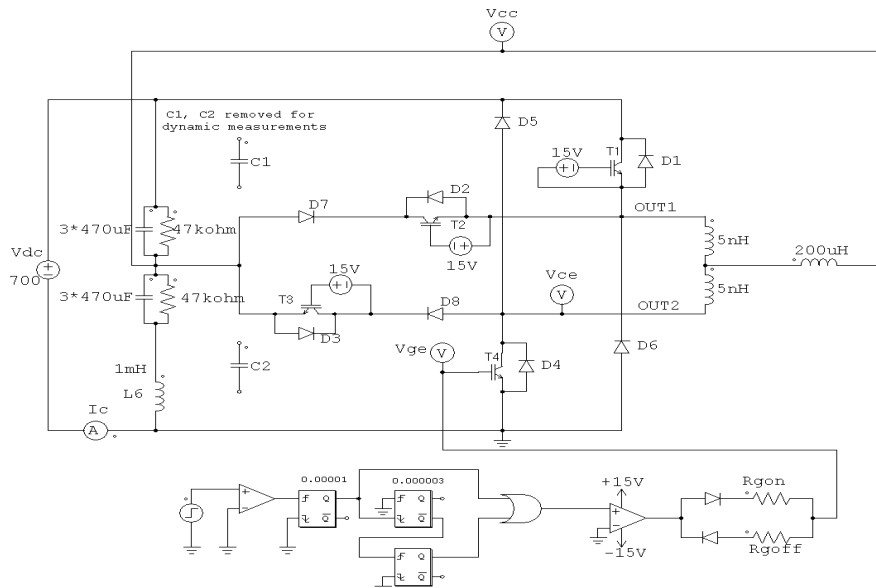
Turn-on Switching Waveforms & definition of t_{Erec}
 (t_{Erec} = integrating time for E_{rec})


P_{rec} (100%) =	70,22	kW
E_{rec} (100%) =	1,01	mJ
t_{Erec} =	0,16	μs

half bridge switching measurement circuit (T1, T4 / D7, D8)

Figure 11

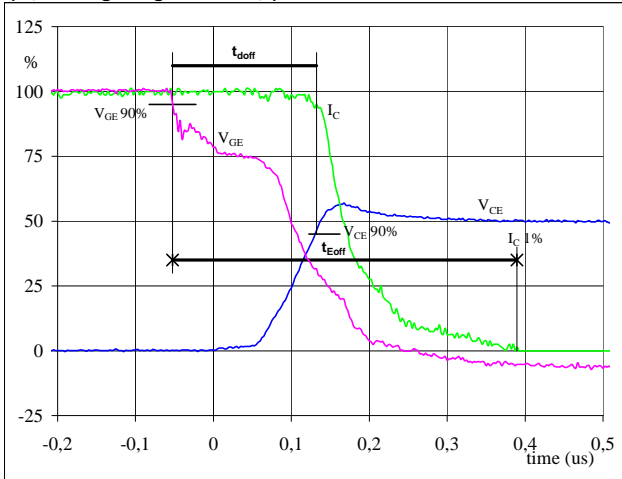
half bridge IGBT



Switching Definitions neutral point IGBT (T2, T3 / D5, D6)

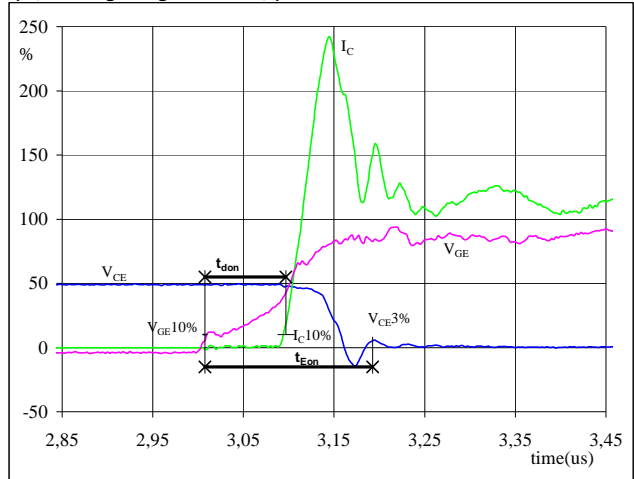
General conditions	
T_j	= 125 °C
R_{gon}	= 4 Ω
R_{goff}	= 4 Ω

Figure 1 neutral point IGBT

Turn-off Switching Waveforms & definition of t_{doff} , t_{Eoff}
 (t_{Eoff} = integrating time for E_{off})


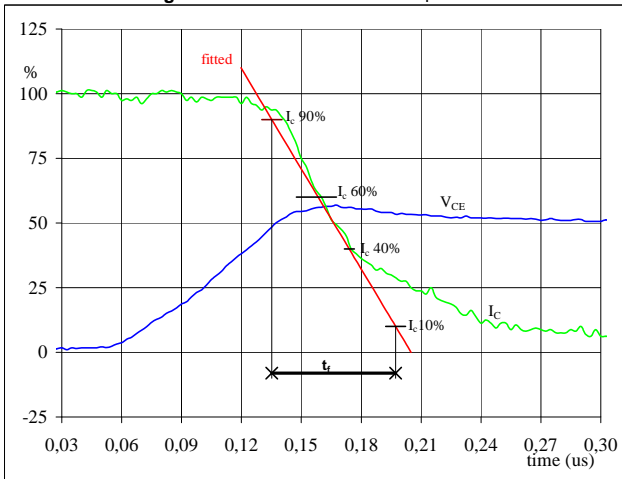
$V_{GE}(0\%) =$	-15	V
$V_{GE}(100\%) =$	15	V
$V_C(100\%) =$	700	V
$I_C(100\%) =$	100	A
$t_{doff} =$	0,18	μs
$t_{Eoff} =$	0,44	μs

Figure 2 neutral point IGBT

Turn-on Switching Waveforms & definition of t_{don} , t_{Eon}
 (t_{Eon} = integrating time for E_{on})


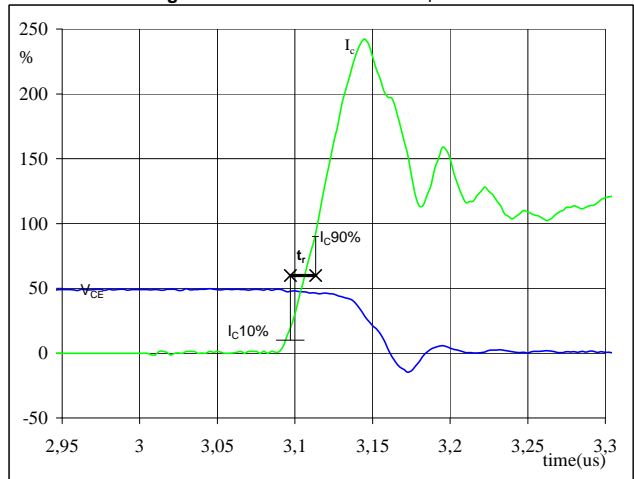
$V_{GE}(0\%) =$	-15	V
$V_{GE}(100\%) =$	15	V
$V_C(100\%) =$	700	V
$I_C(100\%) =$	100	A
$t_{don} =$	0,10	μs
$t_{Eon} =$	0,18	μs

Figure 3 neutral point IGBT

Turn-off Switching Waveforms & definition of t_f


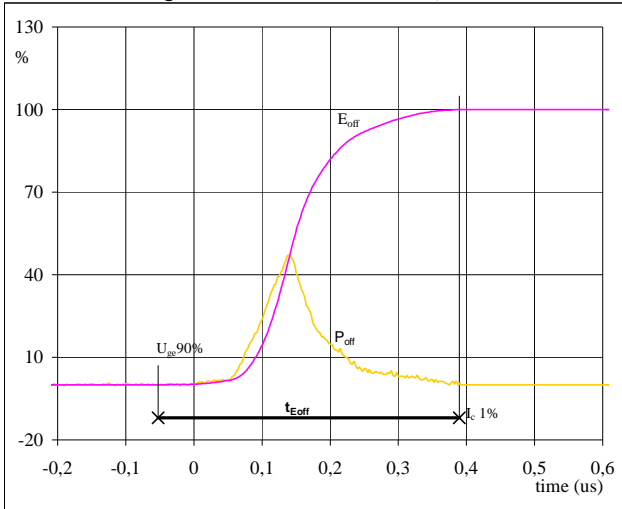
$V_C(100\%) =$	700	V
$I_C(100\%) =$	100	A
$t_f =$	0,064	μs

Figure 4 neutral point IGBT

Turn-on Switching Waveforms & definition of t_r


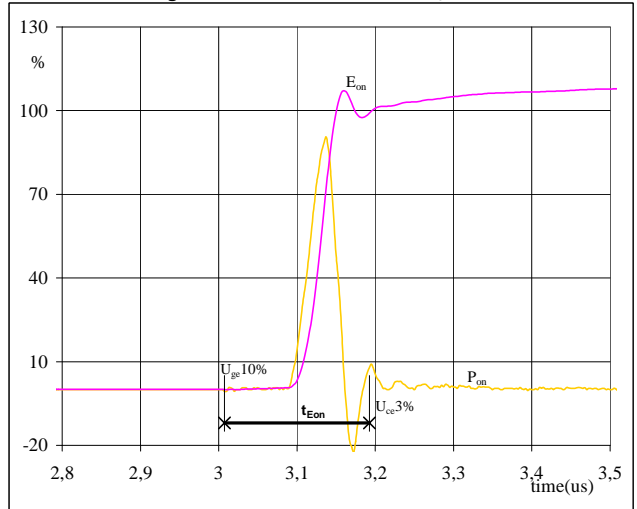
$V_C(100\%) =$	700	V
$I_C(100\%) =$	100	A
$t_r =$	0,019	μs

Switching Definitions neutral point IGBT (T2, T3 / D5, D6)
Figure 5 neutral point IGBT

Turn-off Switching Waveforms & definition of t_{Eoff}


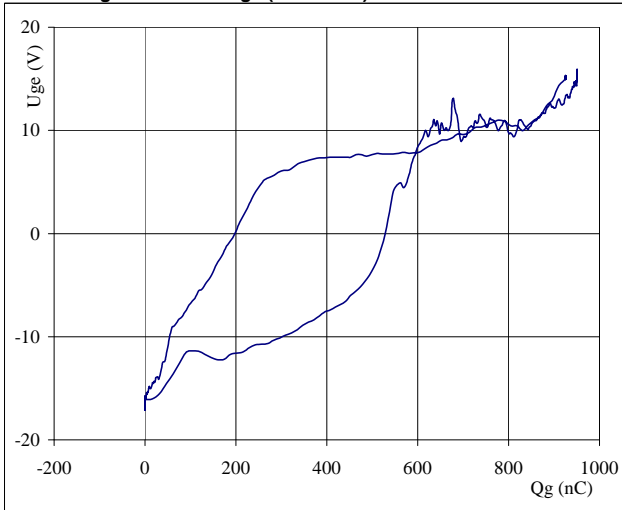
$P_{off}(100\%) = 69,93 \text{ kW}$
 $E_{off}(100\%) = 3,32 \text{ mJ}$
 $t_{Eoff} = 0,44 \text{ }\mu\text{s}$

Figure 6 neutral point IGBT

Turn-on Switching Waveforms & definition of t_{Eon}


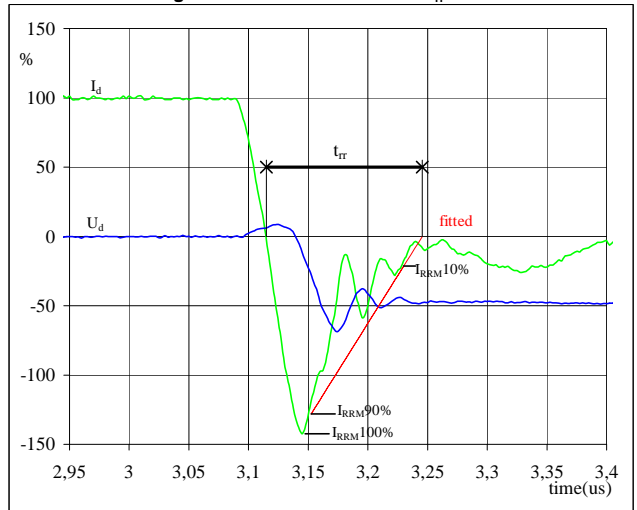
$P_{on}(100\%) = 69,9279 \text{ kW}$
 $E_{on}(100\%) = 1,52 \text{ mJ}$
 $t_{Eon} = 0,18 \text{ }\mu\text{s}$

Figure 7 neutral point IGBT

Gate voltage vs Gate charge (measured)


$V_{GEoff} = -15 \text{ V}$
 $V_{GEon} = 15 \text{ V}$
 $V_C(100\%) = 700 \text{ V}$
 $I_C(100\%) = 100 \text{ A}$
 $Q_g = 950,59 \text{ nC}$

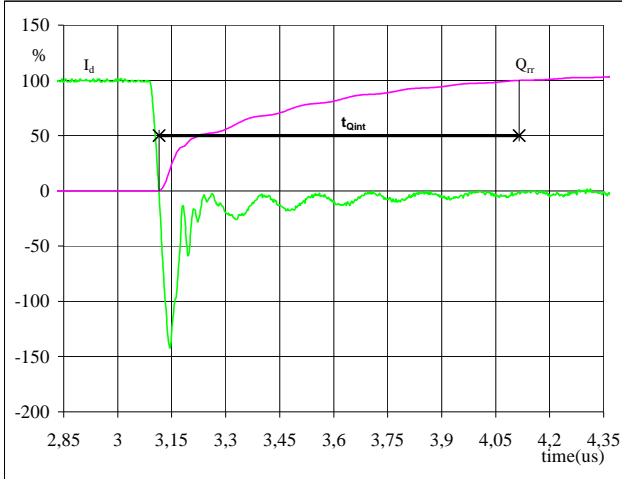
Figure 8 half bridge FWD

Turn-off Switching Waveforms & definition of t_{rr}


$V_d(100\%) = 700 \text{ V}$
 $I_d(100\%) = 100 \text{ A}$
 $I_{RRM}(100\%) = -142 \text{ A}$
 $t_{rr} = 0,07 \text{ }\mu\text{s}$

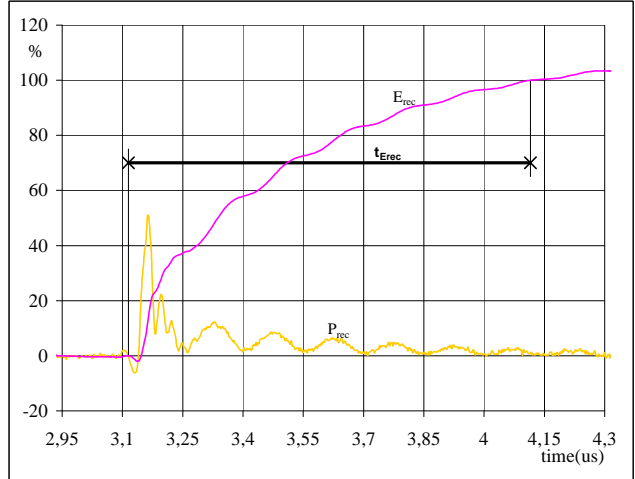
Switching Definitions neutral point IGBT (T2, T3 / D5, D6)

Figure 9 half bridge FWD

Turn-on Switching Waveforms & definition of t_{Qrr}
 (t_{Qrr} = integrating time for Q_{rr})


I_d (100%) =	100	A
Q_{rr} (100%) =	12,71	μC
t_{Qint} =	1,00	μs

Figure 10 half bridge FWD

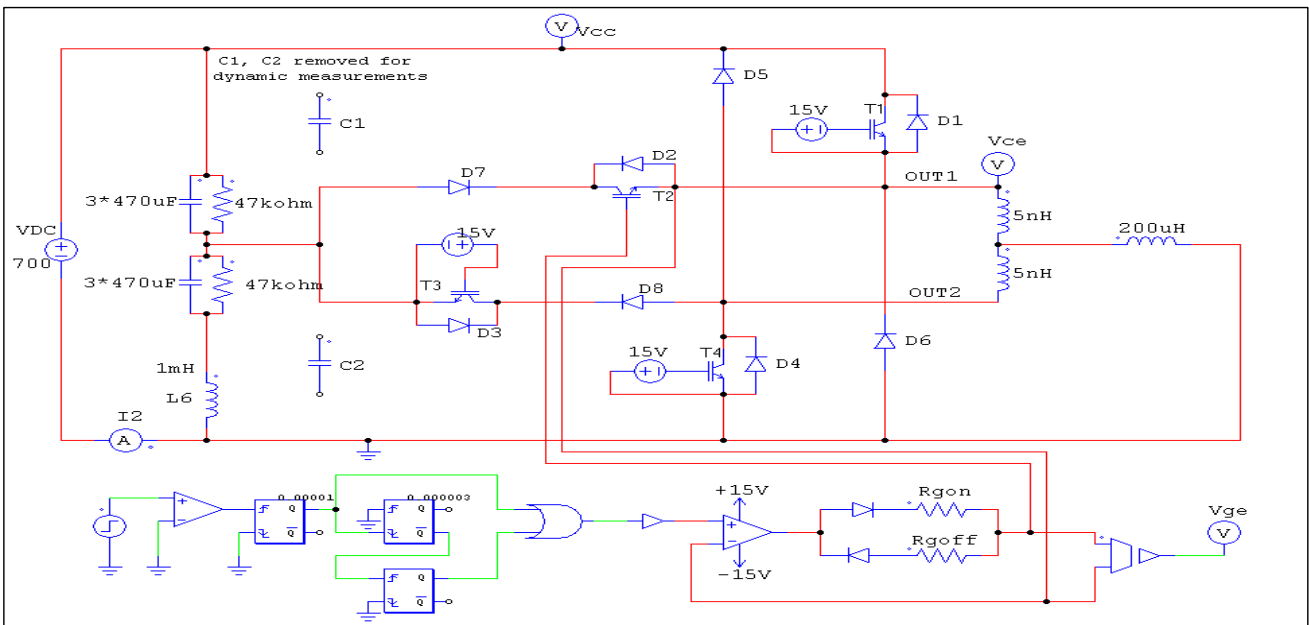
Turn-on Switching Waveforms & definition of t_{Erec}
 (t_{Erec} = integrating time for E_{rec})


P_{rec} (100%) =	69,93	kW
E_{rec} (100%) =	3,61	mJ
t_{Erec} =	1,00	μs

neutral point IGBT switching measurement circuit (T2, T3 / D5, D6)

Figure 11

neutral point IGBT

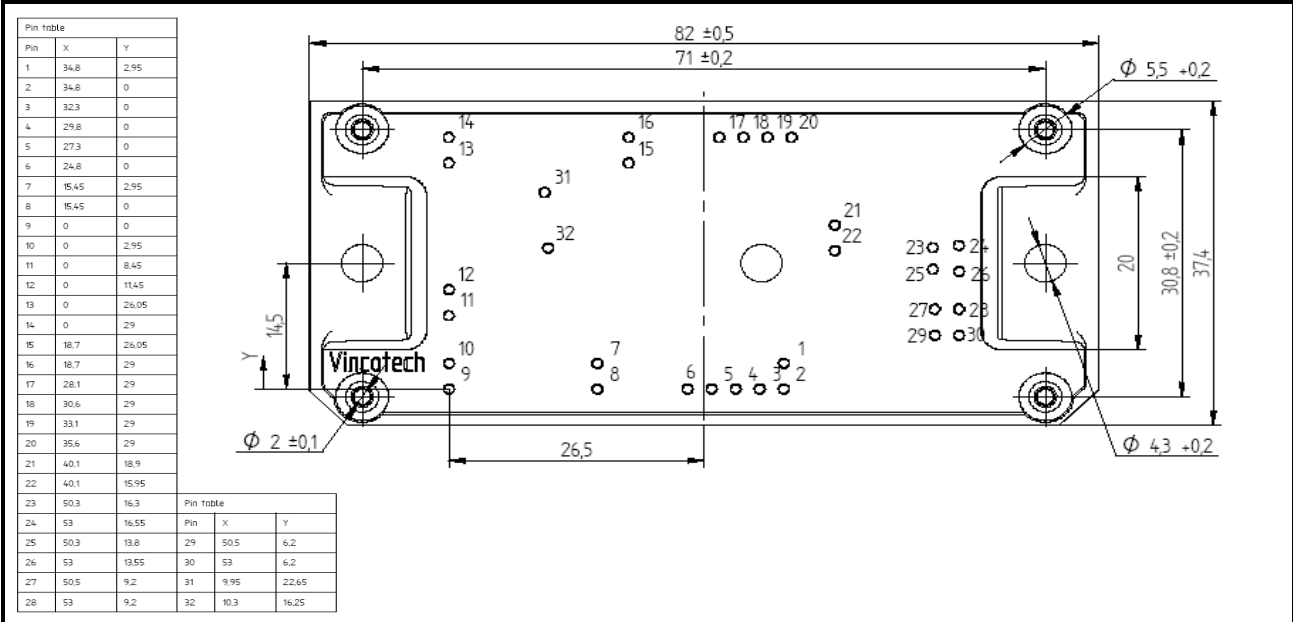


Ordering Code and Marking - Outline - Pinout

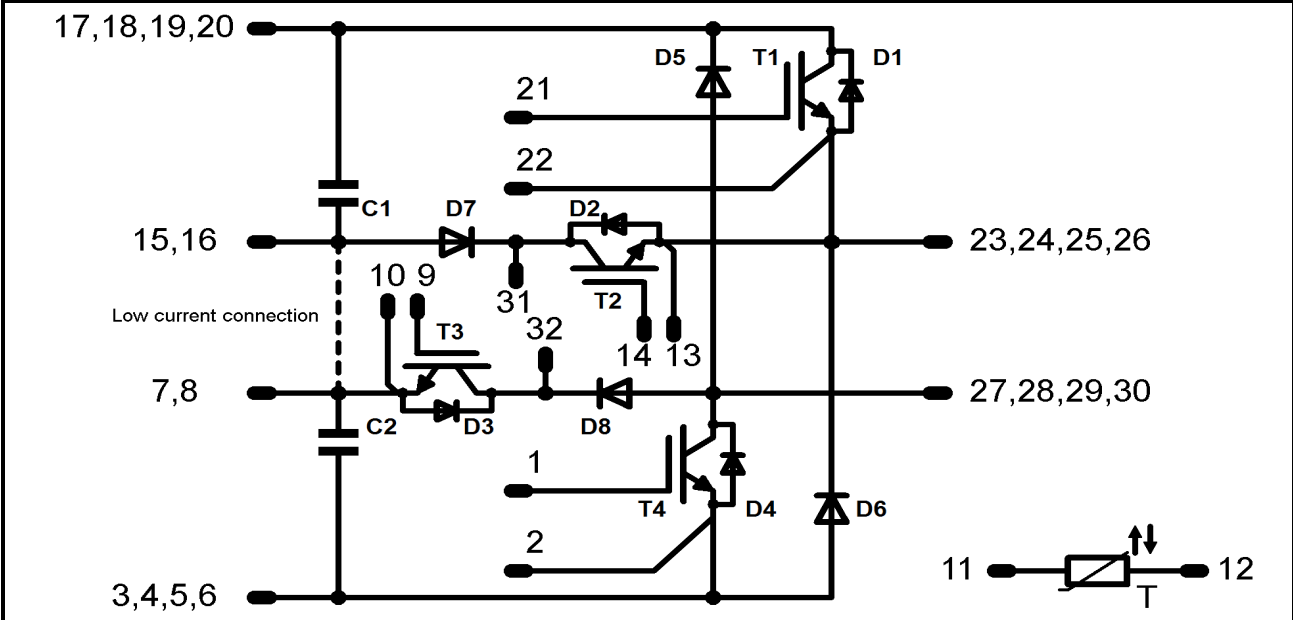
Ordering Code & Marking

Version	Ordering Code	in DataMatrix as	in packaging barcode as
without thermal paste 12mm housing	10-FY12NMA160SH-M420F	M420F	M420F
without thermal paste 12mm housing with PressFIT	10-PY12NMA160SH-M420FY	M420FY	M420FY
with phase change thermal paste 12mm housing	10-FY12NMA160SH-M420F-/3/	M420F	M420F
with phase change thermal paste 12mm housing with PressFIT	10-PY12NMA160SH-M420FY-/3/	M420FY	M420FY

Outline



Pinout



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2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.